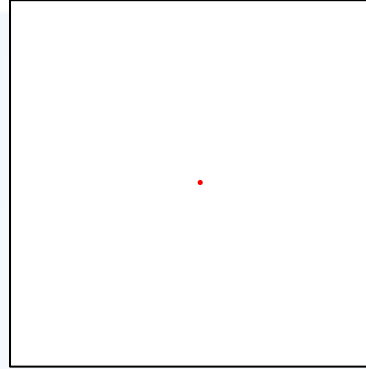


# RFocus: Beamforming using 1000s of Passive Antennas

Venkat Arun, Hari Balakrishnan  
CSAIL, MIT

Transmitter



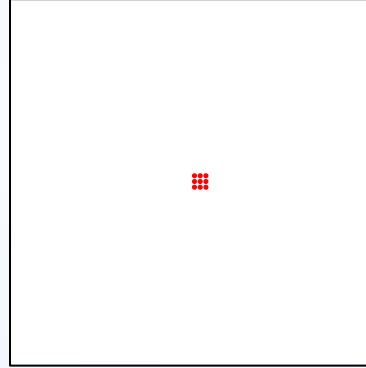
# 1 Antenna

Goal: Maximize signal strength at receiver



Receiver

Transmitter

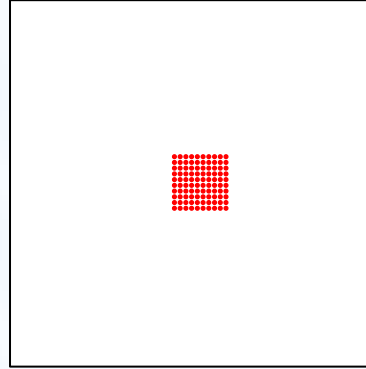


# 9 Antennas



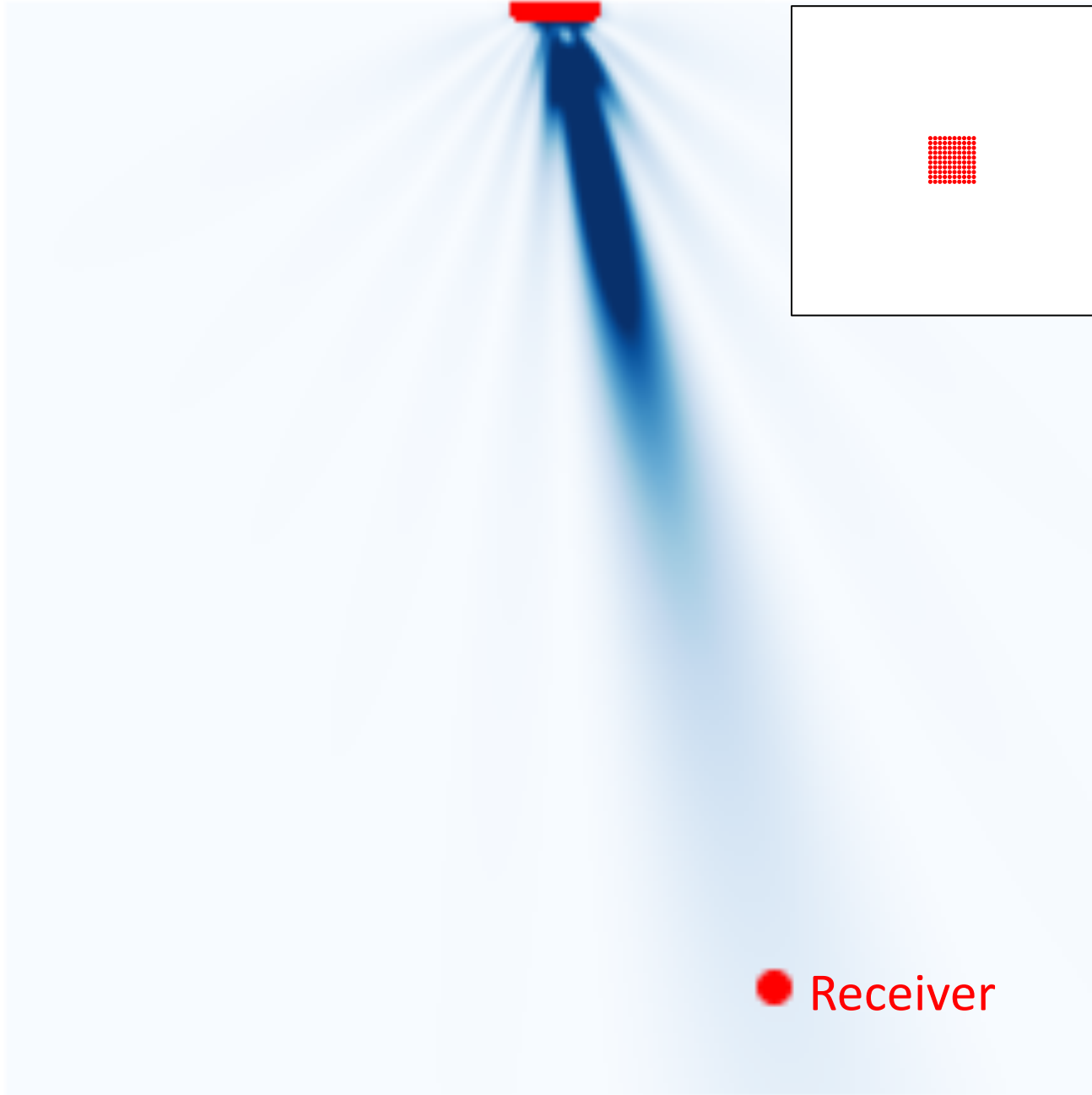
● Receiver

Transmitter



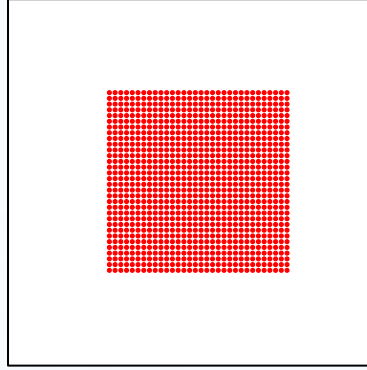
100 Antennas

Receiver





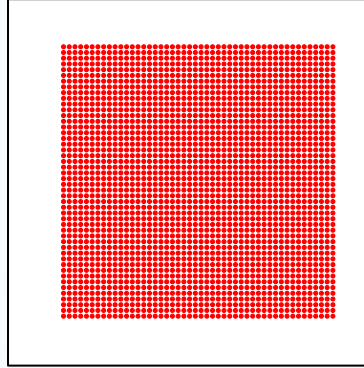
Transmitter



# 1024 Antennas

● Receiver

Transmitter

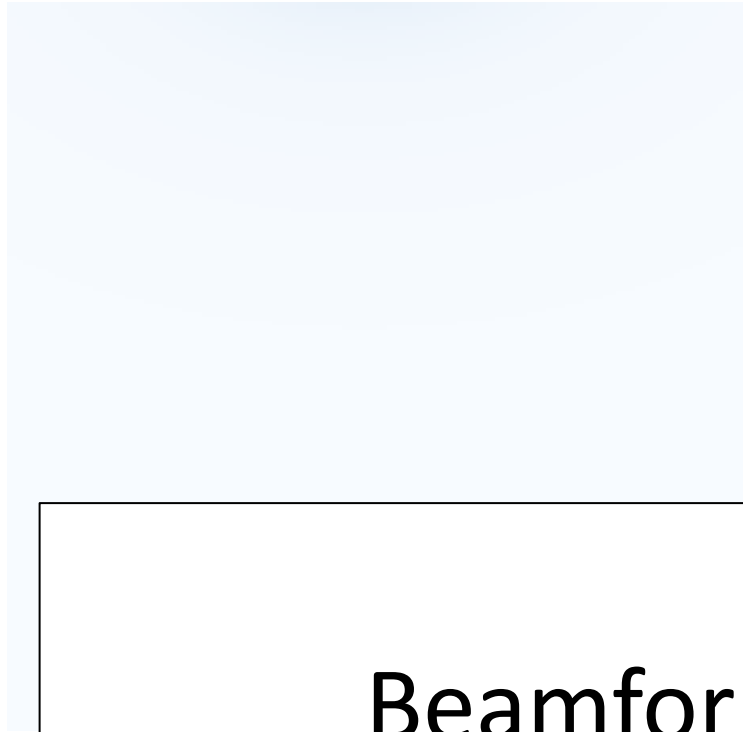


3136 Antennas

Receiver



1 Antenna



100 Antennas



3136 Antennas

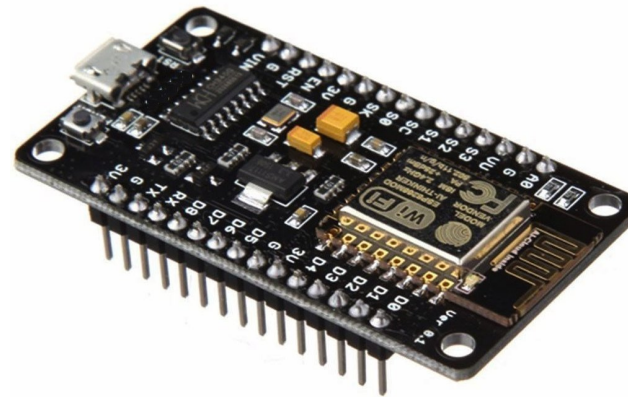
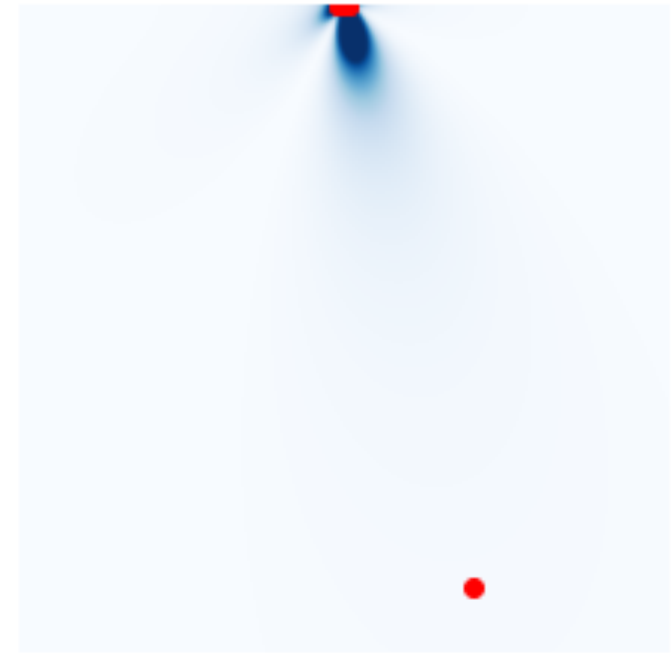
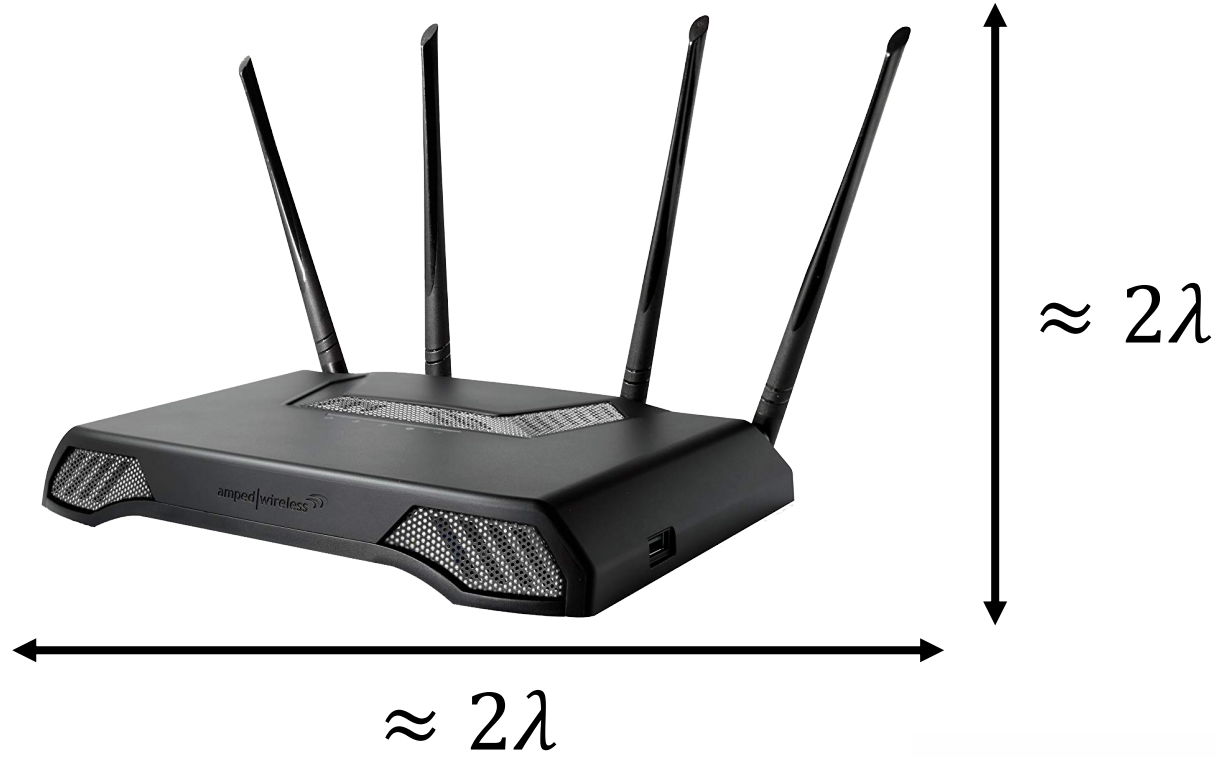


Beamforming ability is a function of the number of wavelengths the device spans

Beamforming ability is a function of the number of wavelengths the device spans

- Directional antennas can only be as directional as their size allows
- Squeezing more antennas into a smaller space doesn't help

< 10 Antennas



More antennas won't fit in our devices

The environment is already big.  
Let's put antennas there!



A photograph of an empty office hallway. The ceiling is a white grid with recessed lights and a smoke detector. The walls are a light beige color. On the right, there is a large window with multiple panes and a wooden frame. Below the window is a long white desk or counter. The floor is covered in blue carpeting. In the background, there is a doorway leading to another room, and a door with a fire exit sign above it. The image is used to illustrate the components of a building's interior: Ceiling, Walls, and Carpets.

Ceiling

Walls

Carpets



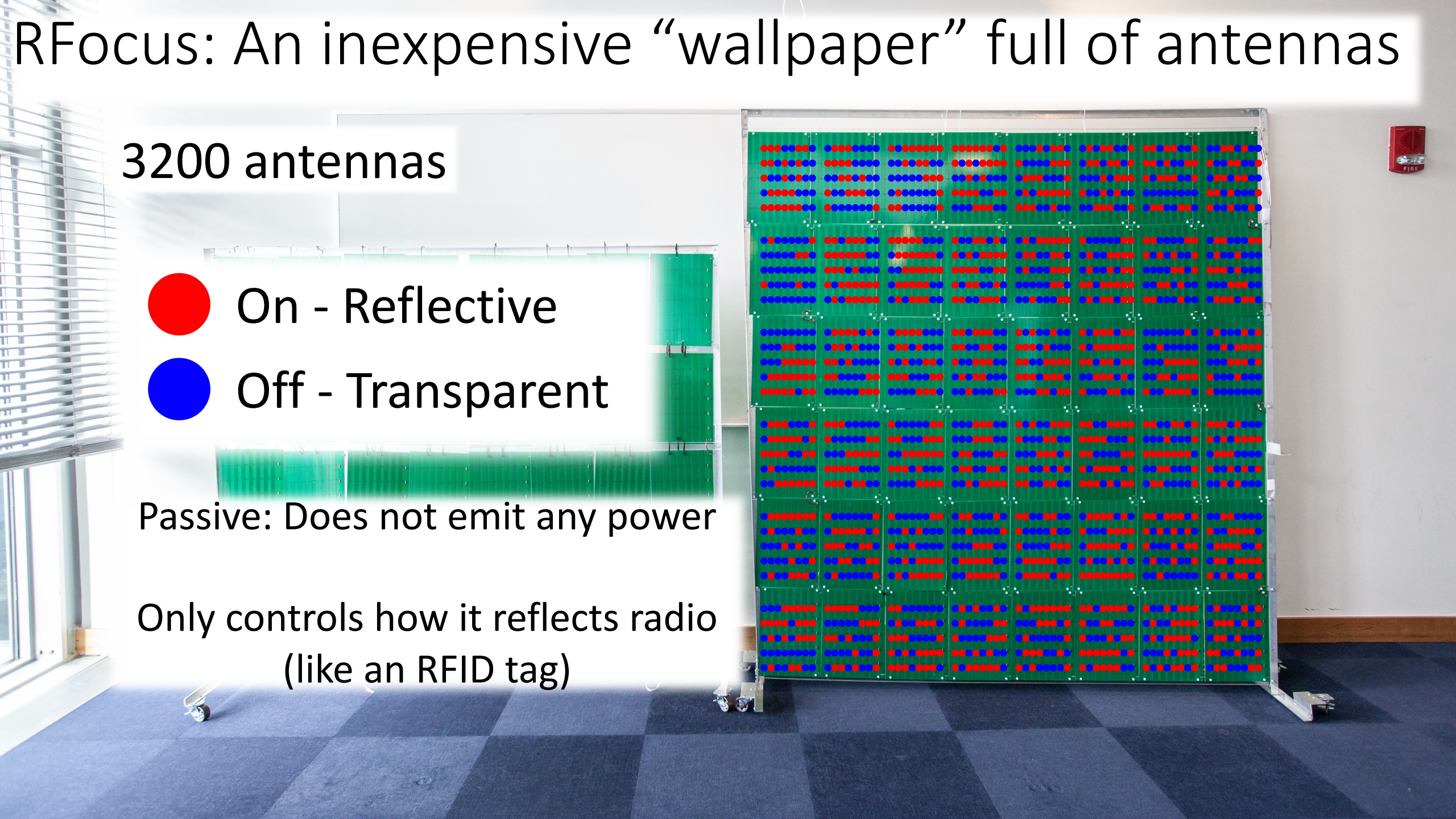
# RFocus: An inexpensive “wallpaper” full of antennas

3200 antennas

- On - Reflective
- Off - Transparent

Passive: Does not emit any power

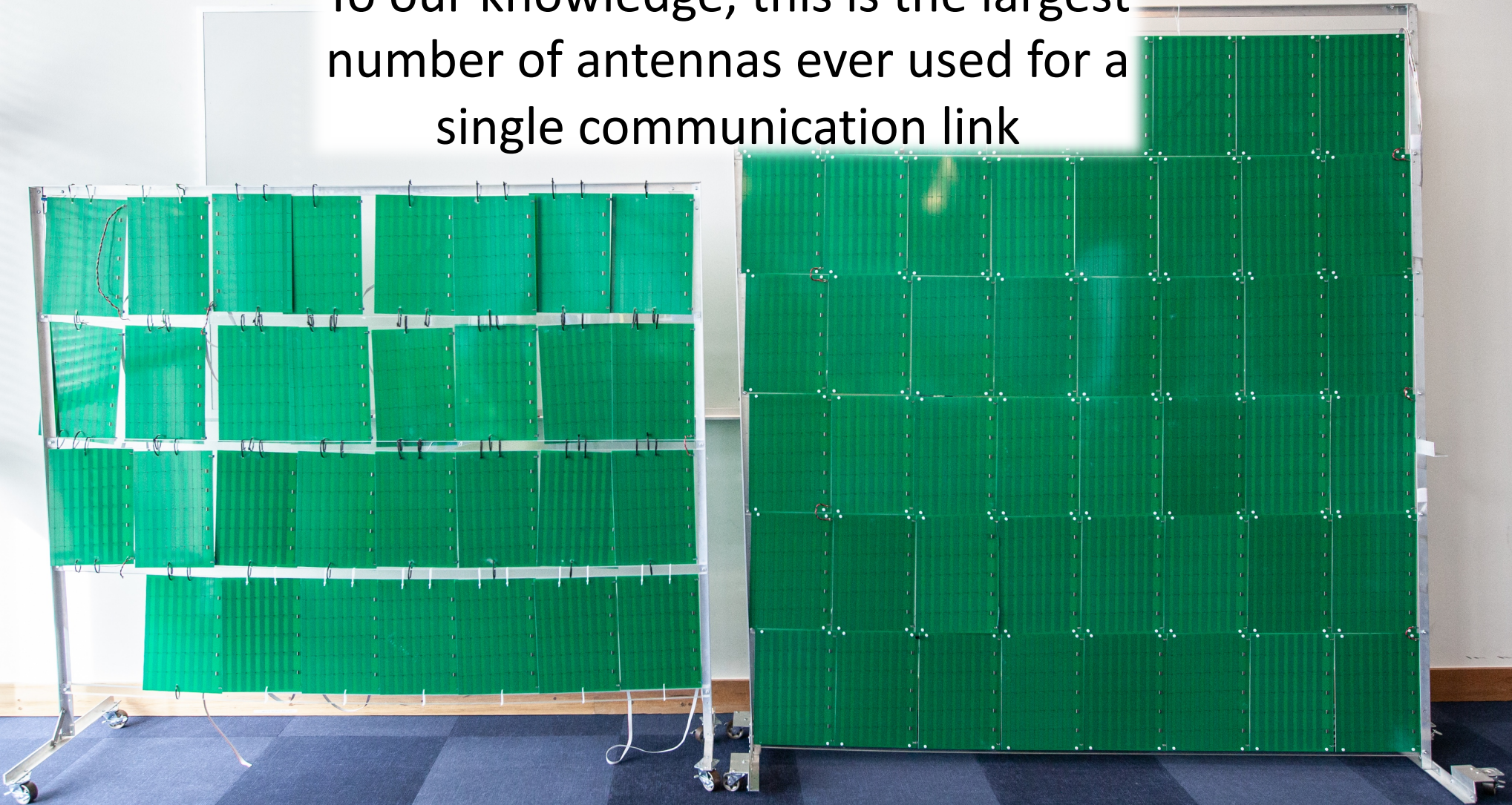
Only controls how it reflects radio  
(like an RFID tag)





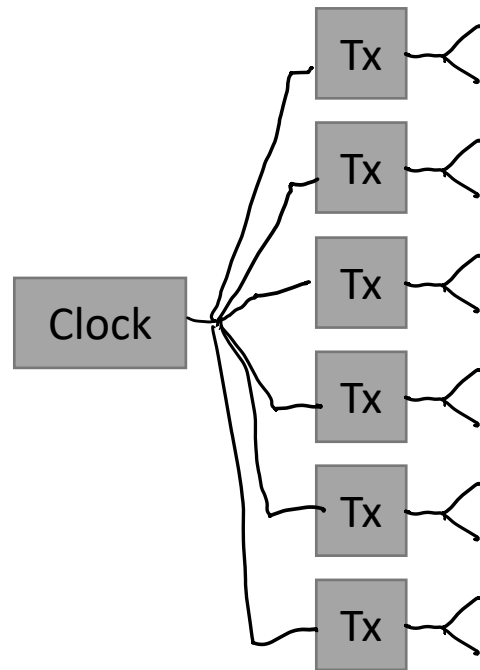
## 3200 Antennas

To our knowledge, this is the largest number of antennas ever used for a single communication link

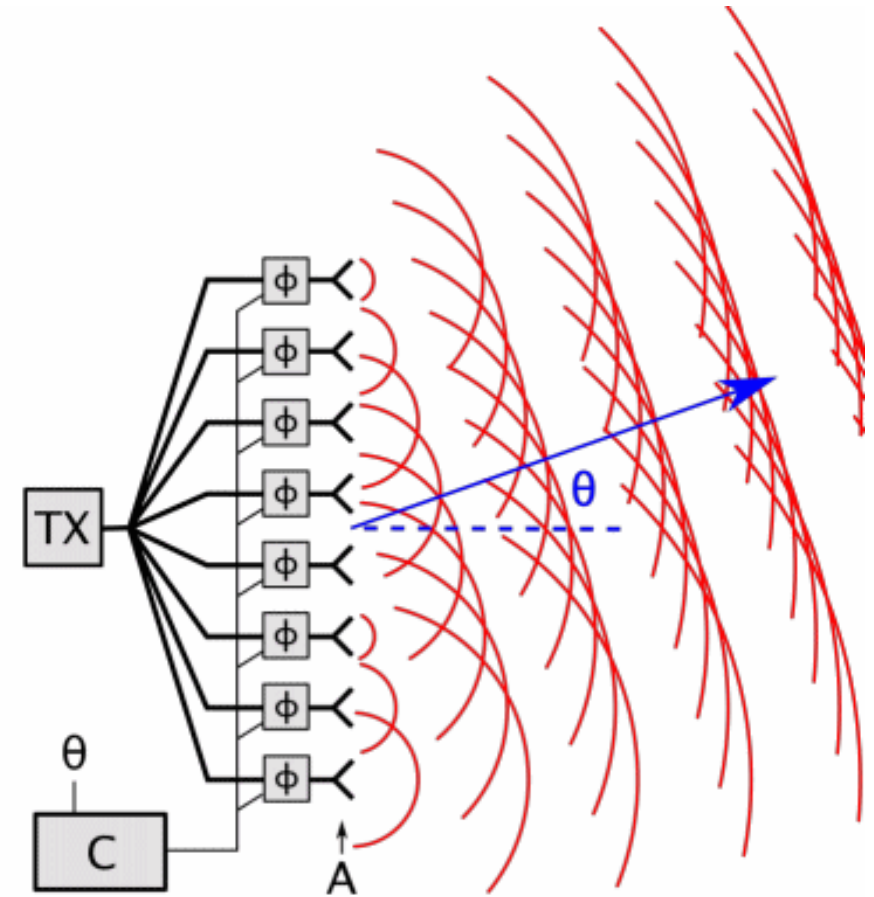




# Other ways to have $>1$ antenna



Independent RF chains

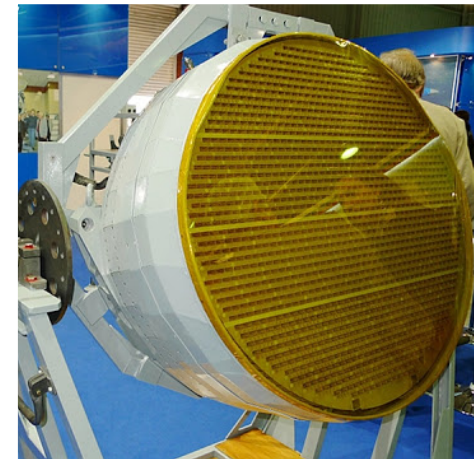


Phased Array

# Why not phased arrays?

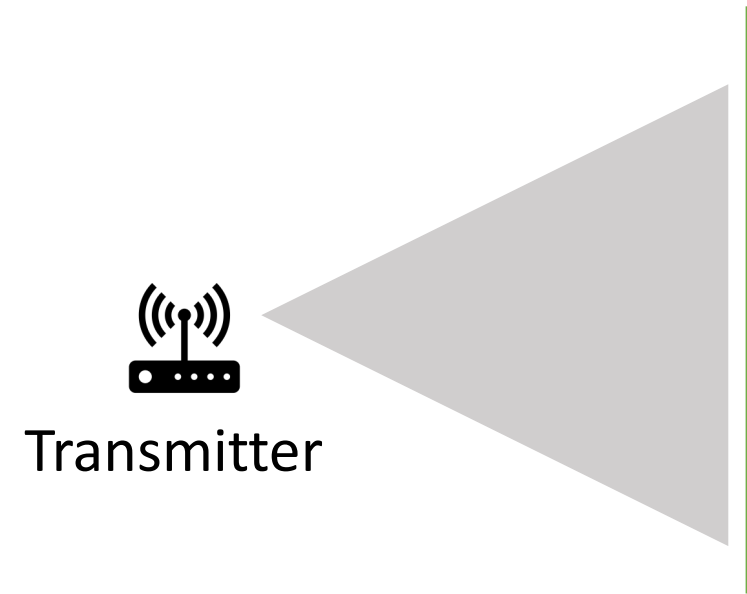
- Needs a splitter network to supply 1000s of antennas.
  - Coax cables or waveguides: expensive and bulky
  - PCB trace transmission line: lossy
  - Cannot be paper thin
- RFocus has a nicer deployment model
  - Endpoints need not be connected to RFocus
  - Allows RFocus to be pre-embedded in the environment. It can even be sold as disconnected pieces, e.g. in carpet squares
  - Currently RFocus' antenna switches are controlled using (low-speed) wires  
In the future, they can be powered and controlled wirelessly like RFID tags

# Radars use large phased arrays today



# RFocus is a phased array

- RFocus is a phased array that uses the air as a splitter network
- At least one of the transmitter and the receiver needs to be close to the surface
- In our case, the loss is 10 dB. This is comparable to a PCB trace transmission line





Target

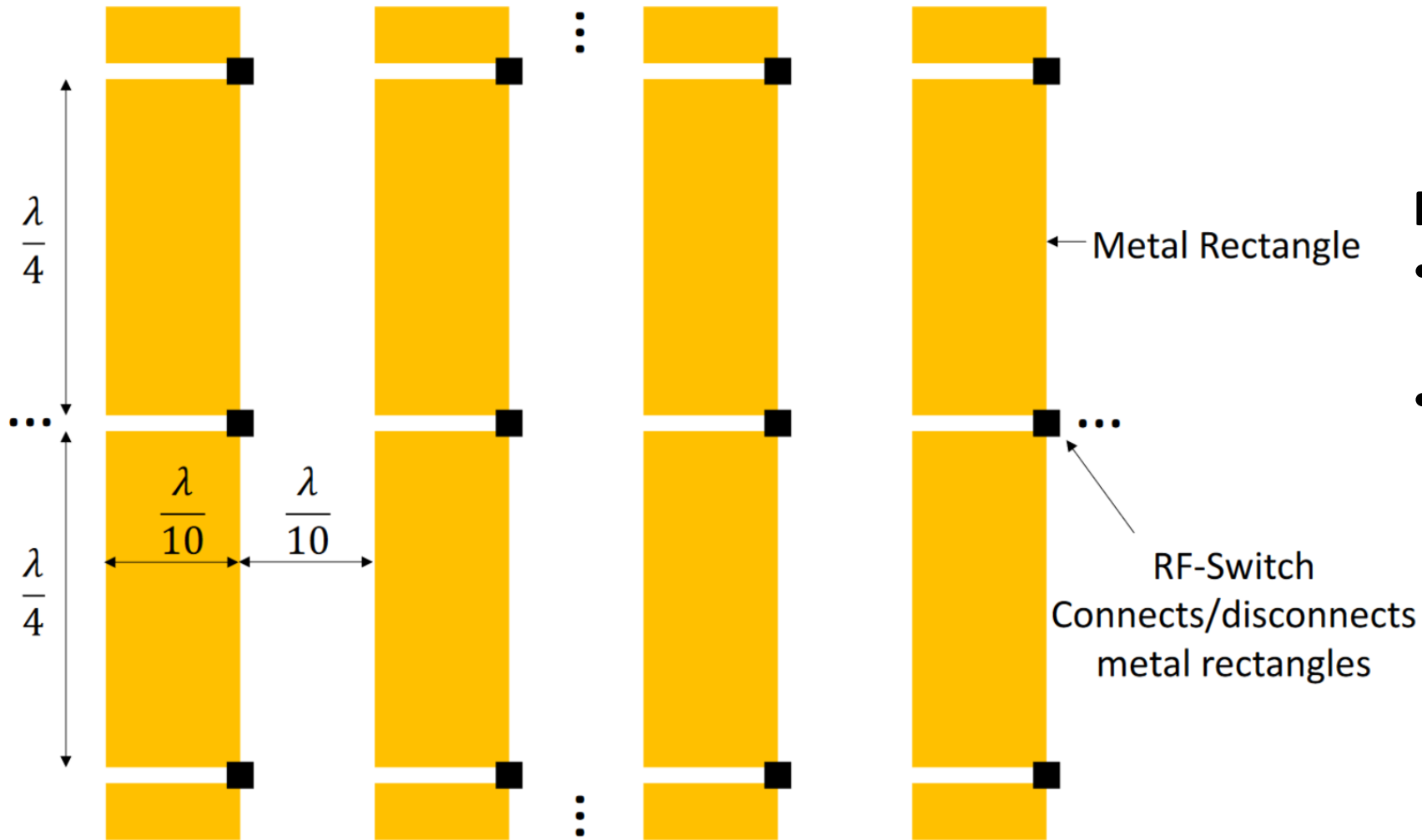
.







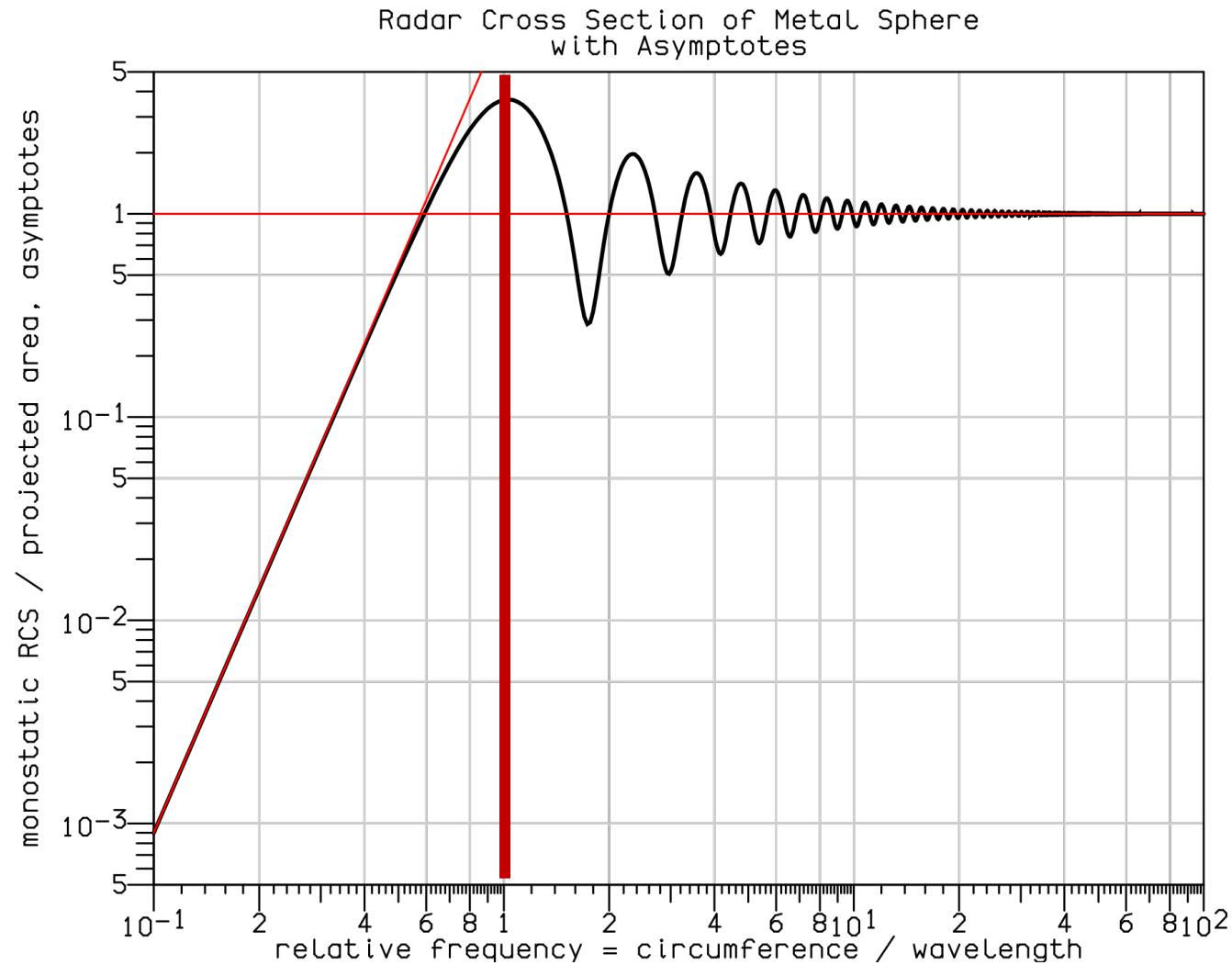
# How RFocus works



## Principles of Operation:

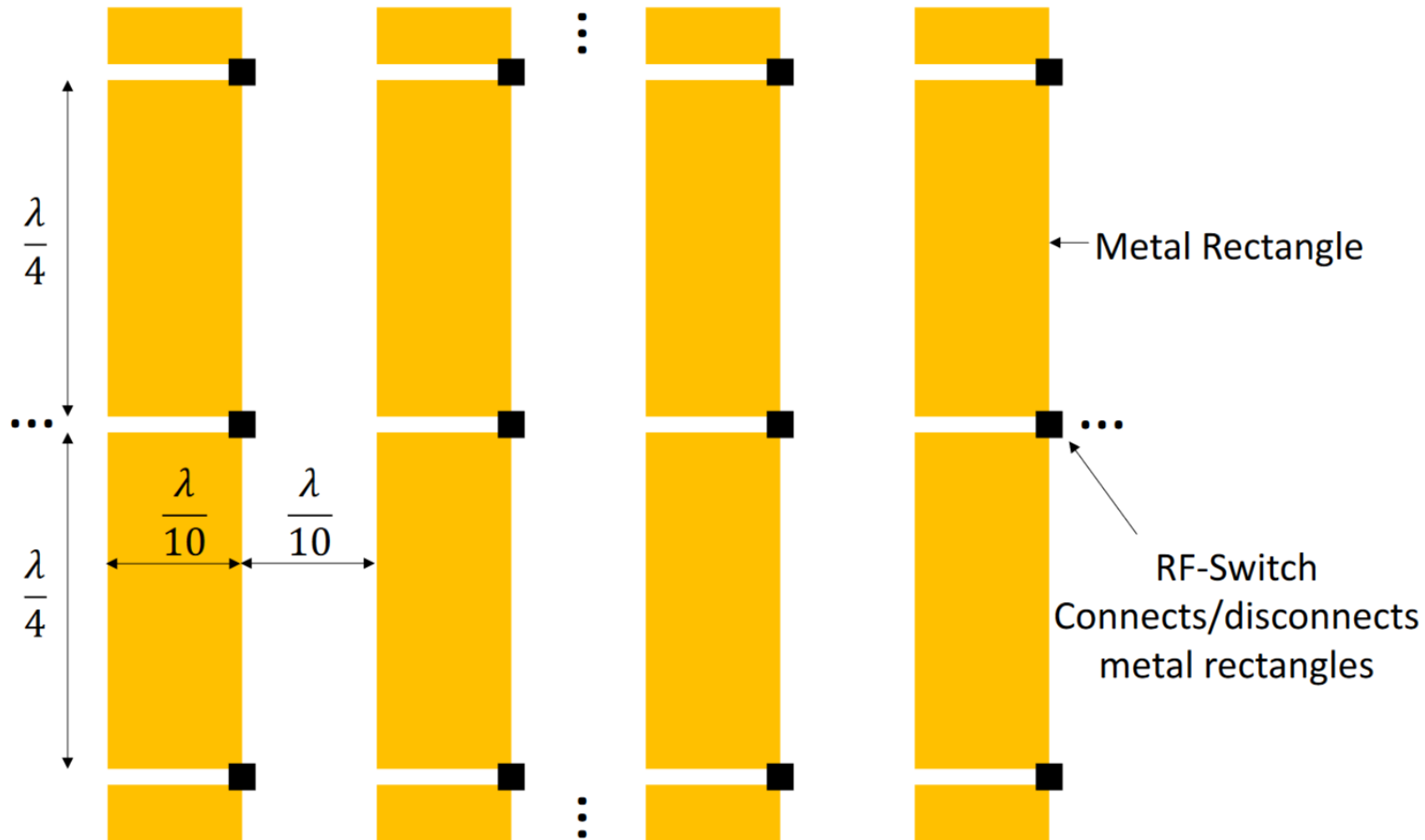
- Small objects are “invisible” to radio
- Small holes in a sheet of metal are ignored

# Small objects are “invisible” to radio



Mie scattering from a sphere

# How Rfocus works



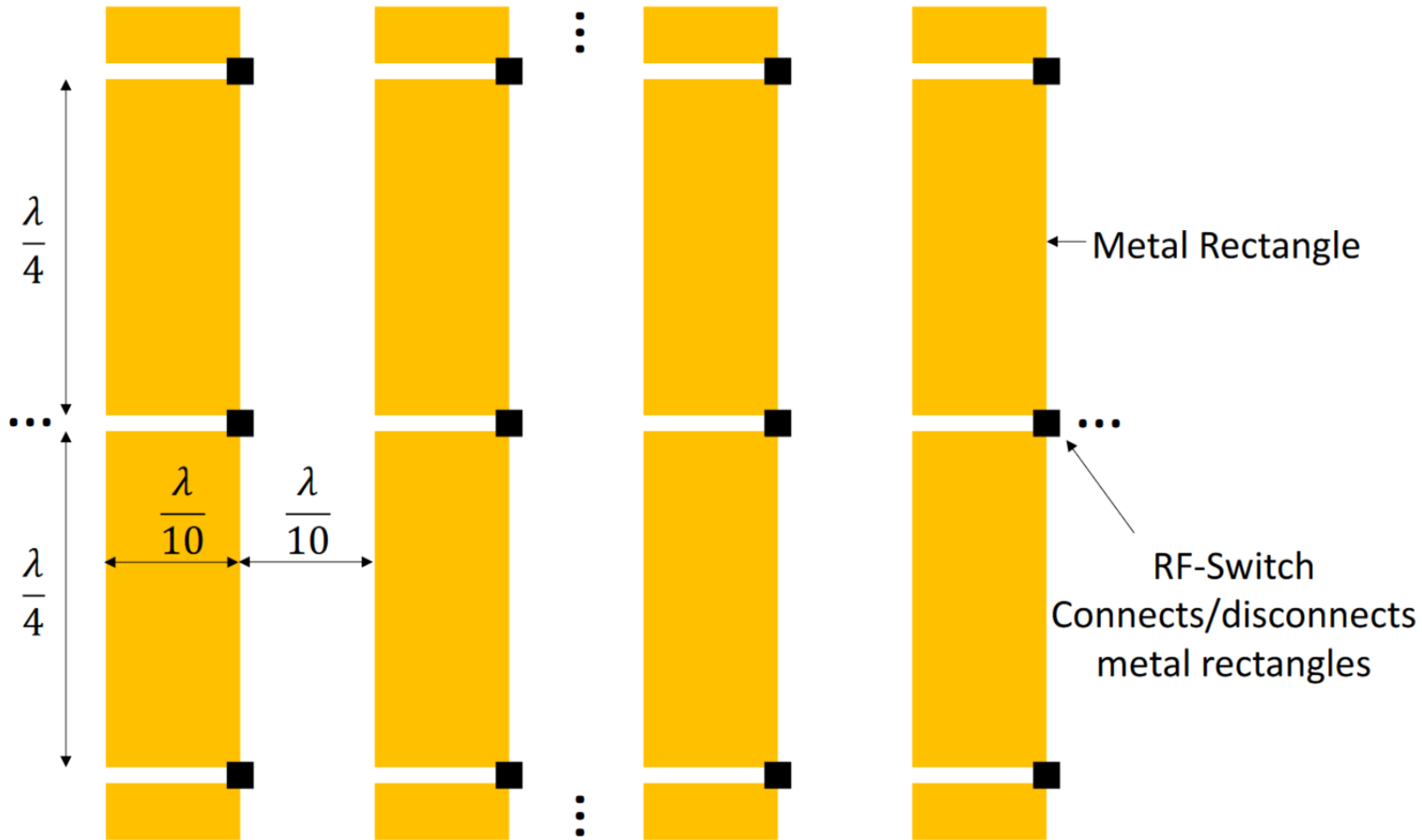
When connected, the pieces of metal are larger

When disconnected, it is (more) transparent

# Radio ignores small holes

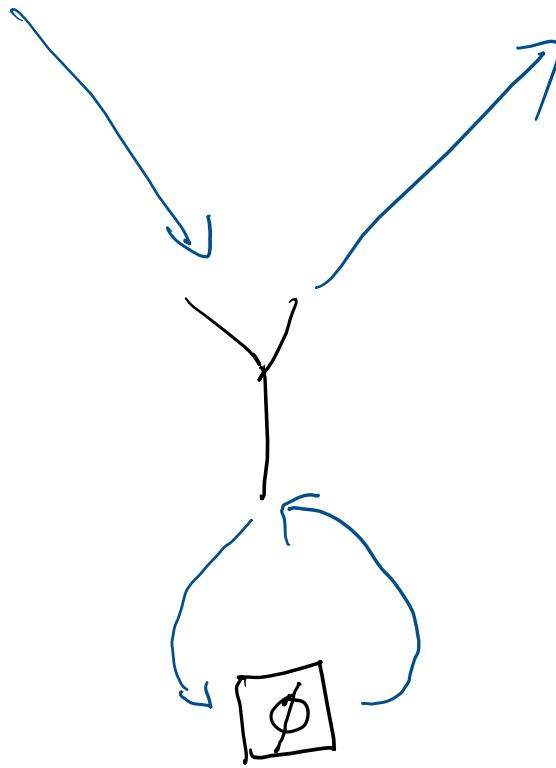


# How Rfocus works



When adjacent columns are activated, the holes cease to matter

# 2-state vs full phase shifter



Phase shifter

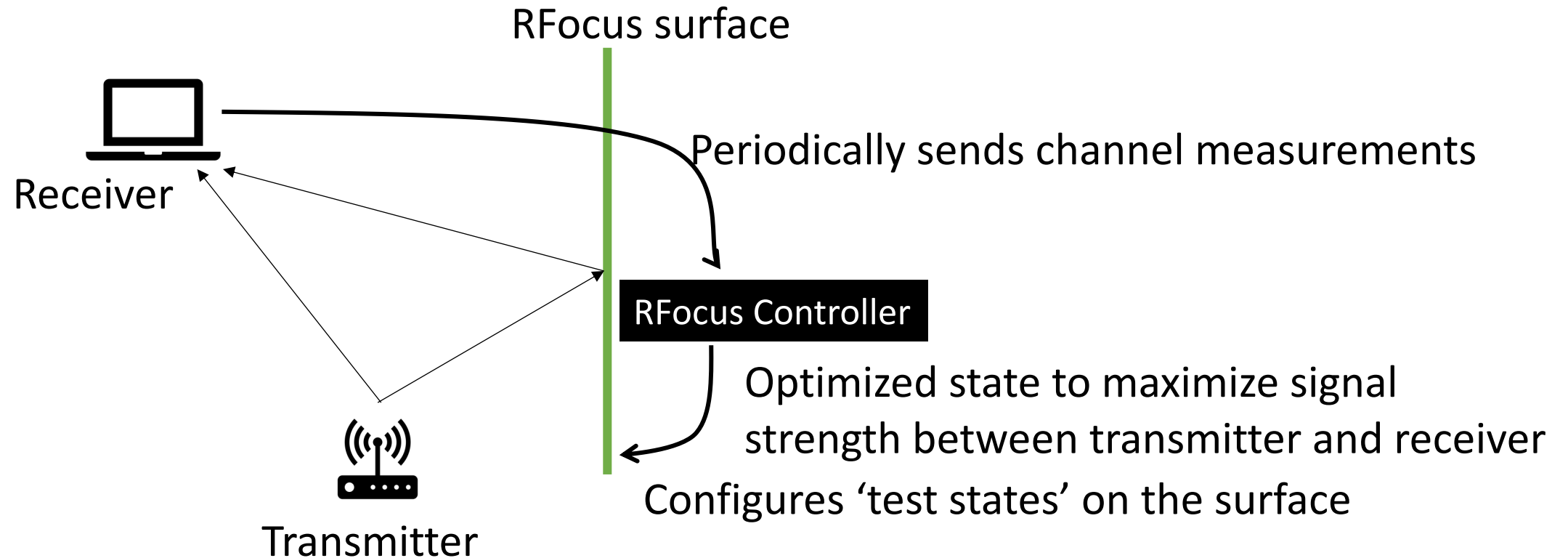
Compare to a full phase shifter, RFocus gets  $\frac{1}{\pi^2}$  of the signal strength improvement

# Goal: Increase Signal Strength

Cellular Networks | WiFi | IoT Sensor Networks

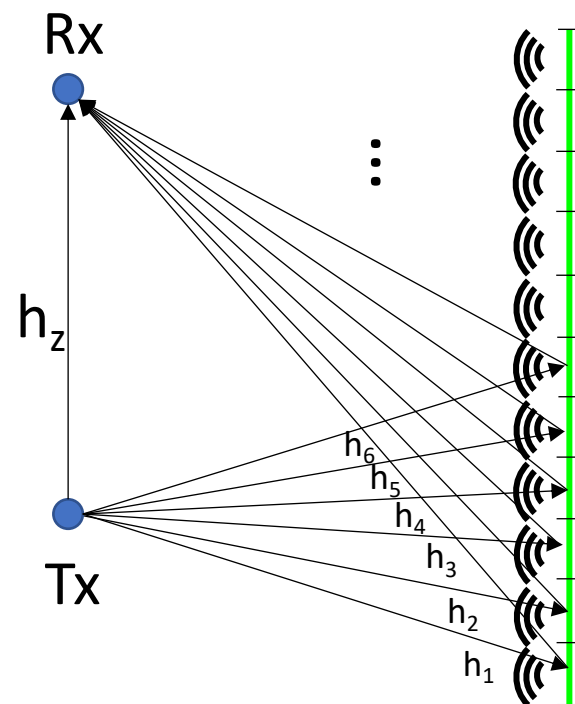
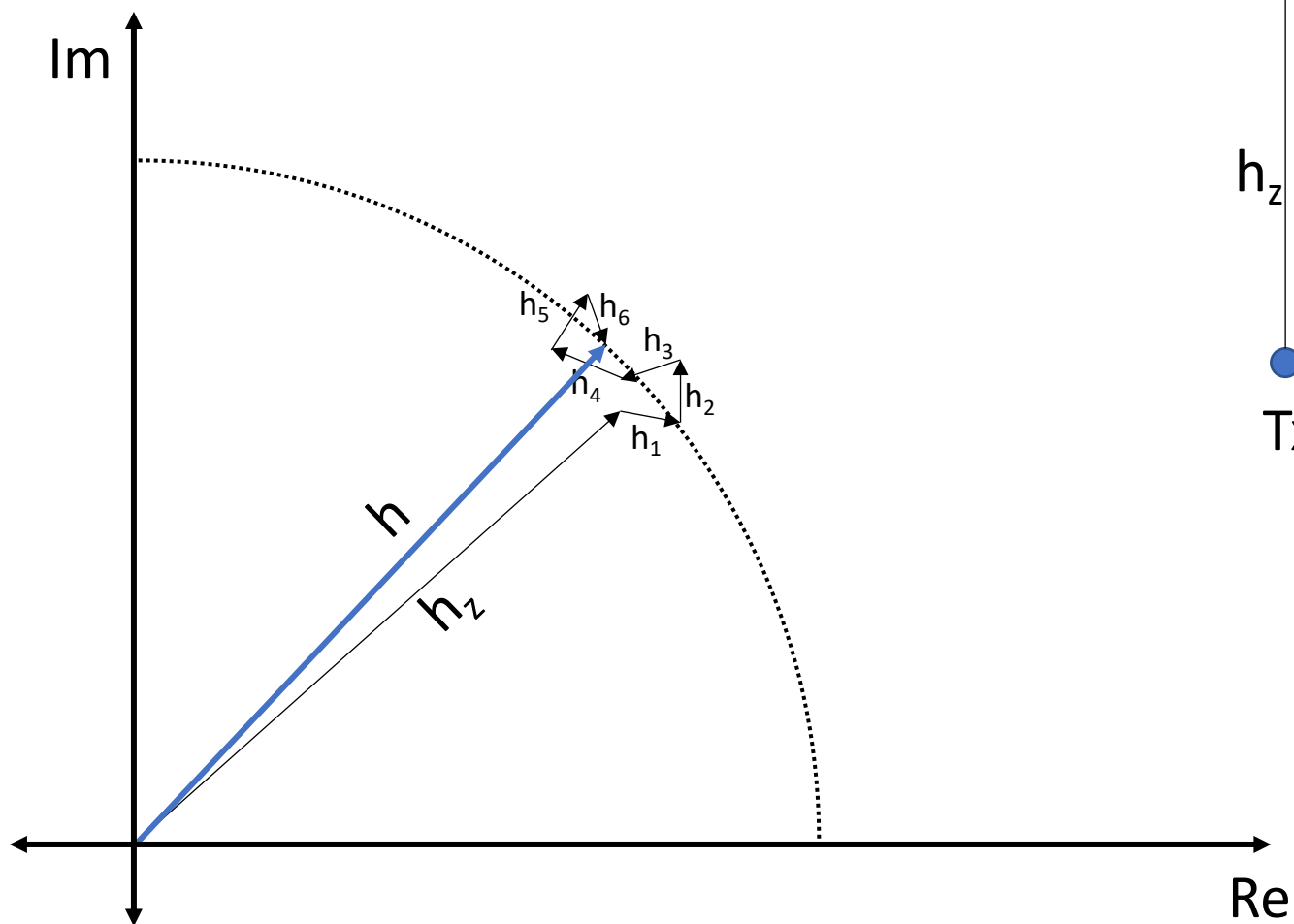


# System Architecture

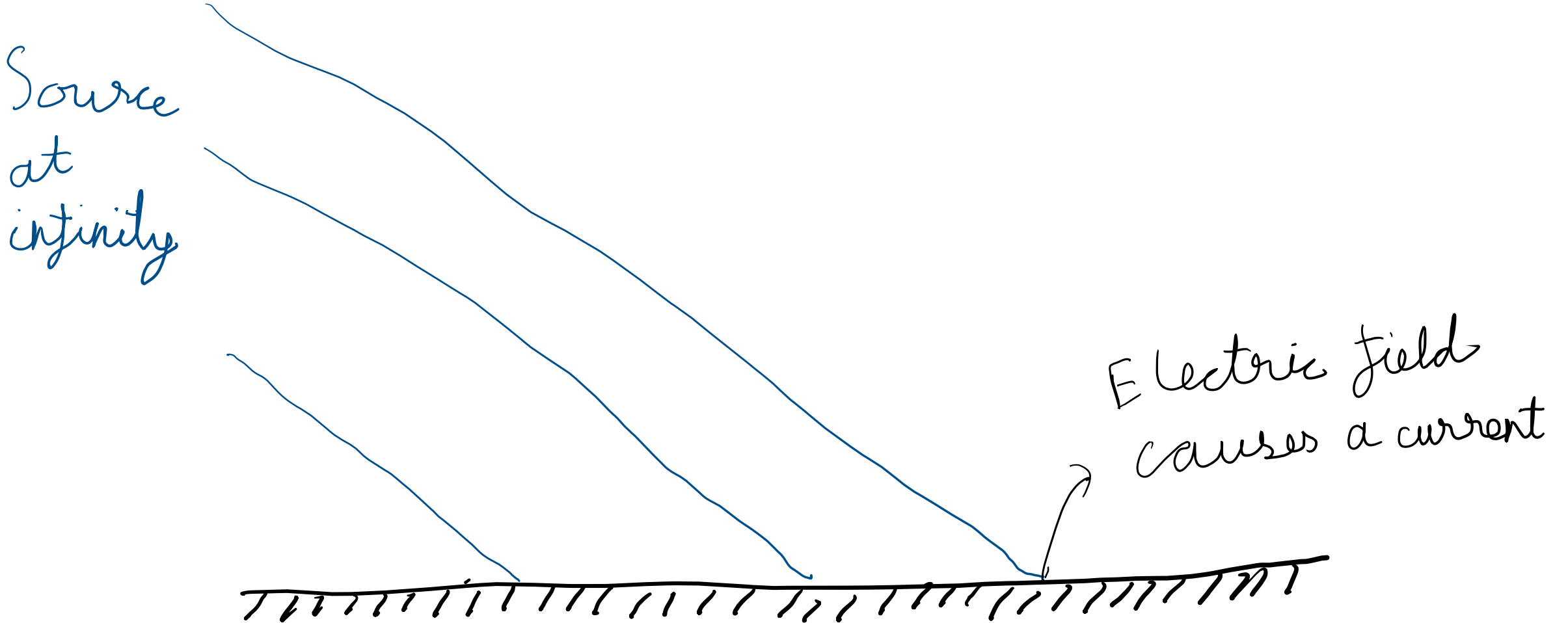




# Reflection from a wall

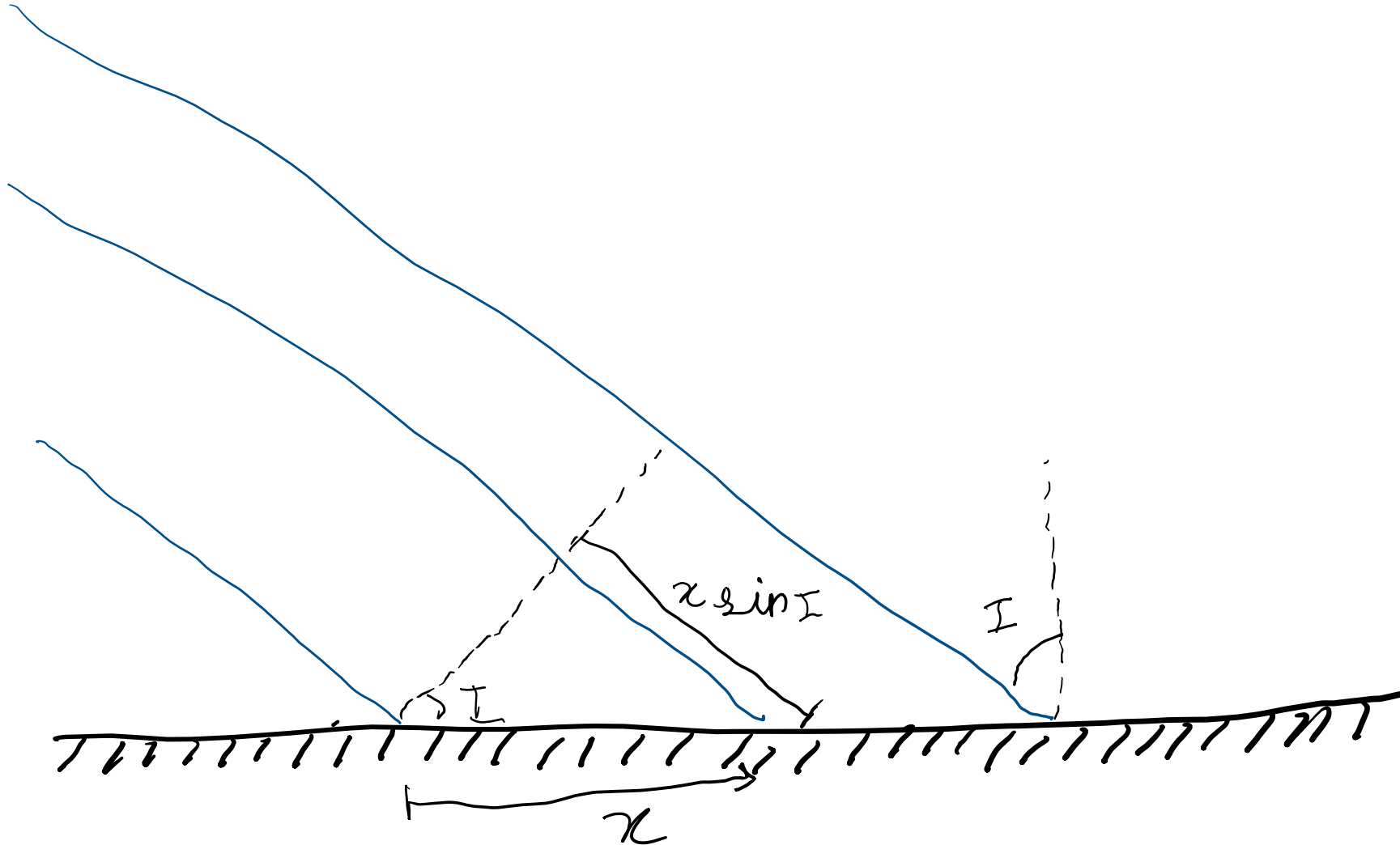


# Reflection from a flat surface (mirror)

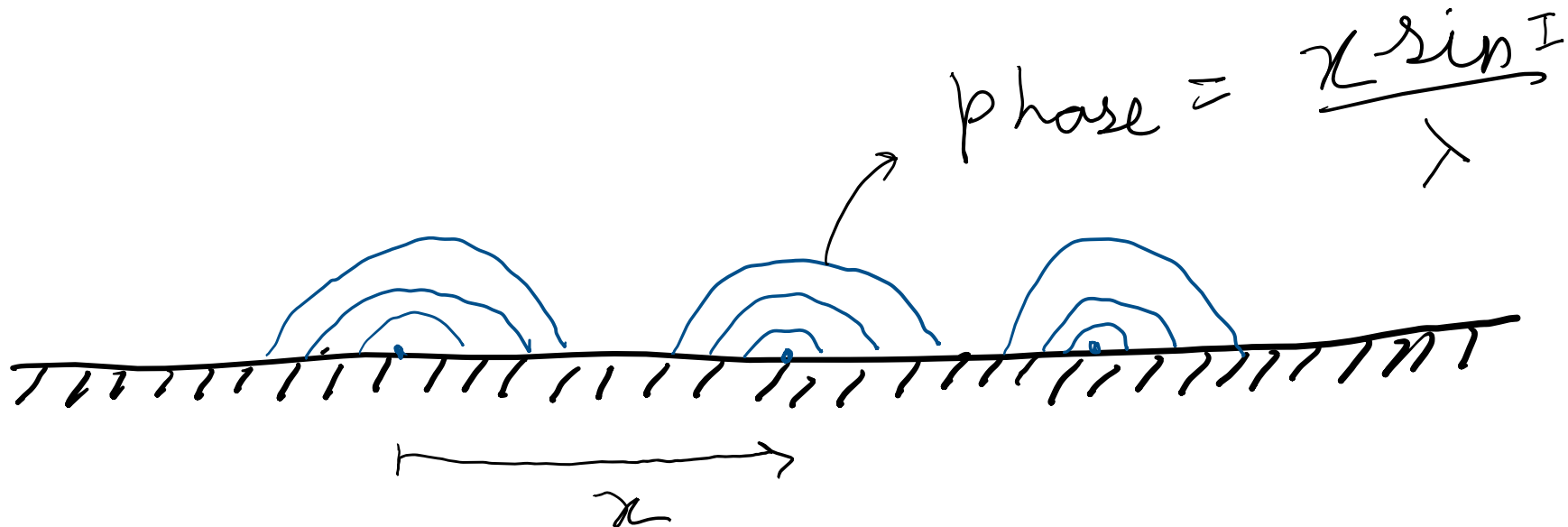


# Reflection from a flat surface (mirror)

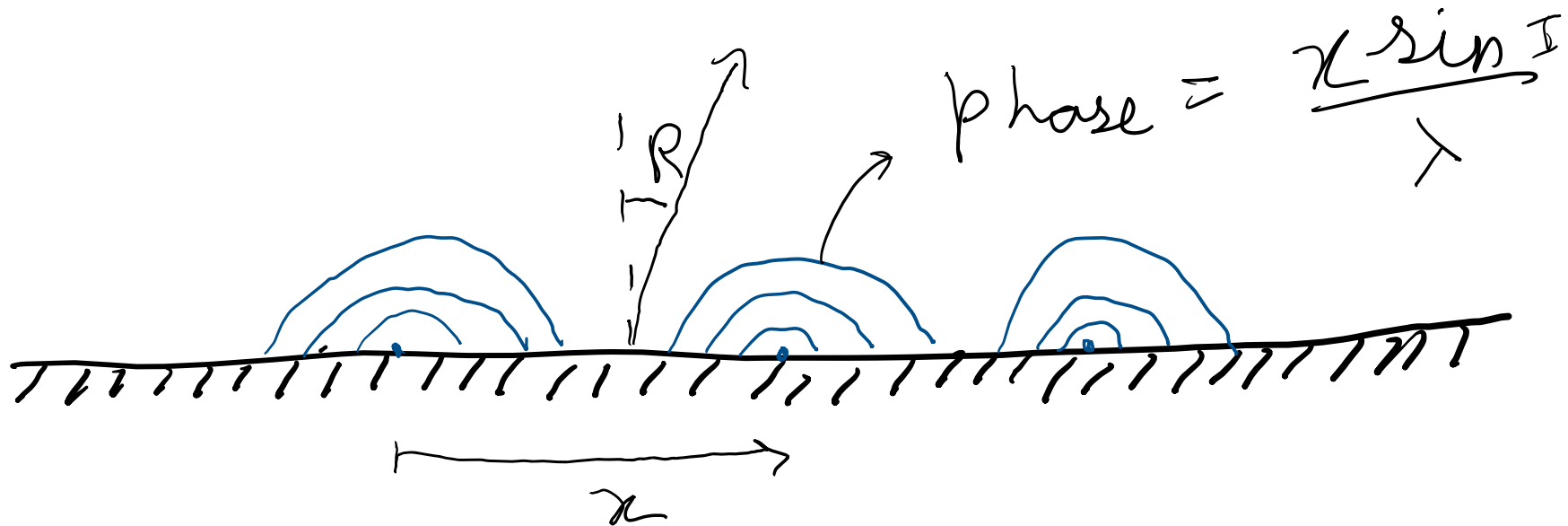
Source  
at  
infinity



# Reflection from a flat surface (mirror)



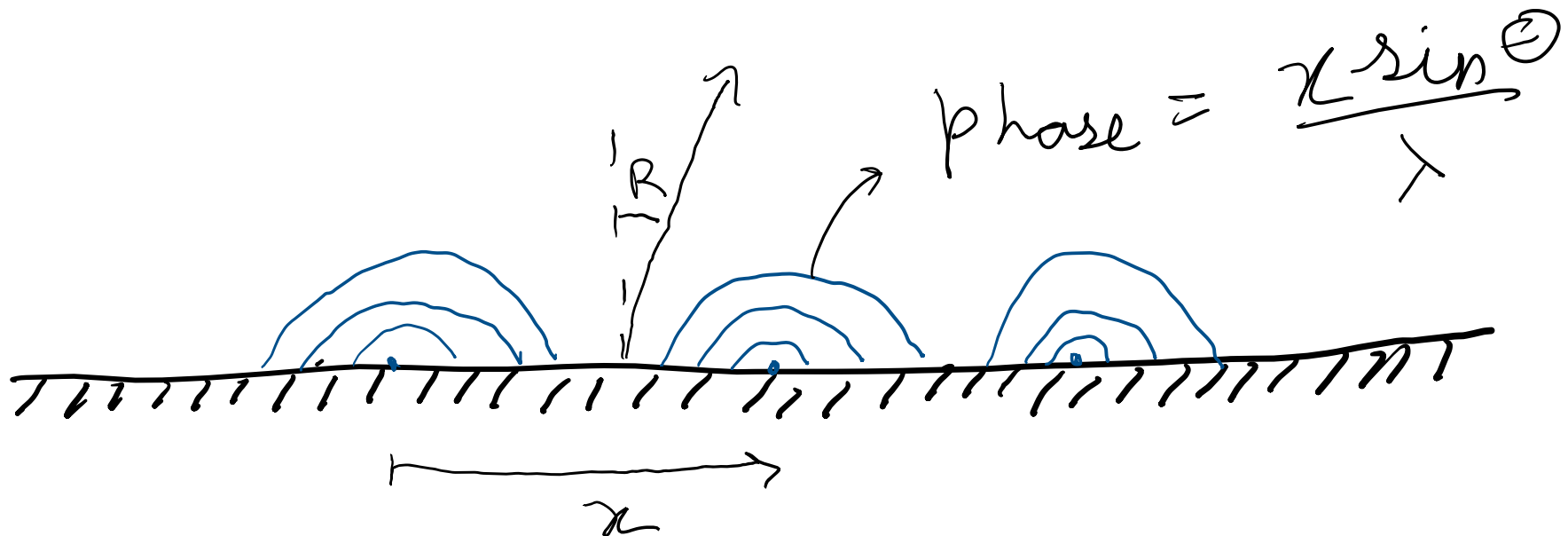
# Reflection from a flat surface (mirror)



# Reflection from a flat surface (mirror)

$$R(\phi) = \int_{-\infty}^{\infty} e^{-jx \sin R / \lambda} e^{jx \sin I / \lambda} \cdot dx$$

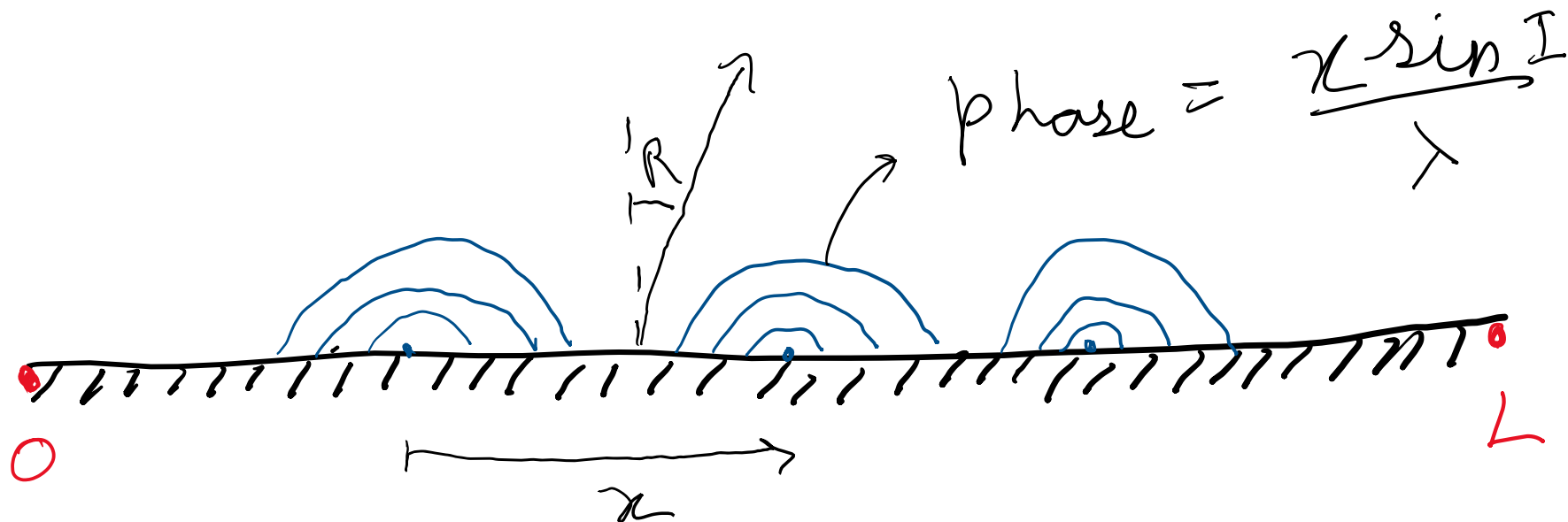
$$= 0 \text{ unless } I = R$$



# Reflection from a flat surface (mirror)

$$R(\phi) = \int_0^L e^{-jx \sin R/\lambda} e^{jx \sin I/\lambda} \cdot dx$$

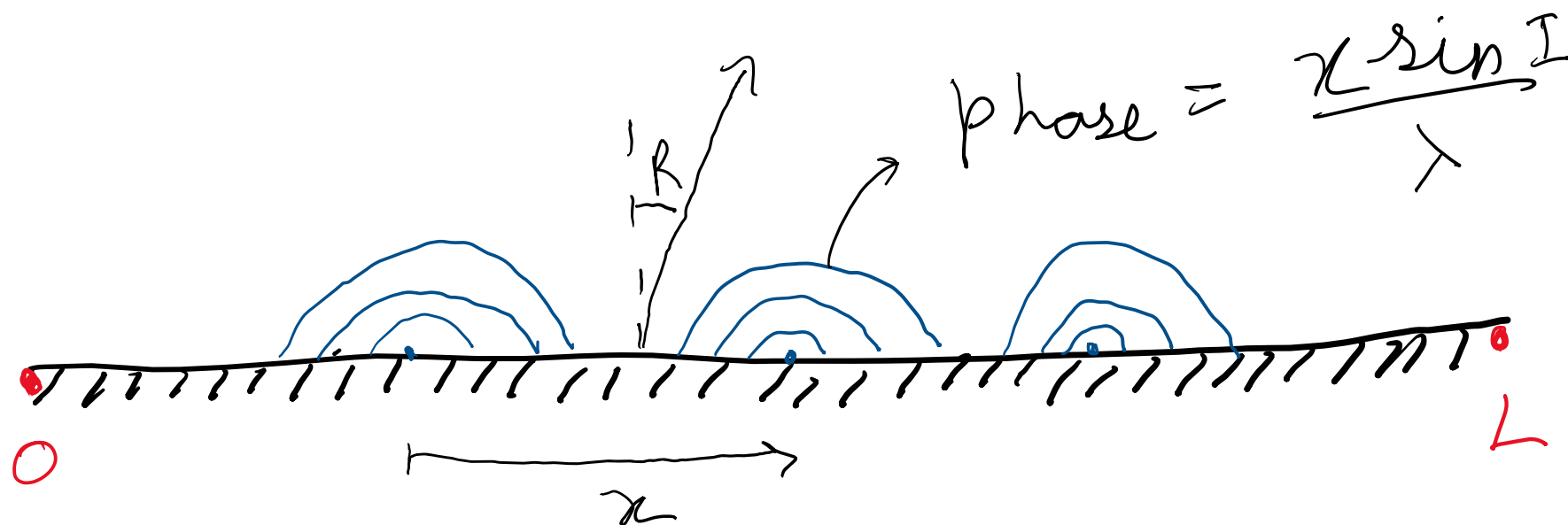
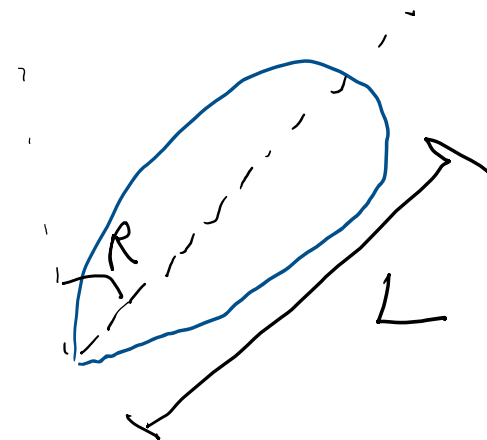
$$\approx L - O((I - R)^2)$$



# Reflection from a flat surface (mirror)

$$R(\phi) = \int_0^L e^{-jx \sin R/\lambda} e^{jx \sin I/\lambda} dx$$

$$\approx L - O((I - R)^2)$$

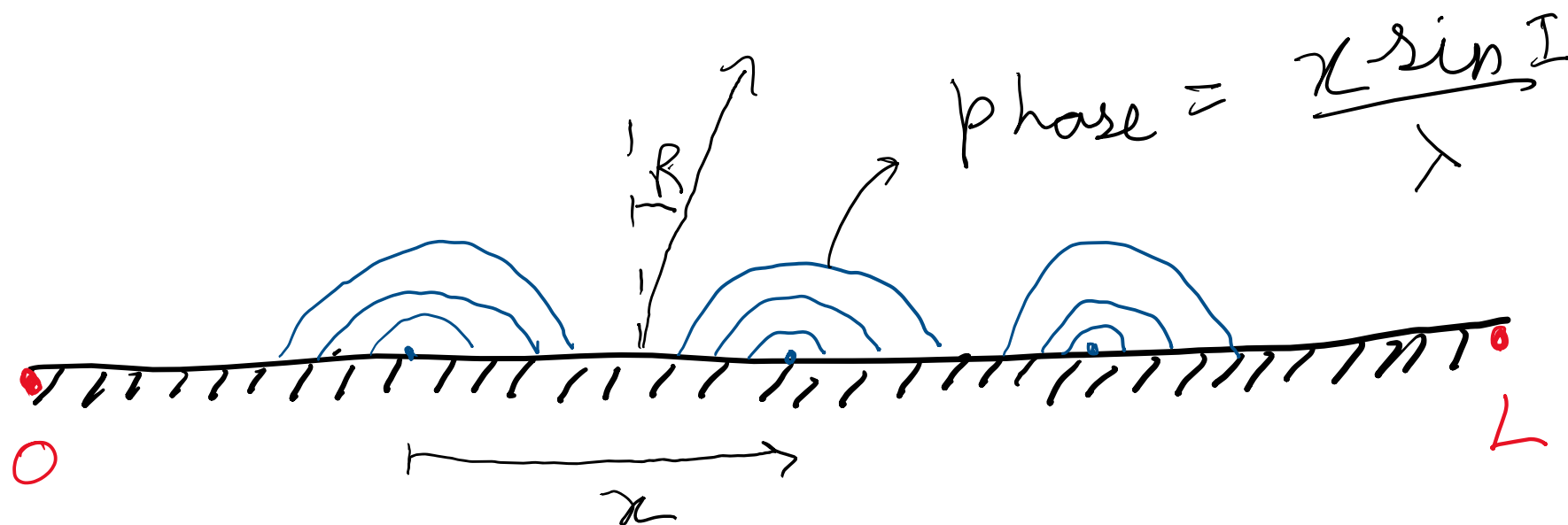
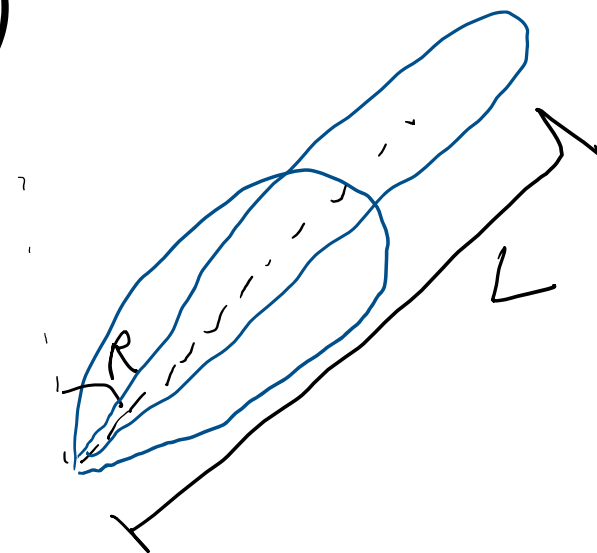




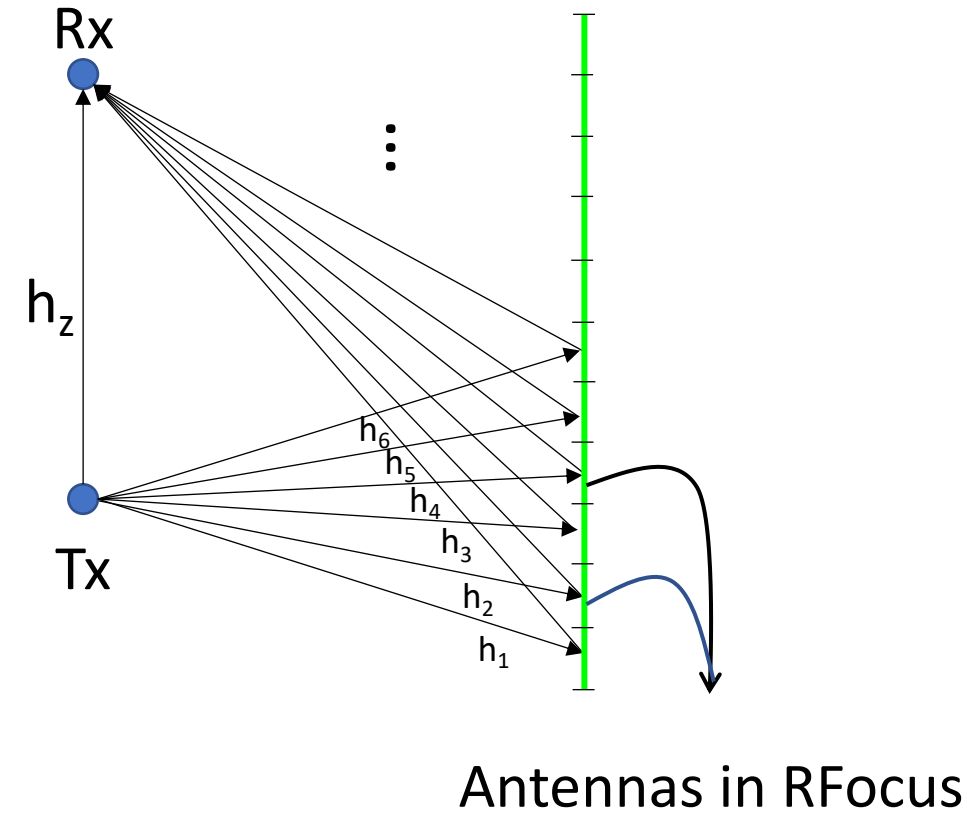
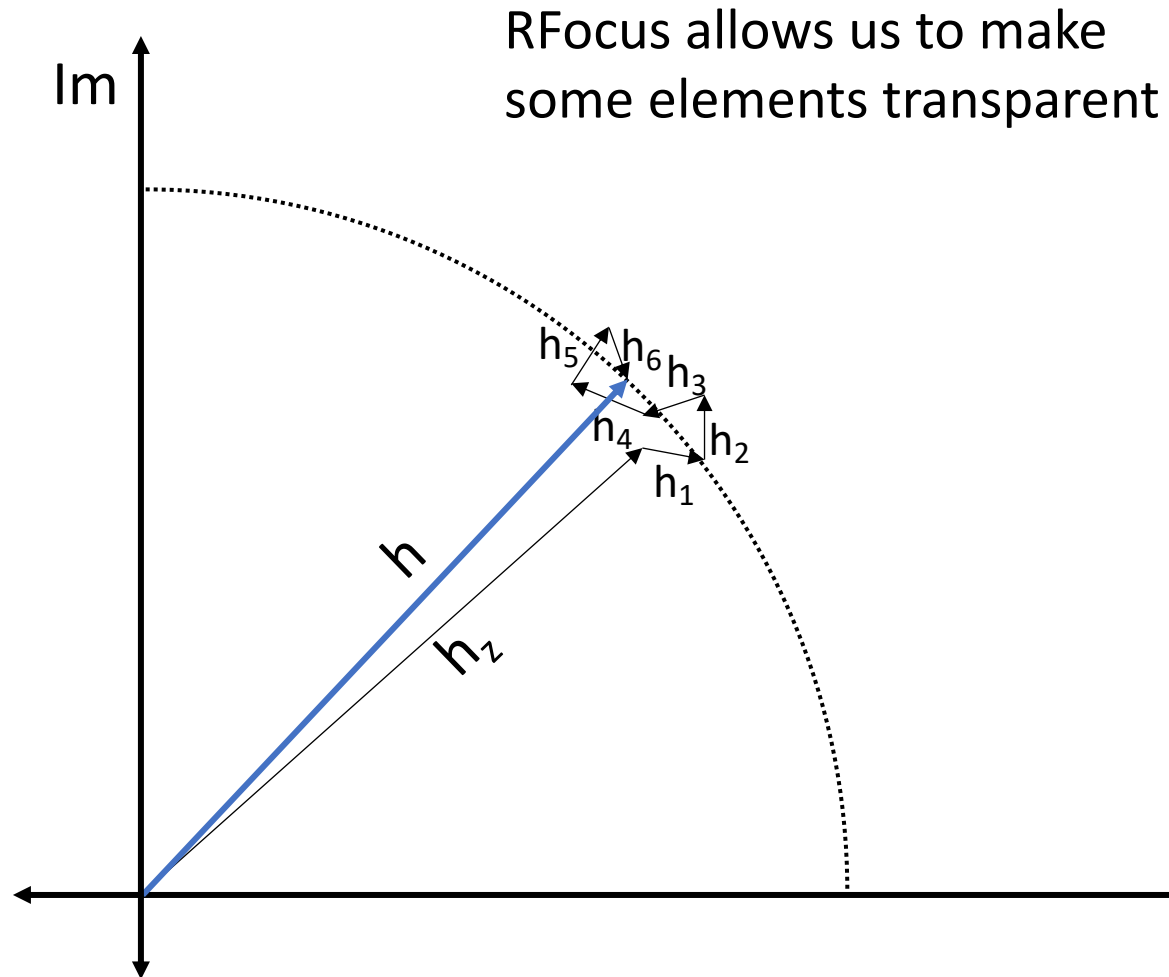
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$$R(\phi) = \int_0^L e^{-jx \sin R/\lambda} e^{jx \sin I/\lambda} dx$$

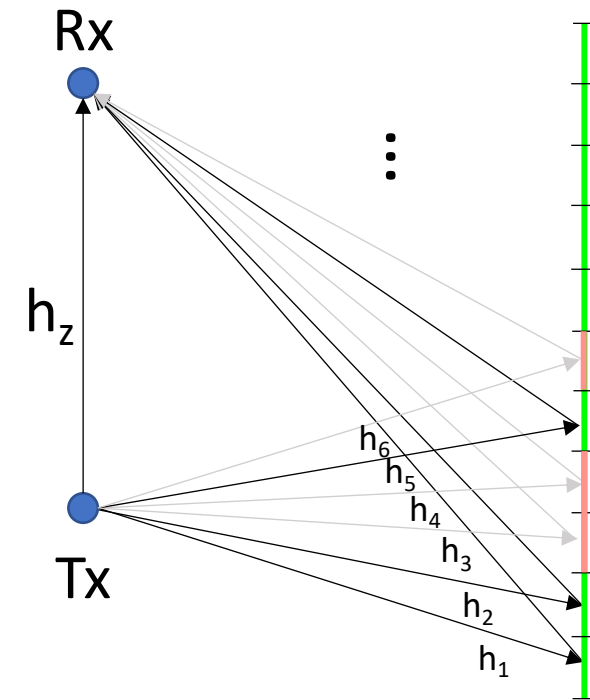
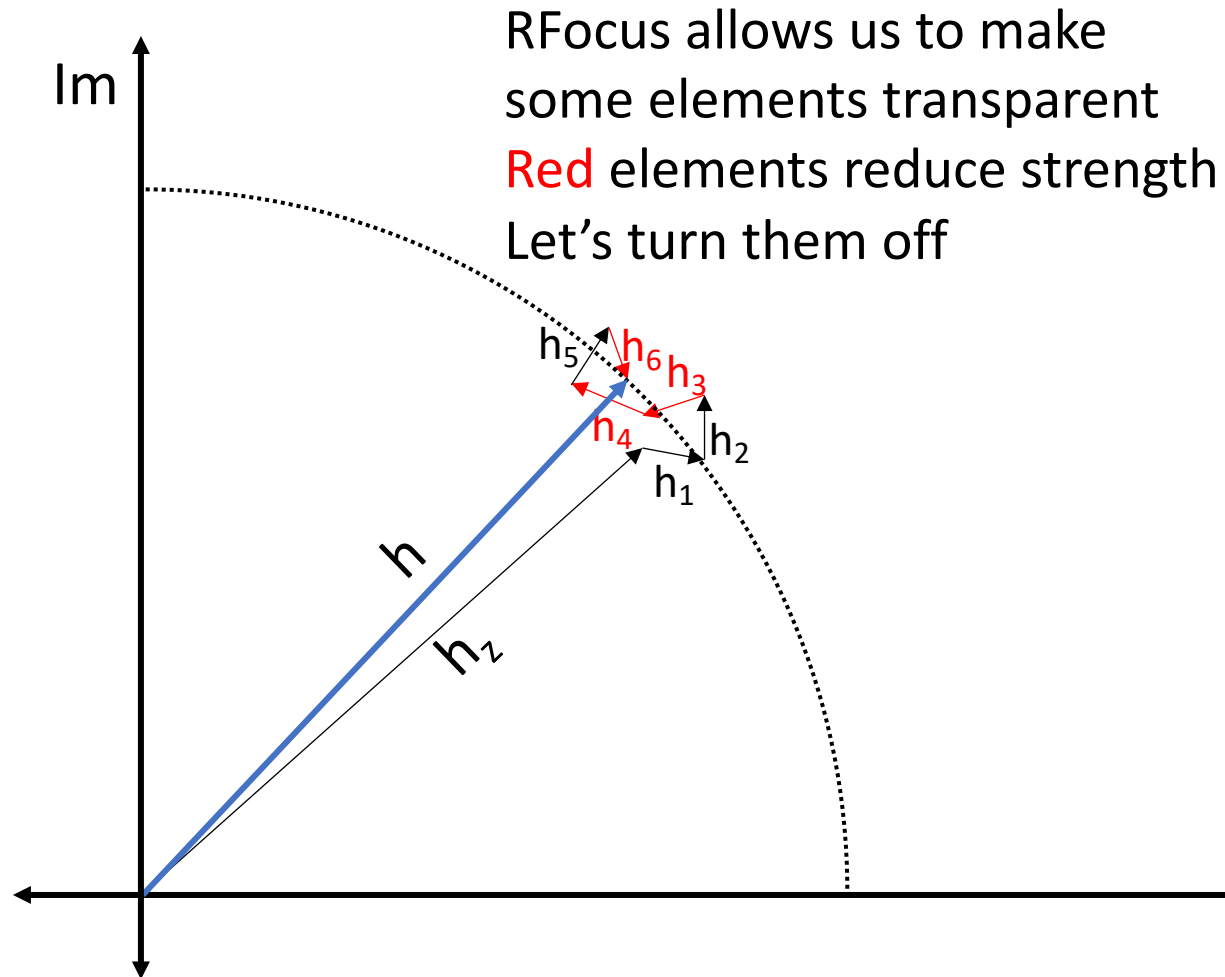
$$\approx L - O((I - R)^2)$$



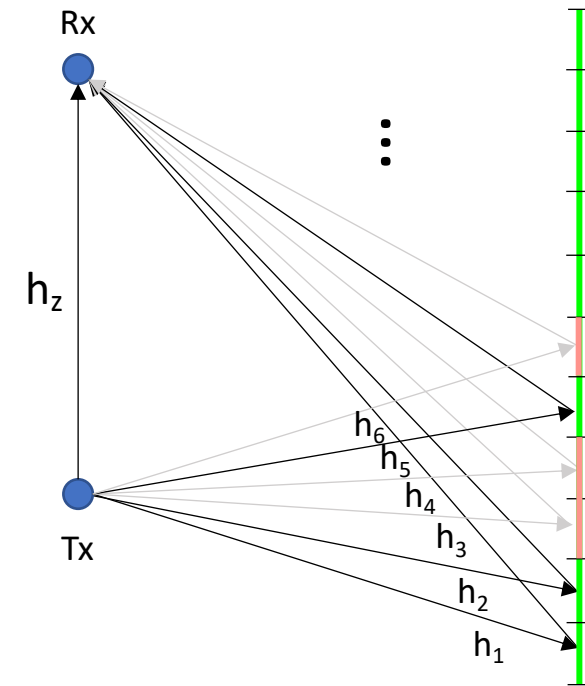
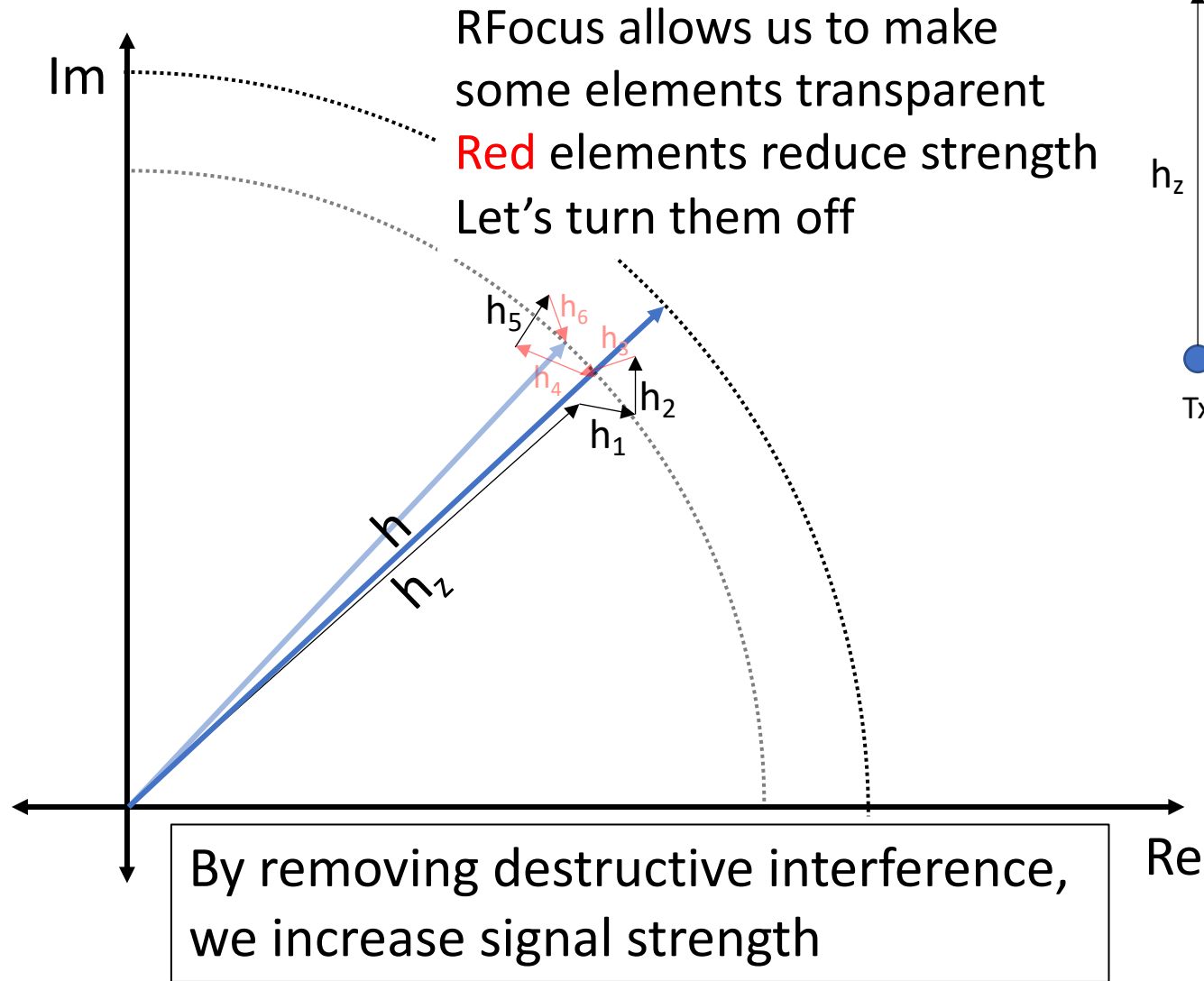
# Improving the Reflection



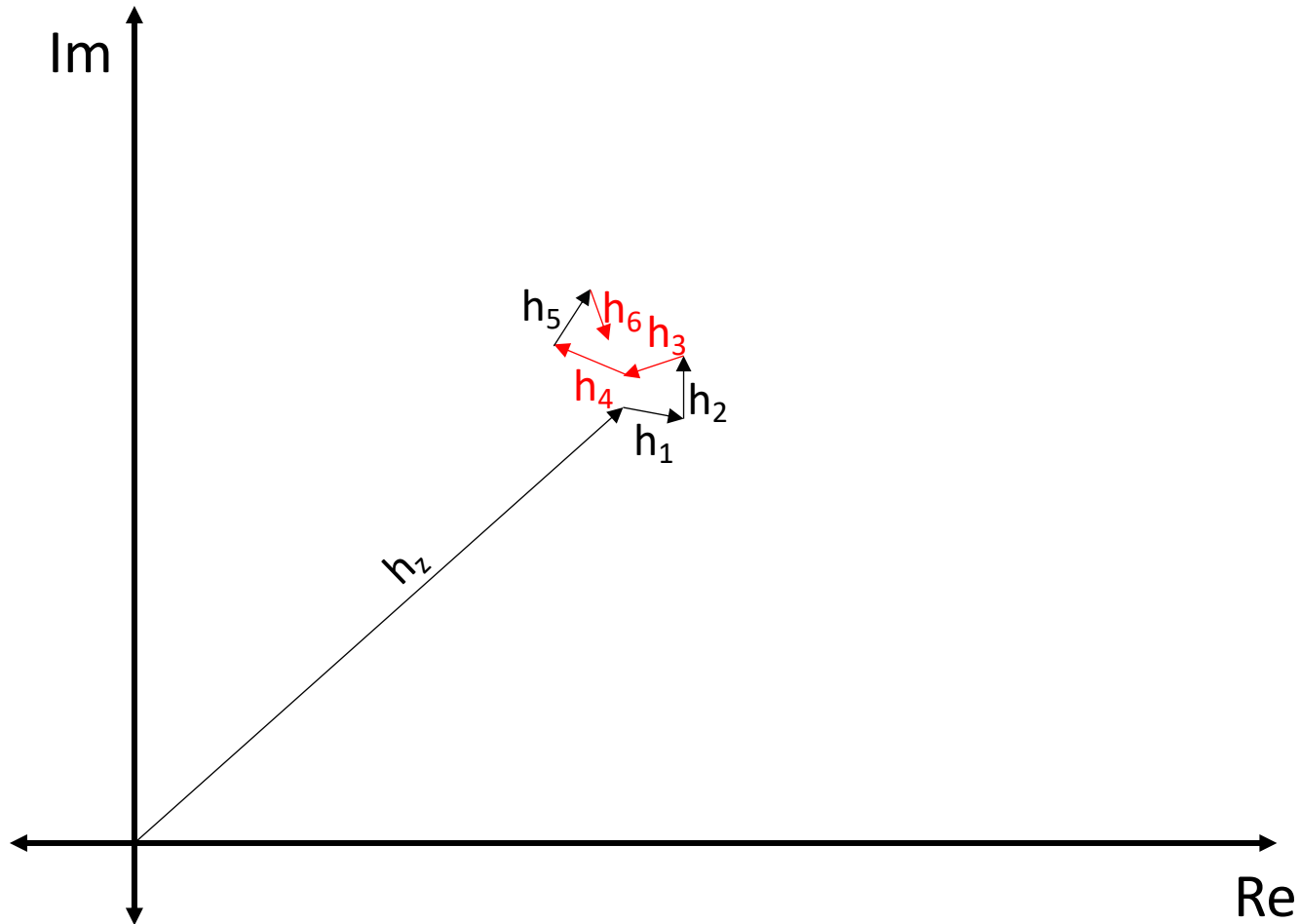
# Improving the Reflection



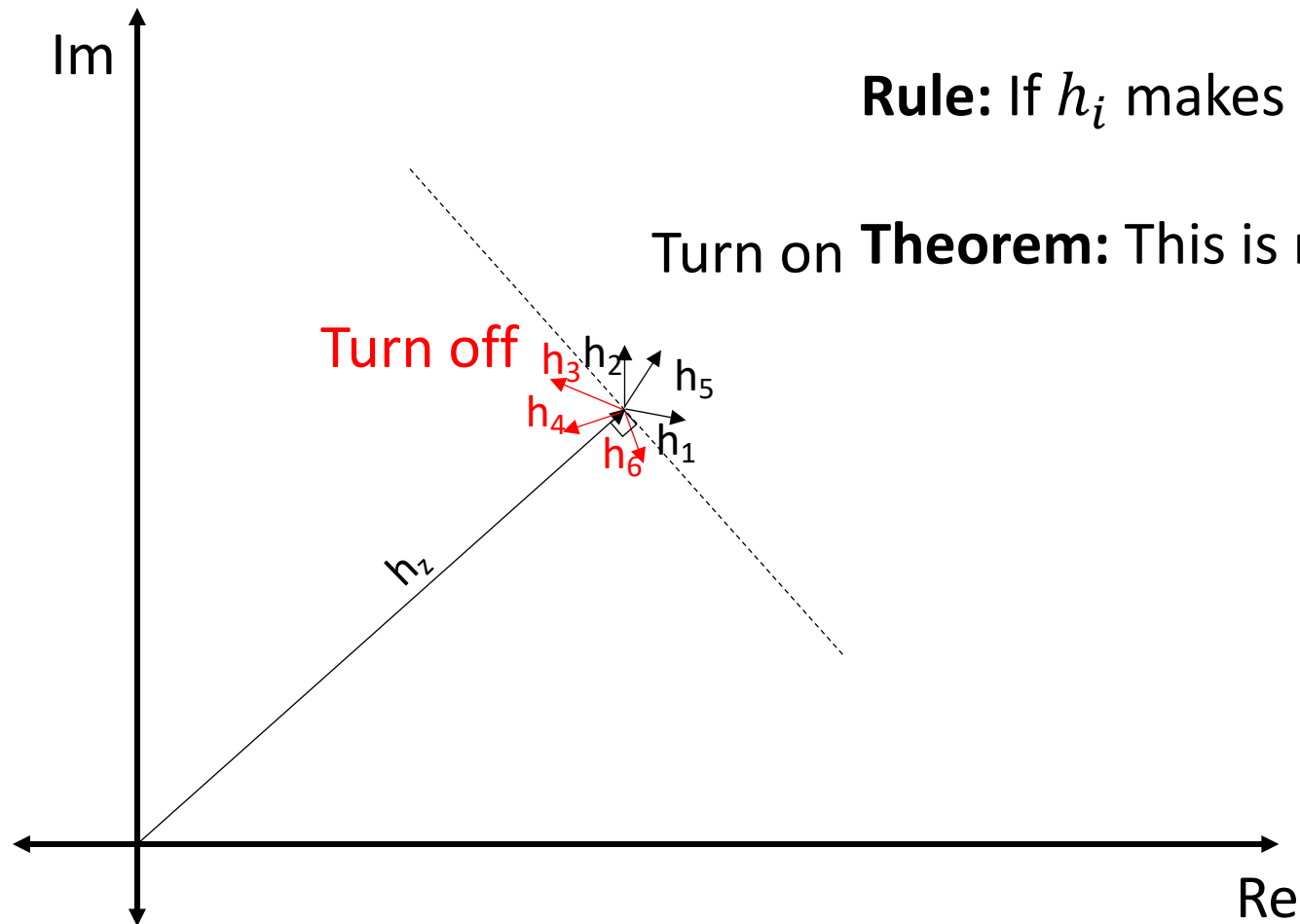
# Improving the Reflection



Which antennas should we turn off?



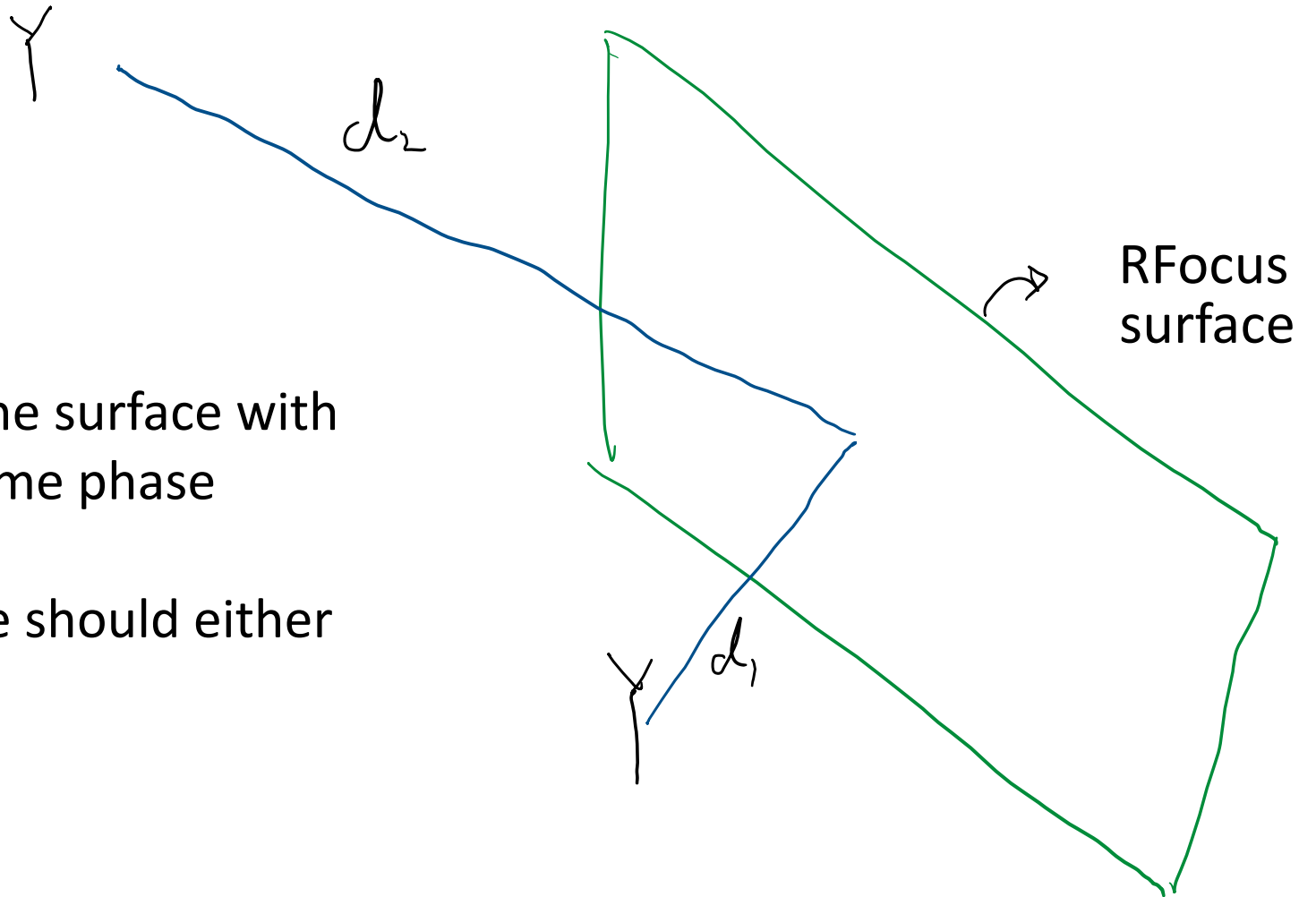
# Which antennas should we turn off?



**Rule:** If  $h_i$  makes an acute angle with  $h_z$ , turn it on

Turn on **Theorem:** This is near optimal

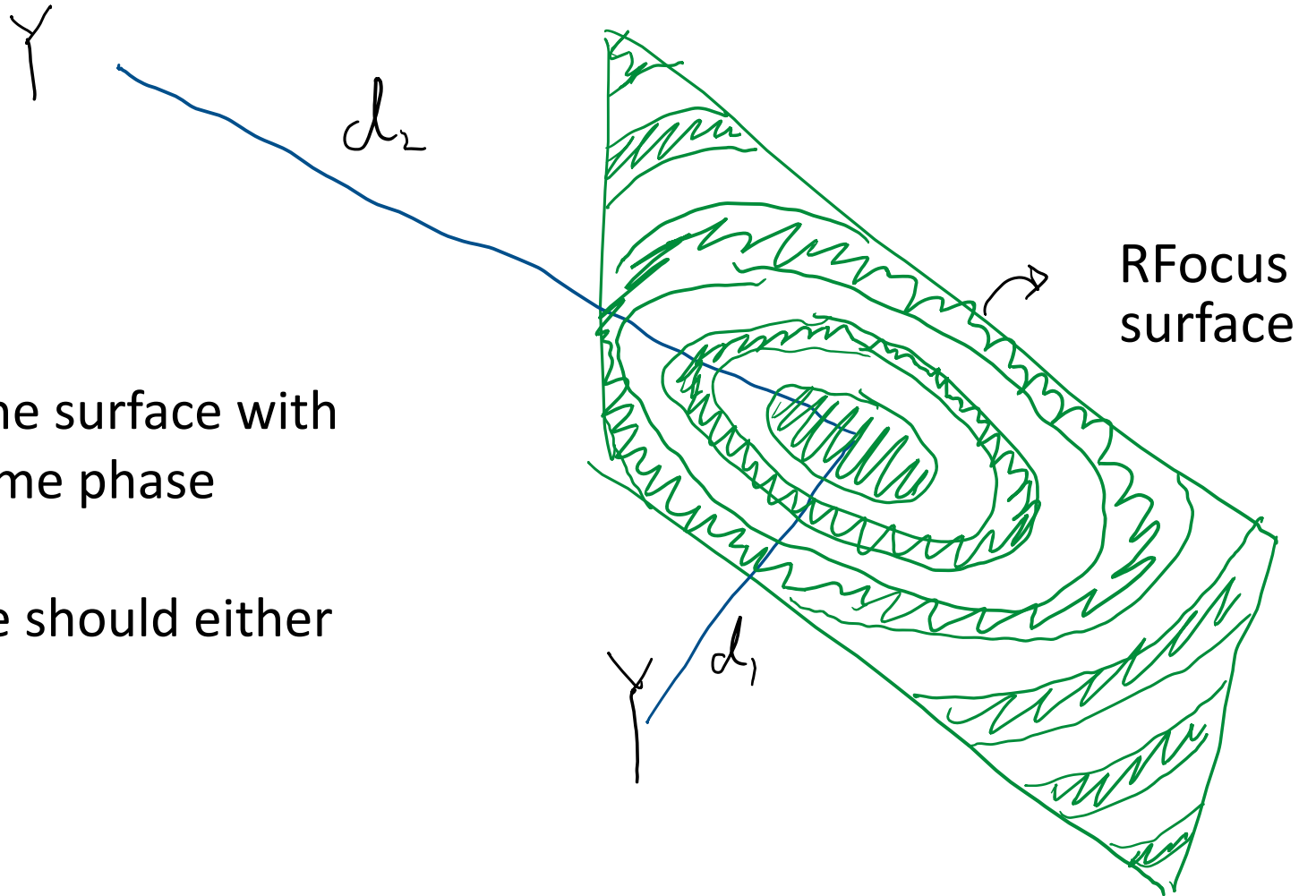
# RFocus in empty space



The paths from all points on the surface with the same  $d_1 + d_2$  have the same phase

All points with the same phase should either be on or off

# RFocus in empty space

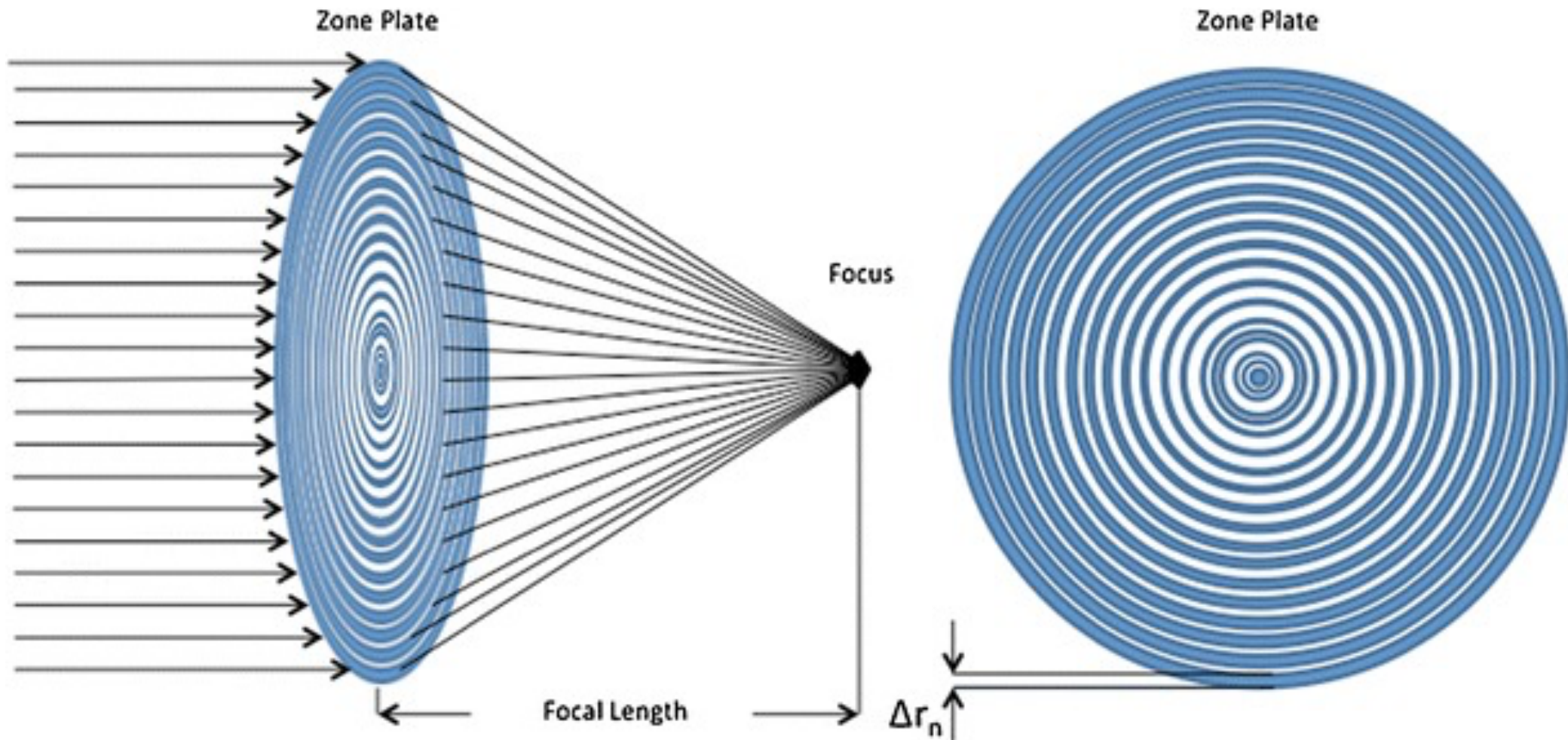


The paths from all points on the surface with the same  $d_1 + d_2$  have the same phase

All points with the same phase should either be on or off

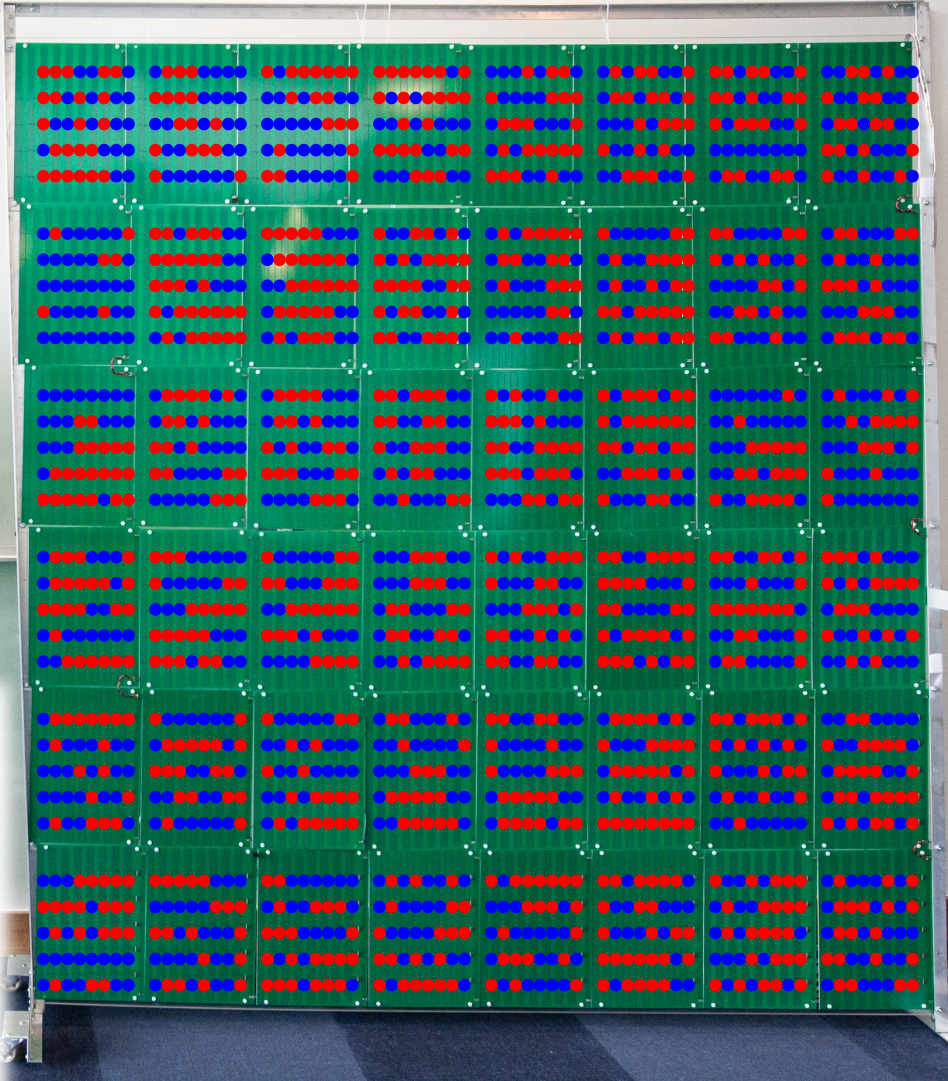


# RFocus in empty space



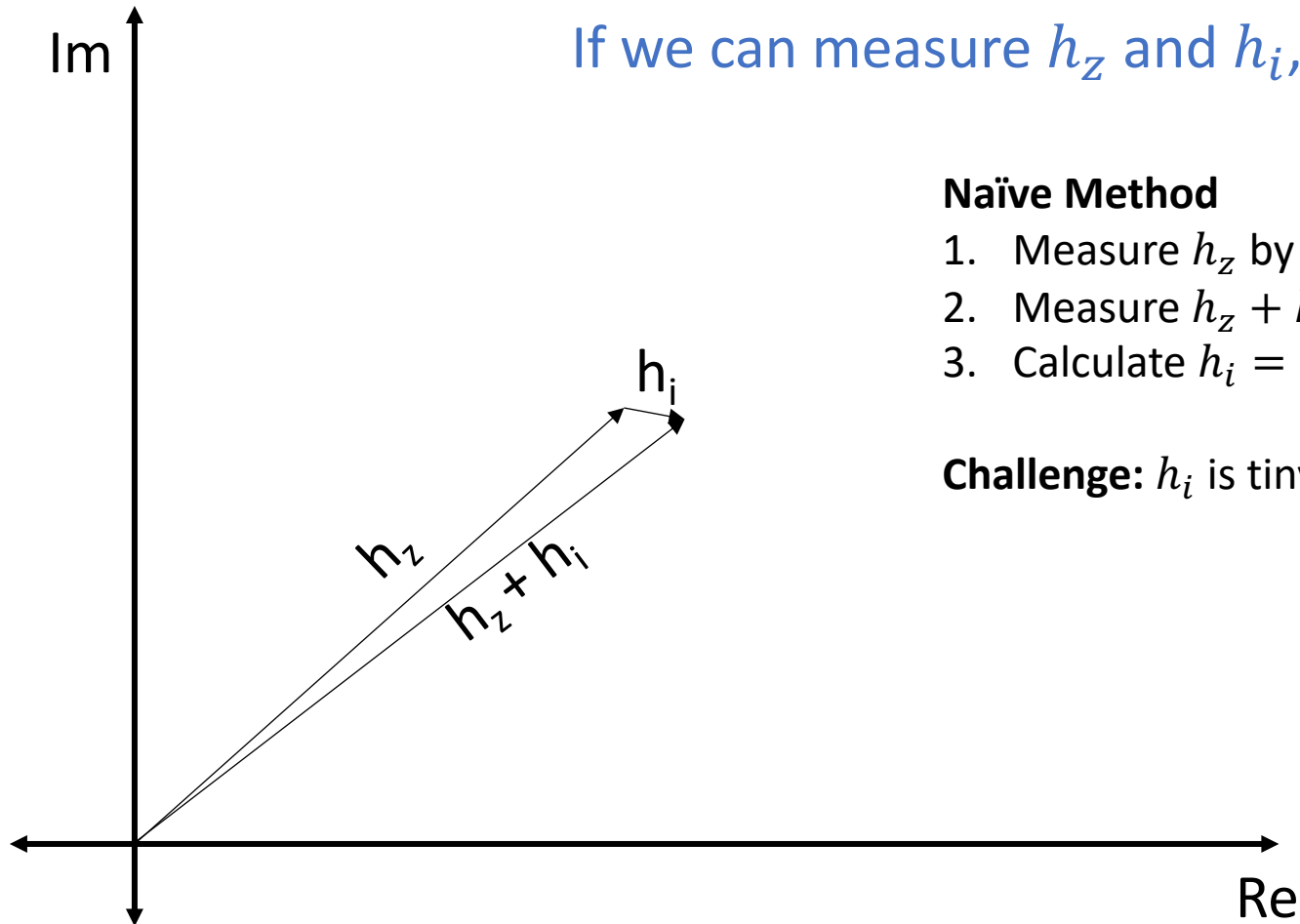




- 
- On - Reflective
- Off - Transparent

In practice, we don't  
operate in empty space.  
The state shown is a real  
example

# Strawman 1: Prior Work



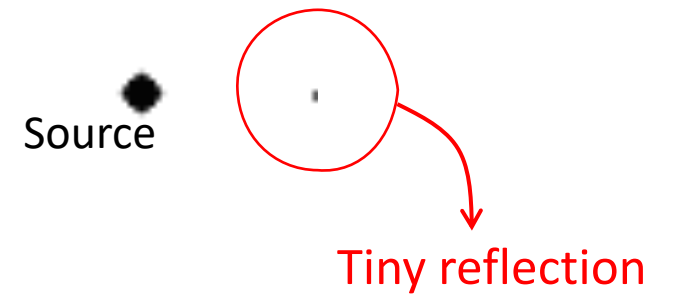
If we can measure  $h_z$  and  $h_i$ , we can find the optimum state

## Naïve Method

1. Measure  $h_z$  by turning all antennas off
2. Measure  $h_z + h_i$  by turning just the  $i^{th}$  antenna on
3. Calculate  $h_i = (h_z + h_i) - h_z$

**Challenge:**  $h_i$  is tiny!

$h_i$  are tiny and hard to measure





# Prior Work

- Large body of theoretical work and some experimental work
- They do not address the problem of measuring  $h_i$
- Hence cannot scale to larger distances and larger number of antennas
- RFocus is the first large-scale prototype to demonstrate the feasibility of such a system
- Increasing interest in the CS community

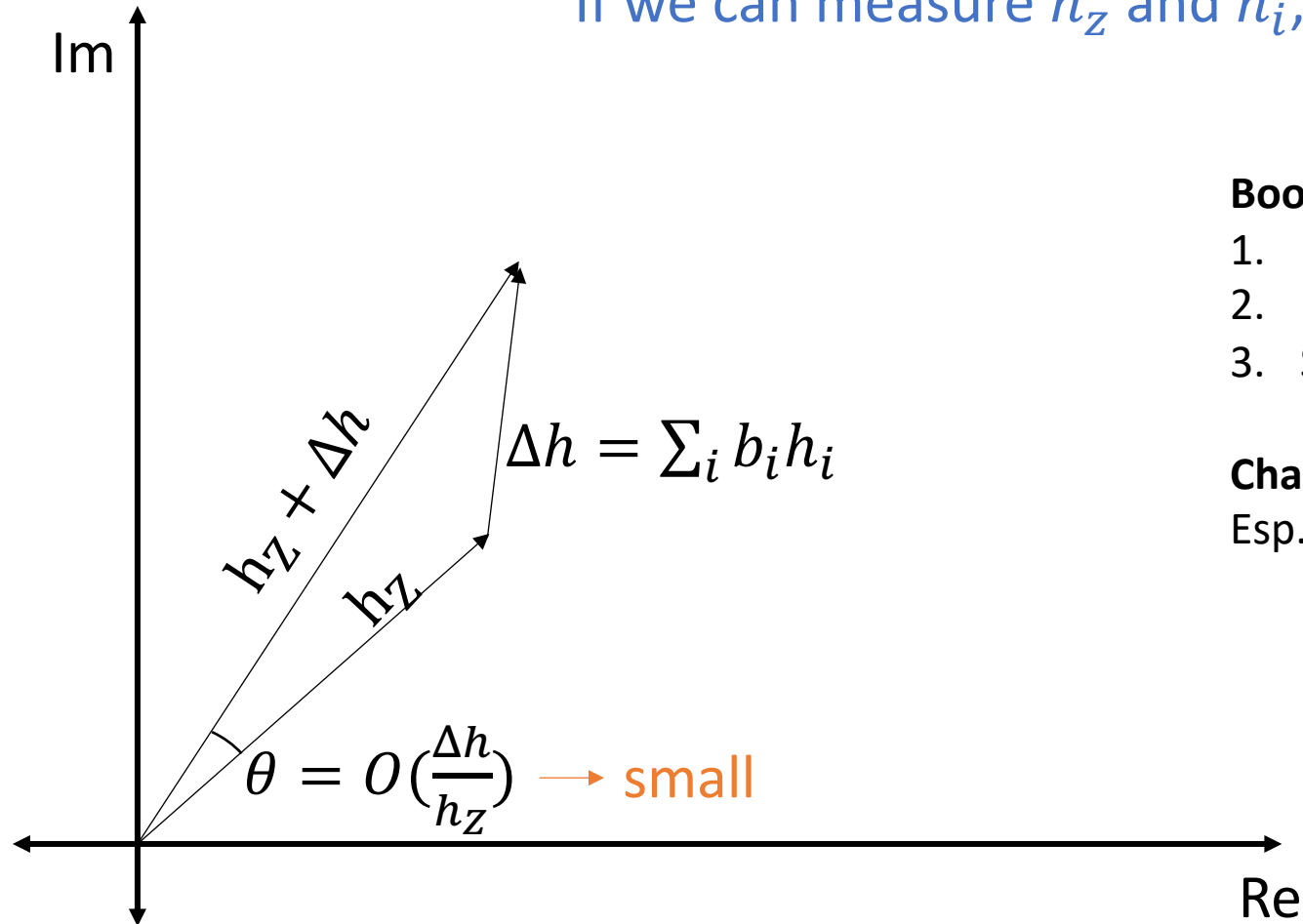
- “Optimally diverse communication channels in disordered environments with tuned randomness”, P. del Hougne, M. Fink, and G. Lerosey
- “Shaping complex microwave fields in reverberating media with binary tunable metasurfaces”, N. Kaina, M. Dupré, G. Lerosey, and M. Fink
- “Increasing indoor spectrum sharing capacity using smart reflect-array”, X. Tan, Z. Sun, J. M. Jornet, and D. Pados

# Key Ideas: to measure tiny $h_i$

- **Boosting:** Instead of measuring the effect of one antenna, turn on a random subset of antennas and measure the effect
  - This effect is  $O(N)$  times bigger (here,  $N = 3200\times$ )
- **Signal Strength:** The effect on phase is still small, esp. due to clock jitter and CFO drift. Rely only on signal strength (RSSI) instead.

# Strawman 2: Boosting

If we can measure  $h_z$  and  $h_i$ , we can find the optimum state



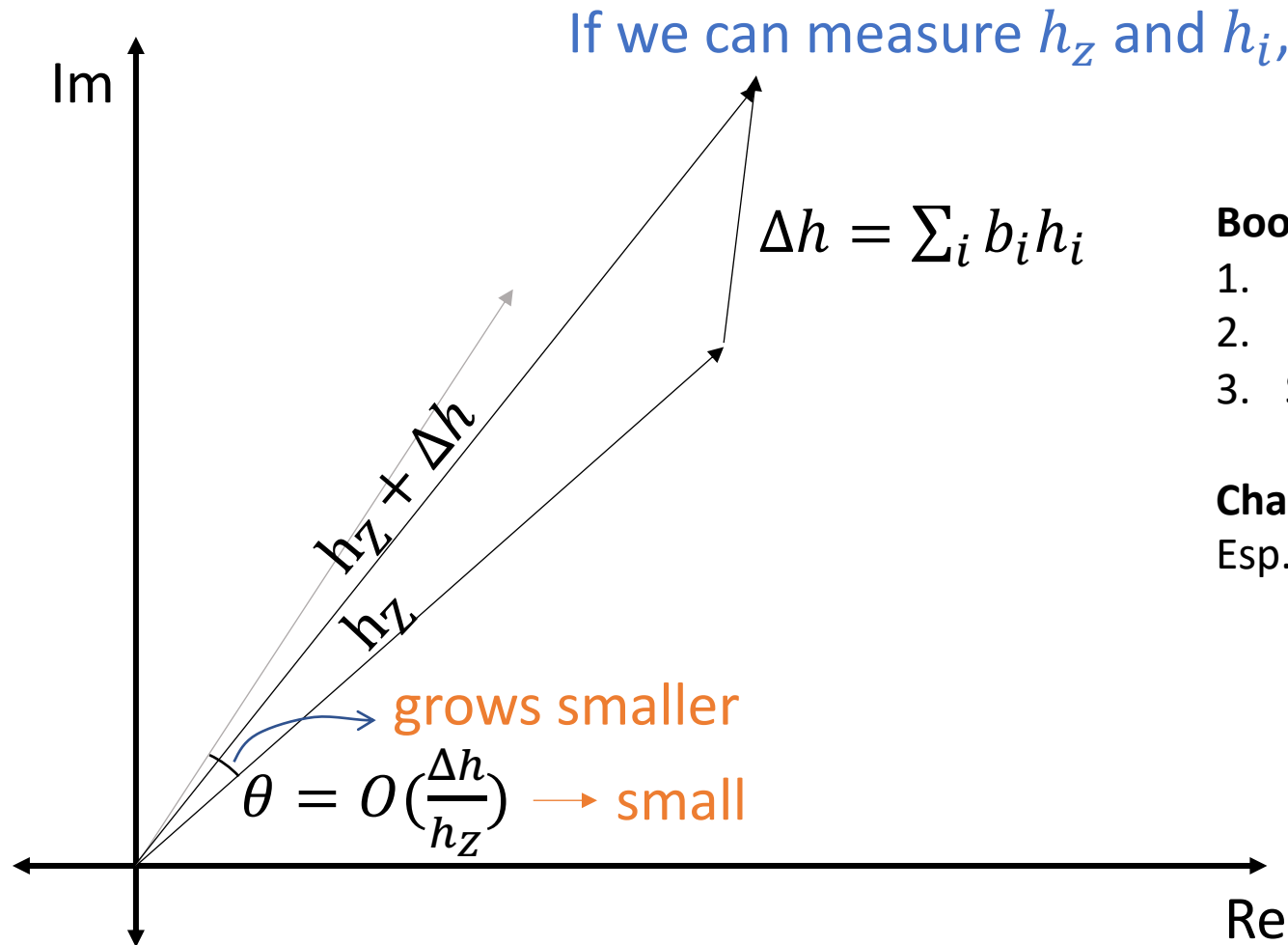
## Boost the Signal:

1. Measure  $h_z$
2. Measure  $h_z + \sum_i b_i h_i$  for many random  $b_i$
3. Solve linear equations for  $h_i$

**Challenge:**  $\sum_i b_i h_i$  is still small!  
Esp. phase change is small



# Strawman 2: Boosting



## Boost the Signal:

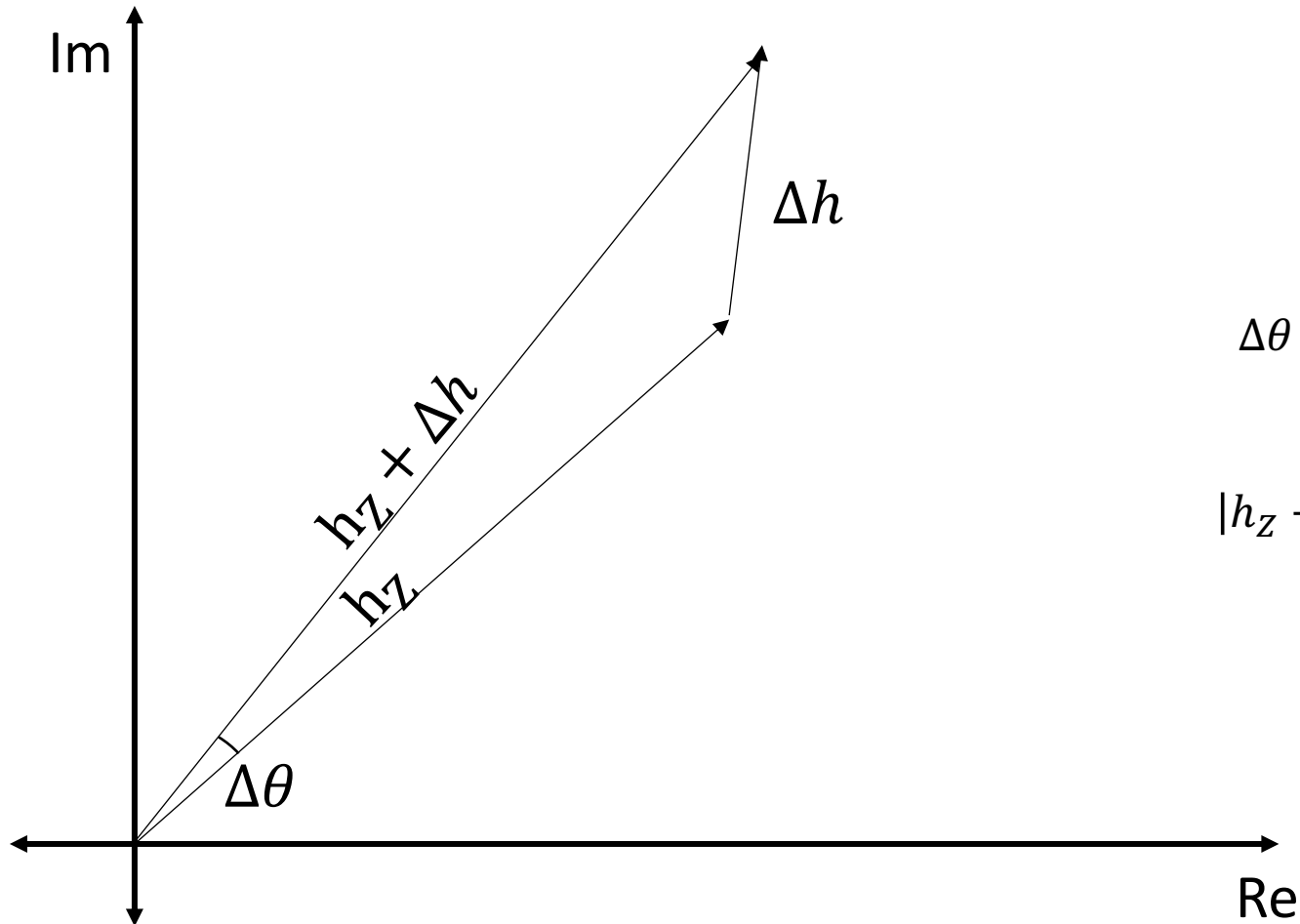
1. Measure  $h_z$
2. Measure  $h_z + \sum_i b_i h_i$  for many random  $b_i$
3. Solve for linear equations  $h_i$

**Challenge:**  $\sum_i b_i h_i$  is still small!  
Esp. phase change is small

# Assumptions

1. The phases of  $h_i$  are uniformly distributed random variables in  $(-\pi, \pi]$
2. For uniformly random  $\mathbf{b}$ ,  $|\sum_i b_i h_i| \ll |h_Z|$  with high probability
3.  $|h_i|$  is bounded above by a constant, even as  $N \rightarrow \infty$

# Measuring signal strength is easier than measuring phase



When  $\Delta h \ll h_z$ ,

$$\Delta\theta = \text{Arg}(h_z + \Delta h) - \text{Arg}(h_z) \approx \Im\left(\frac{\Delta h}{h_z}\right)$$

$$|h_z + \Delta h| - |h_z| \approx |h_z| \Re\left(\frac{\Delta h}{h_z}\right) = \frac{\Re(\Delta h h_z^*)}{|h_z|}$$

# How we take measurements

Antenna ID →

1	2	3	4	signal strength Measurement
0	1	1	0	0.8
1	1	1	0	0.9
0	0	1	1	1
1	0	0	0	1.1
1	1	0	1	1.1

# Strawman 3: Take the max of all rows

Antenna ID →

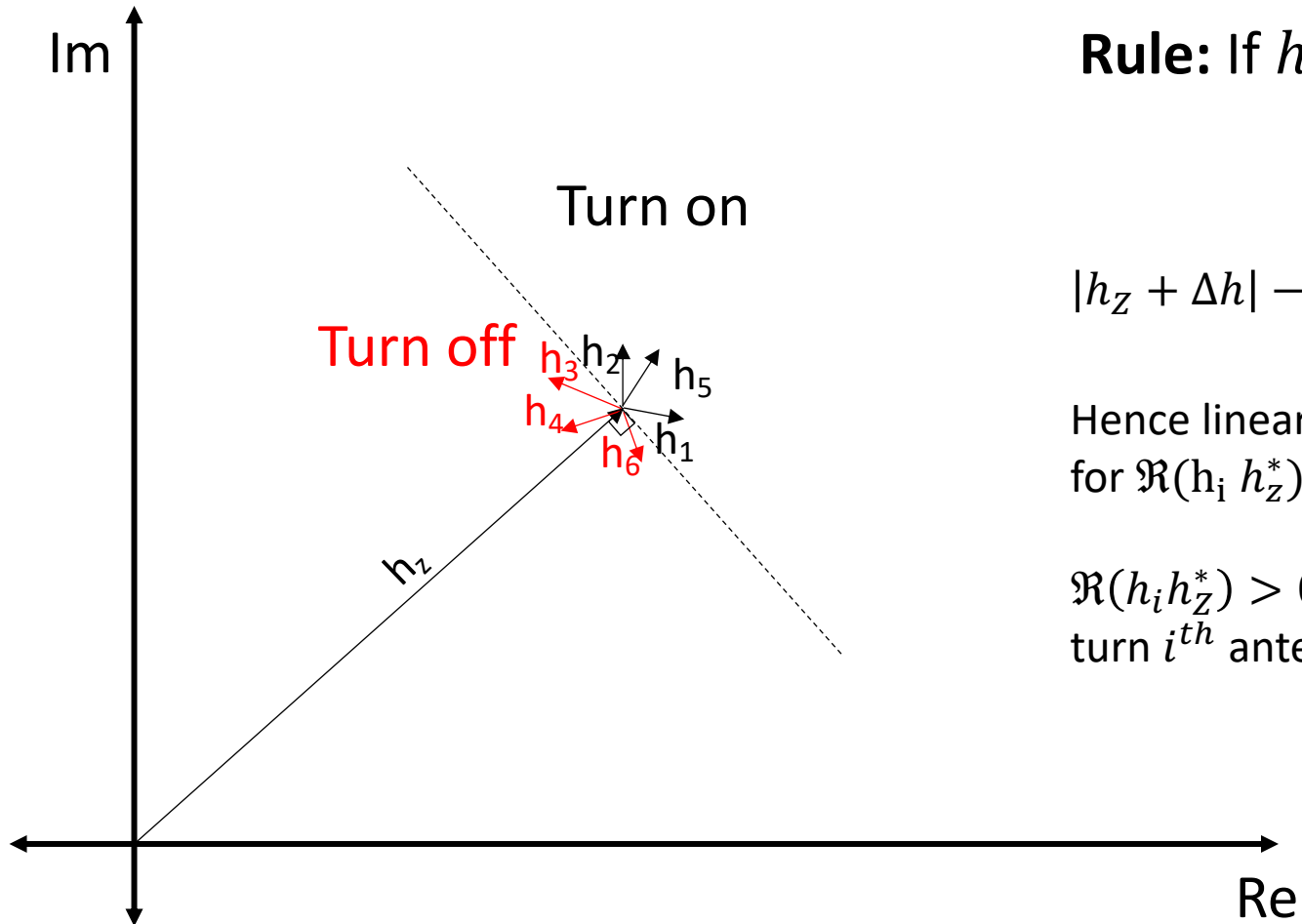
1	2	3	4	signal strength Measurement
0	1	1	0	0.8
1	1	1	0	0.9
0	0	1	1	1
1	0	0	0	1.1
1	1	0	1	1.1

Very far from optimal

Most random states have 1% the impact of the zeros state ( $h_Z$ )

The optimal state is  $\sim 10\times$  bigger

# Approach: Linear Regression



**Rule:** If  $h_i$  makes an acute angle with  $h_z$ , turn it on

$$|h_z + \Delta h| - |h_z| \approx |h_z| \Re\left(\frac{\Delta h}{h_z}\right) = \frac{\Re(\Delta h h_z^*)}{|h_z|}$$

Hence linear regression on  $|h_z + \sum_i b_i h_i|$  will solve for  $\Re(h_i h_z^*)$ , which gives us our answer

$\Re(h_i h_z^*) > 0 \Rightarrow h_i$  makes an acute angle with  $h_z \Rightarrow$  turn  $i^{th}$  antenna on

# Why not linear regression?

- Sensitive to outliers because it minimizes RMS error. Our approach looks at the median
  - This was a problem with an earlier version of RFocus
- Computationally expensive
- Our approach can fix the state of an antenna as soon as we are 95% sure about its value

# Why not linear regression?

- Our approach can fix the state of an antenna as soon as we are 95% sure about its value
  - If an antenna is faulty, it doesn't affect the others
  - Maybe only a small fraction of antennas are close enough to have an effect. Here, we can set those as soon as we are sure
  - The optimization algorithm is embarrassingly parallel for different antennas



# Our Approach: Majority Voting

Antenna ID →

1	2	3	4	signal strength Measurement	
0	1	1	0	0.8	
1	1	1	0	0.9	
0	0	1	1	1	→ Median
1	0	0	0	1.1	
1	1	0	1	1.1	

# Our Approach: Majority Voting

Antenna ID →	1	2	3	4	signal strength Measurement	
	0	1	1	0	0.8	This is bad, flip what we were doing
	1	1	1	0	0.9	
	0	0	1	1	1	→ Median
	1	0	0	0	1.1	This is good, Keep doing the same things
	1	1	0	1	1.1	

# Our Approach: Majority Voting

Antenna ID →	1	2	3	4	signal strength Measurement	
	01	10	10	01	0.8	This is bad, flip what we were doing
	10	10	10	01	0.9	
	0	0	1	1	1	→ Median
	1	0	0	0	1.1	This is good, Keep doing the same things
	1	1	0	1	1.1	
Majority Vote	1					

# Our Approach: Majority Voting

Antenna ID →	1	2	3	4	signal strength Measurement	
	01	10	10	01	0.8	This is bad, flip what we were doing
	10	10	10	01	0.9	
	0	0	1	1	1	→ Median
	1	0	0	0	1.1	This is good, Keep doing the same things
	1	1	0	1	1.1	
Majority Vote	1	0				

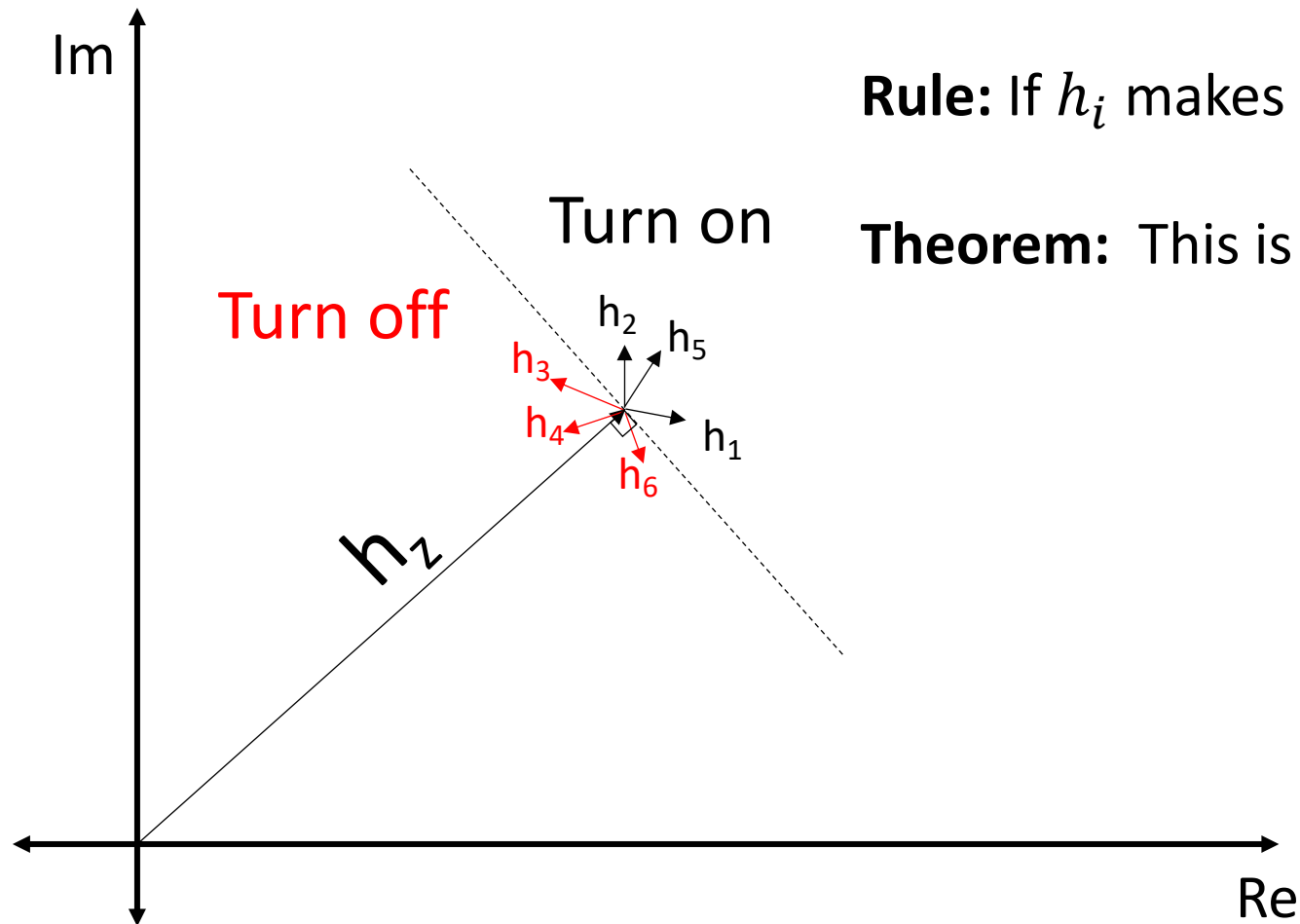
# Our Approach: Majority Voting

Antenna ID →	1	2	3	4	signal strength Measurement	
	01	10	10	01	0.8	This is bad, flip what we were doing
	10	10	10	01	0.9	
	0	0	1	1	1	→ Median
	1	0	0	0	1.1	This is good, Keep doing the same things
	1	1	0	1	1.1	
Majority Vote	1	0	0	1	→ Optimized State	

**Theorem 1:** Let assumptions 1, 2 and 3 hold. Then as  $N \rightarrow \infty$  and  $K \rightarrow \infty$ , majority voting finds a near-optimal solution. Here  $K$  is the number measurements.



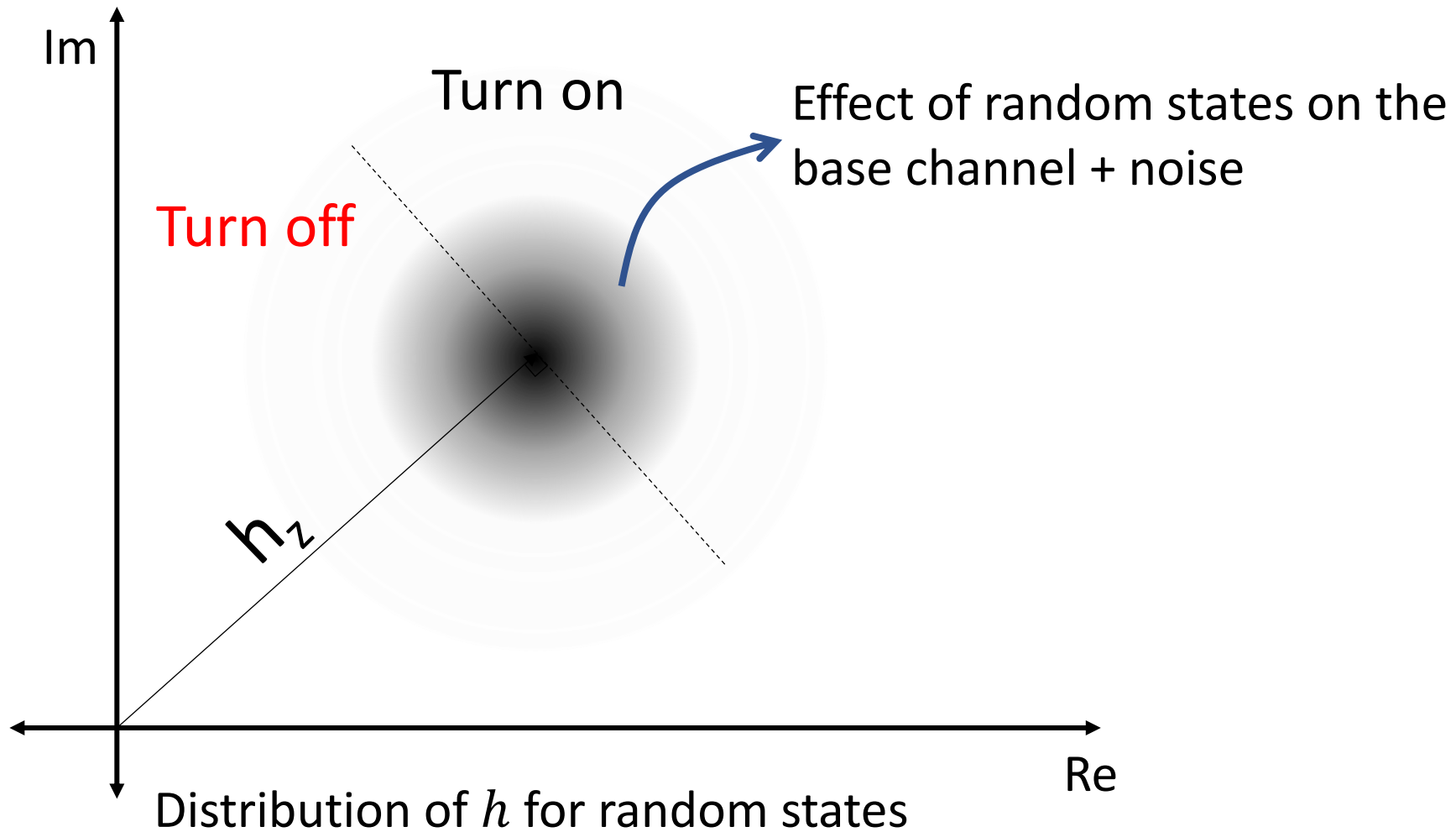
# Why Majority Voting is Optimal



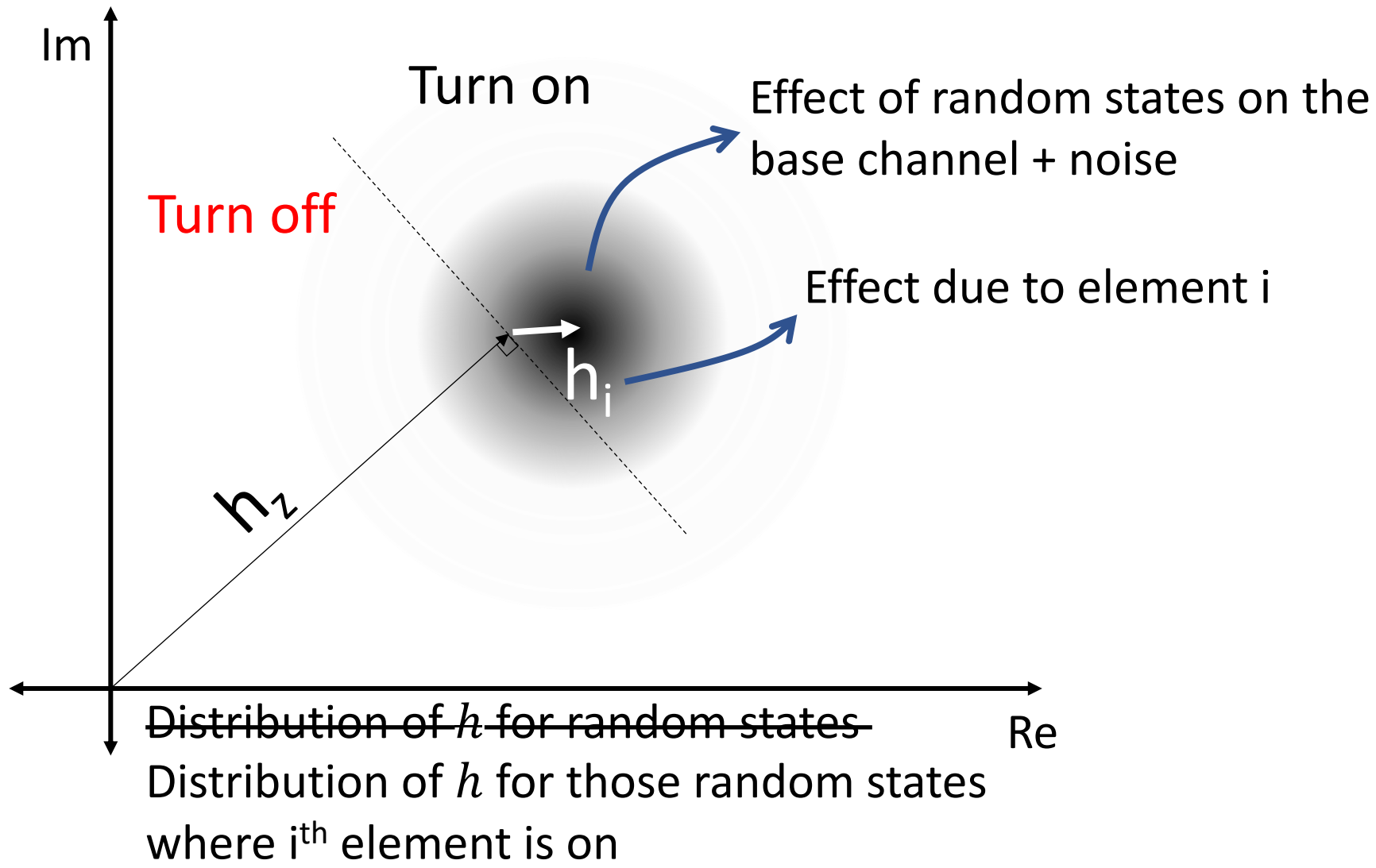
**Rule:** If  $h_i$  makes an acute angle with  $h_z$ , turn it on

**Theorem:** This is near optimal

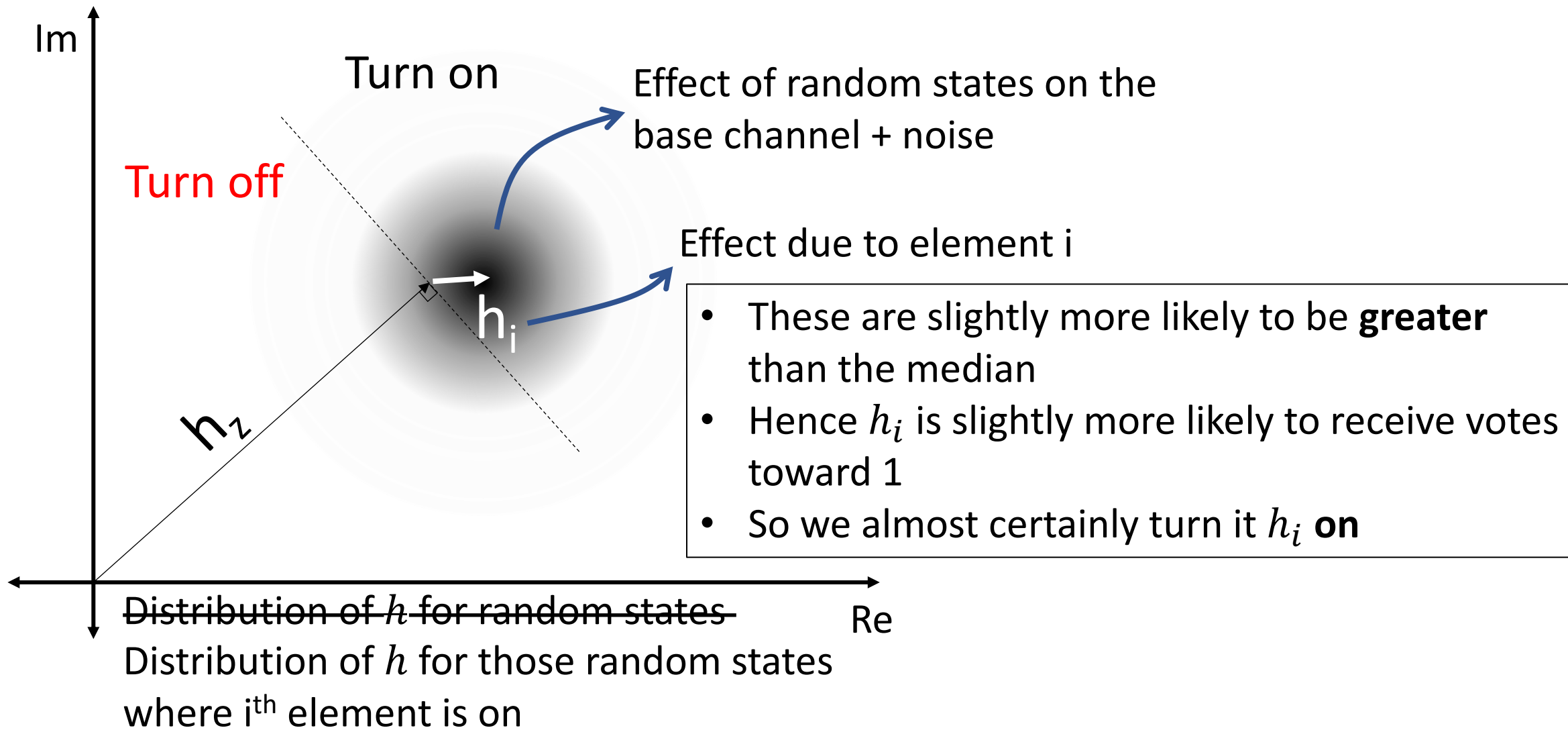
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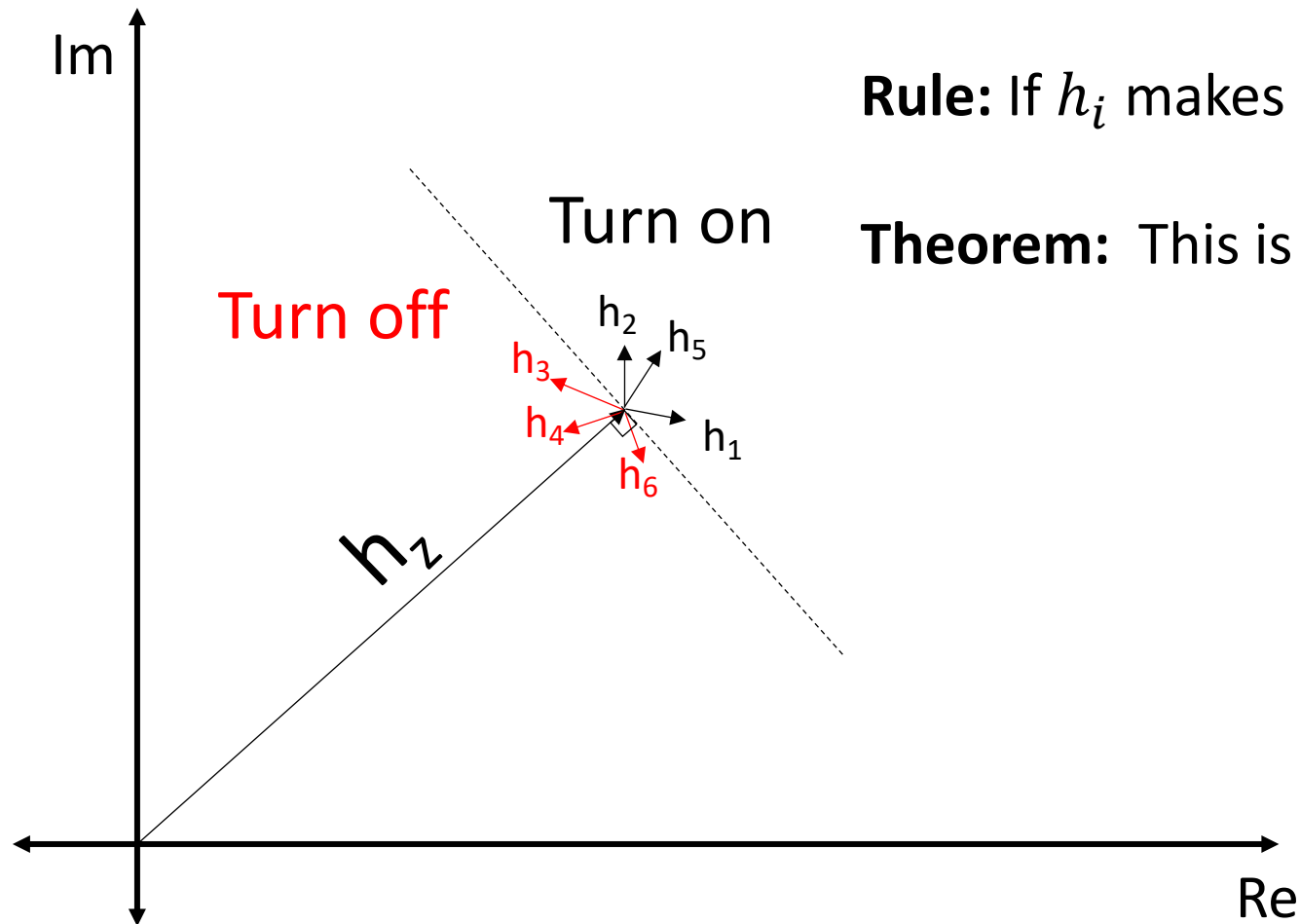
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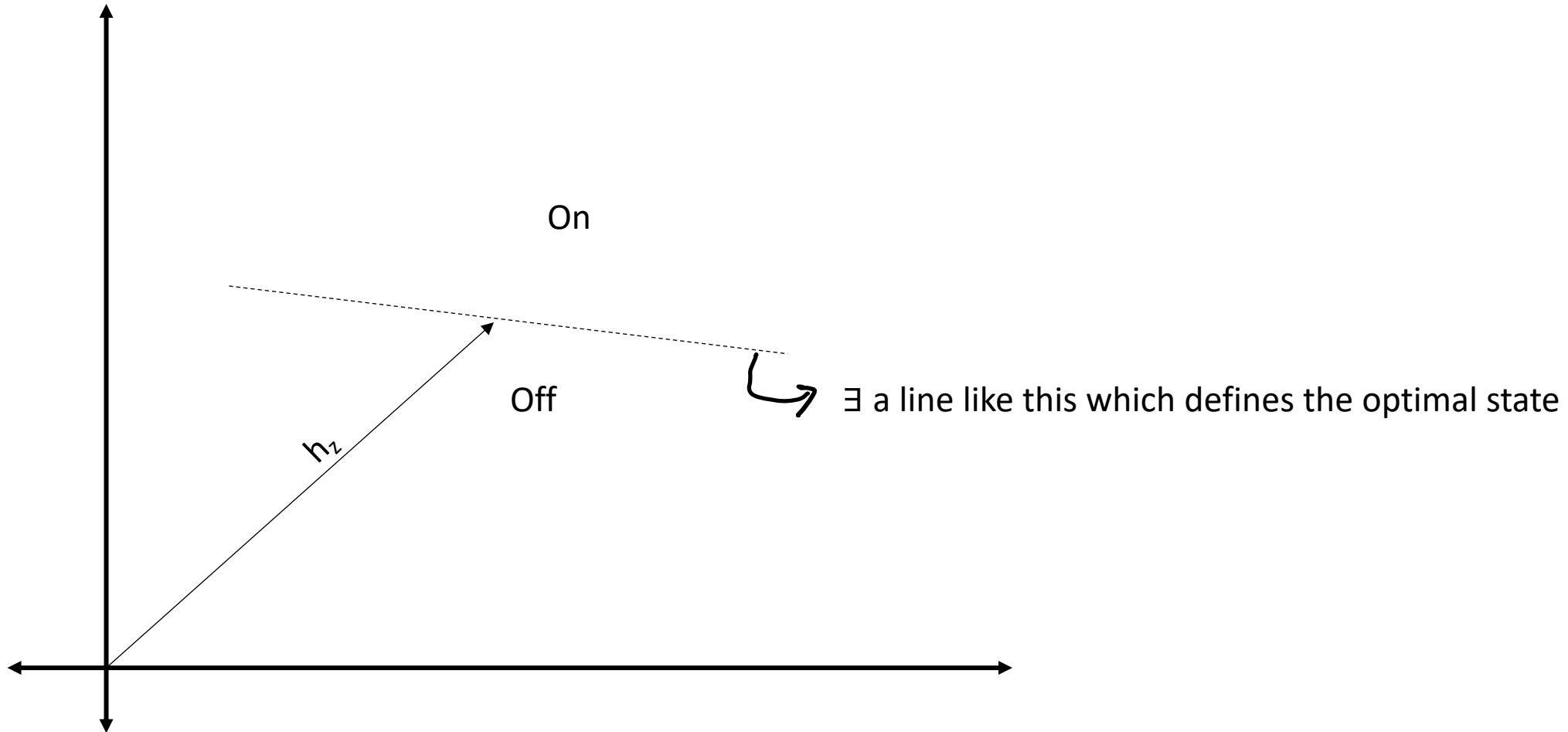
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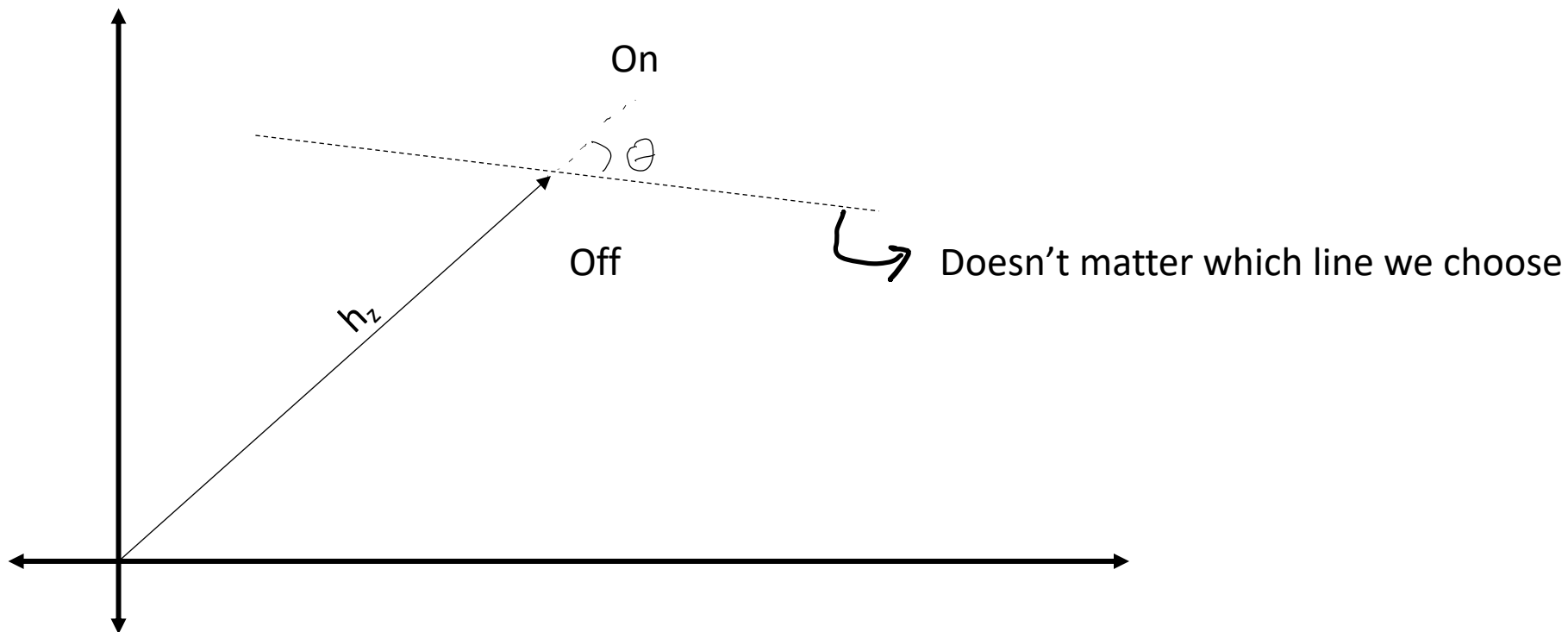
**Lemma 1:** Under assumptions 2 and 3, let  $\mathbf{b}_{OPT}$  be an optimal state assignment. Then  $b_{OPT,i} = 1$  if and only if  $\Re(h_i \cdot H(\mathbf{b}_{OPT})^*) > 0$ , where  $H(\mathbf{b}) = h_Z + \mathbf{h} \cdot \mathbf{b}$ .



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**Lemma 2:** Let  $\mathbf{b}_{OPT}$  be the optimal assignment that maximizes  $|h_z + \mathbf{h} \cdot \mathbf{b}|$  and  $\mathbf{b}_\perp$  be such that the  $i^{th}$  component  $\mathbf{b}_{\perp,i} = 1$  if and only if  $\Re(h_i \cdot h_z^*) > 0$ . As  $N \rightarrow \infty$ , if assumptions 1 and 3 hold, then  $\frac{|H(\mathbf{b}_{OPT})|}{|H(\mathbf{b}_\perp)|} < 1 + \epsilon \forall \epsilon > 0$ , with high probability.

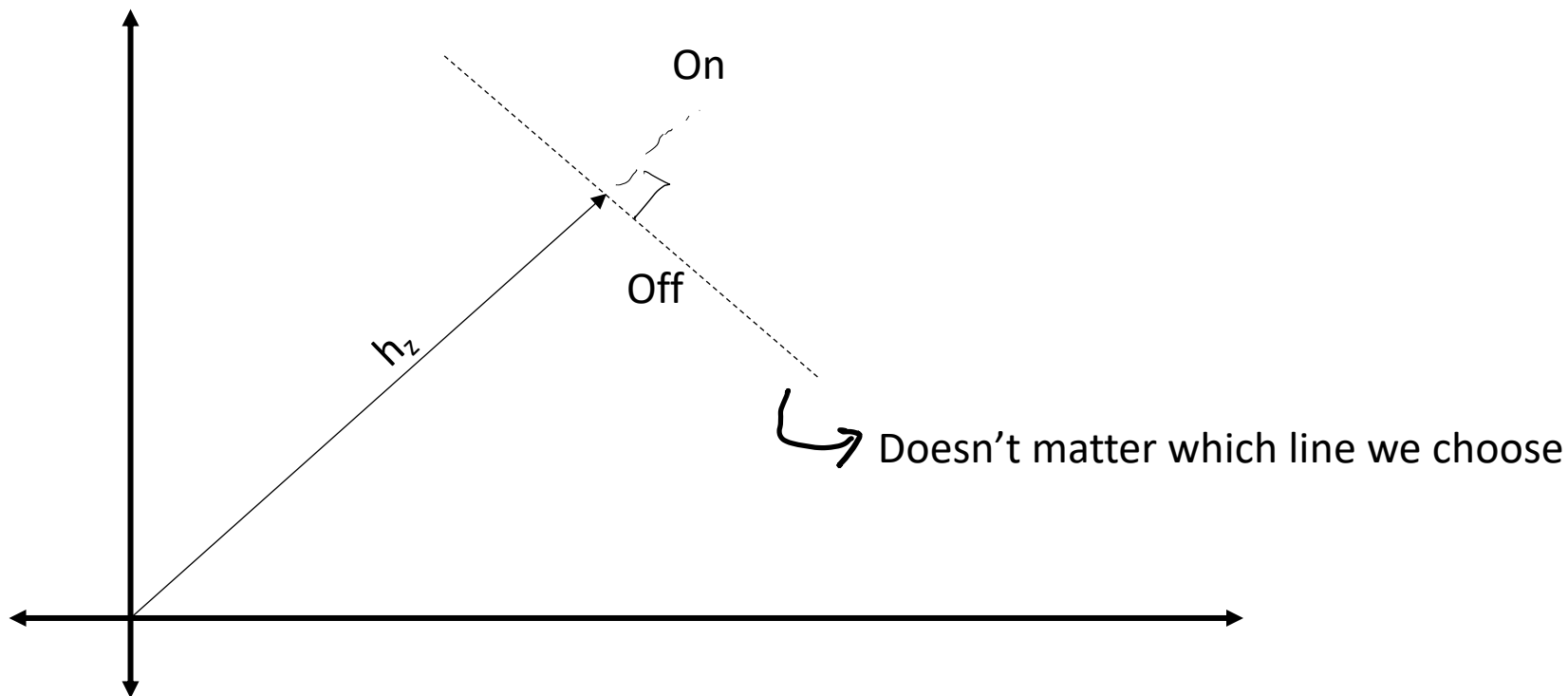
**Proof Intuition:**  $\max_{\theta} |\mathbf{b}_{\theta} \cdot \mathbf{h}| \approx \min_{\theta} |\mathbf{b}_{\theta} \cdot \mathbf{h}|$



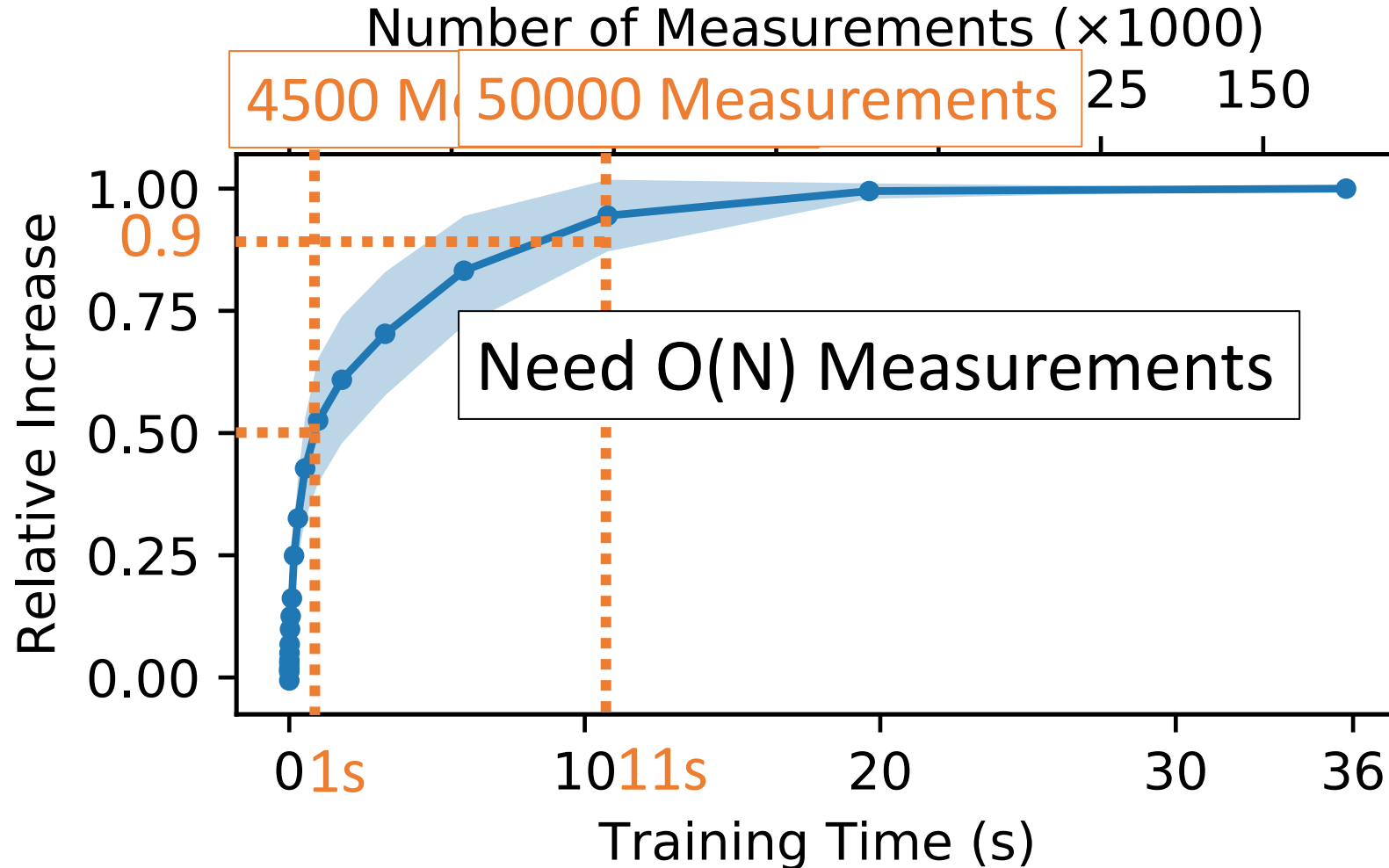


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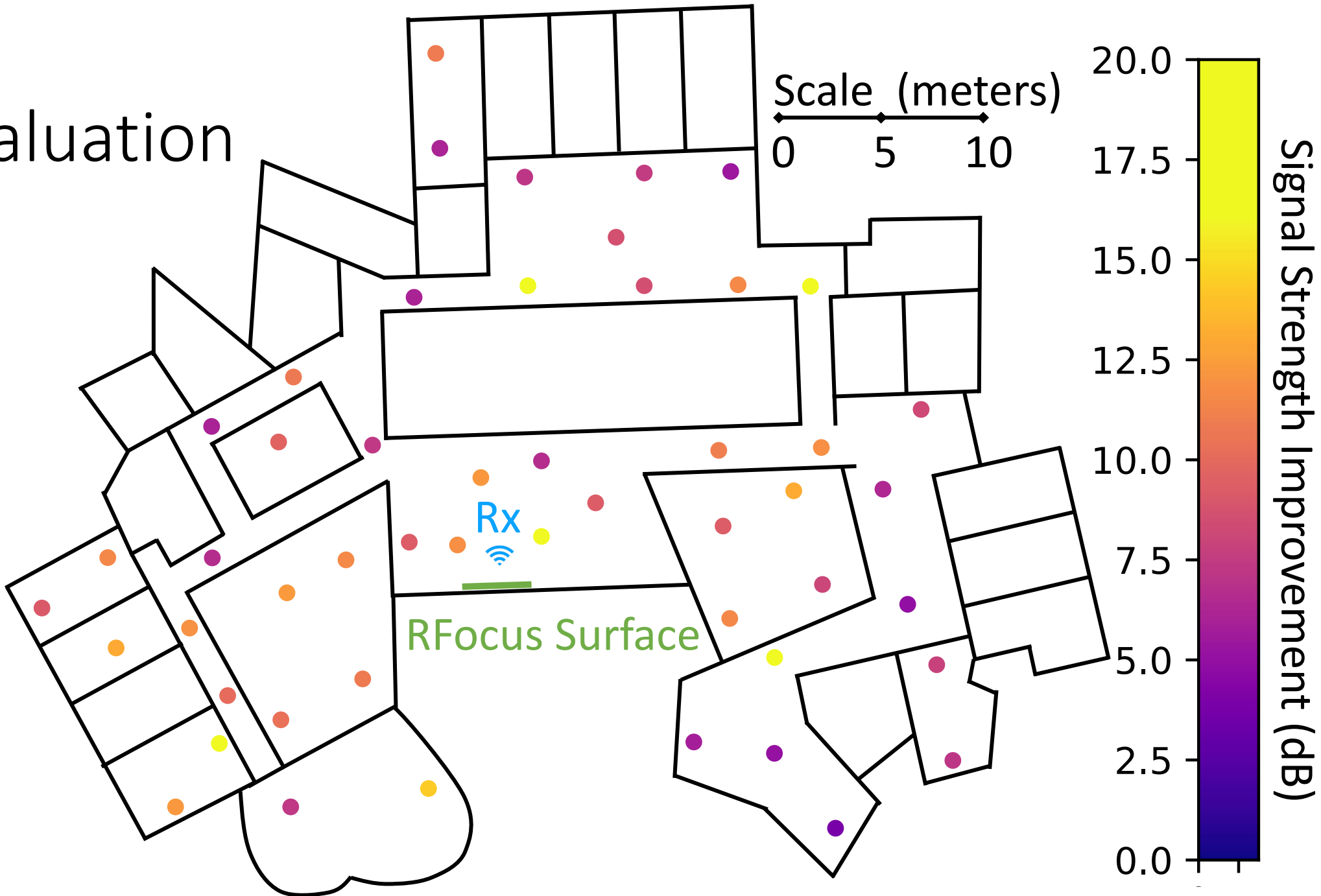
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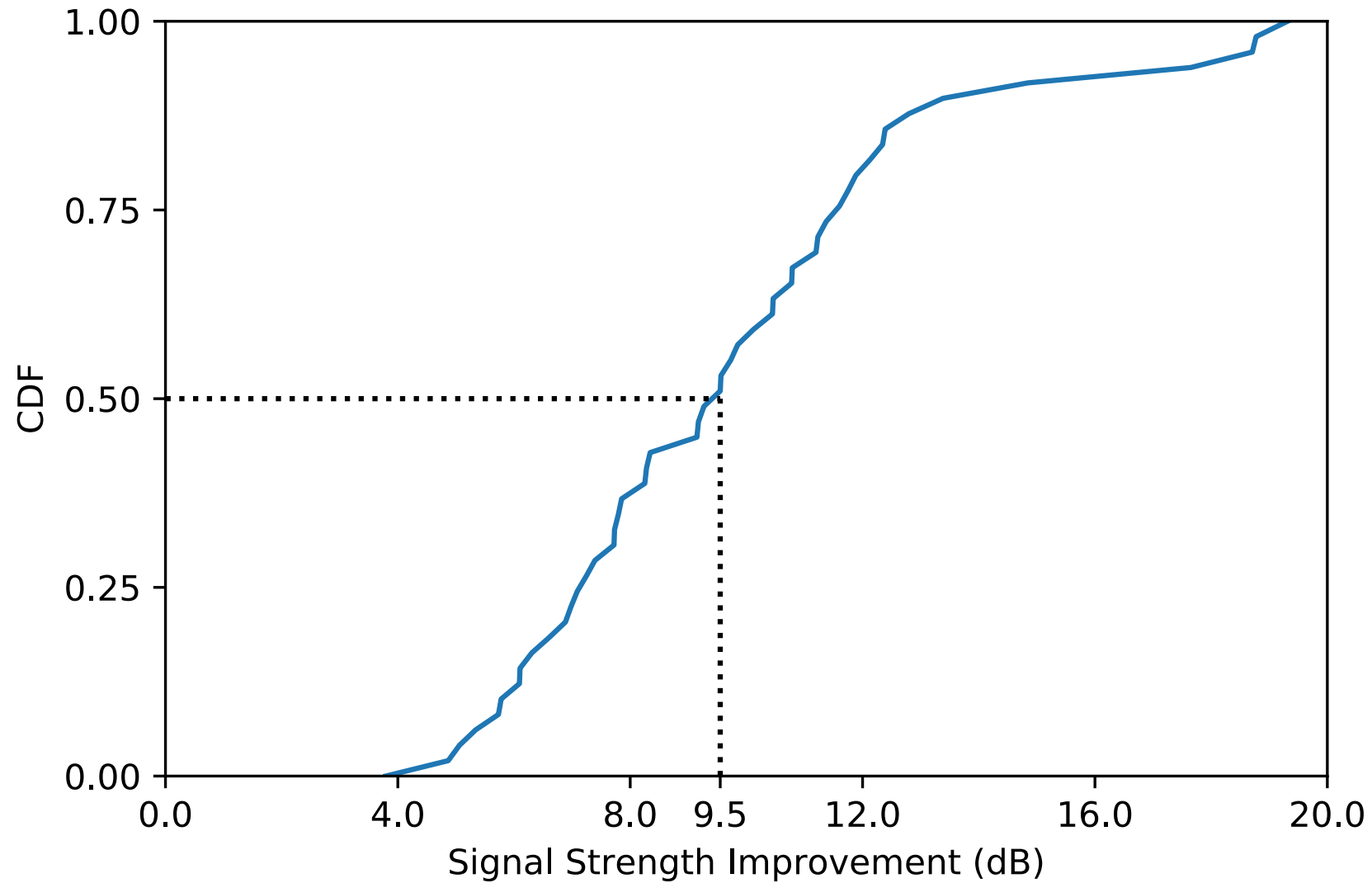
# How long does it take to train?



# Evaluation



# Evaluation



# Contributions

- Design of the antenna surface
- Near-optimal optimization algorithm that improves signal strength by  $\approx 10\times$ 
  - **Challenge:** Quantities we need to measure,  $h_i$ , are  $\sim 1$  million times smaller than the channel

This is just the beginning!





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