

ECE 515/ME 540: Problem Set 11: Problems and Solutions  
Linear Quadratic Regulators (LQR)

**Due:** Wednesday, December 4, 11:59pm

**Reading:** Course notes, Sections 10.3-10.7

1. **[LQR example]**

Consider the LTI system

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ 0 & -1 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u \quad x(0) = x_0$$

and infinite horizon cost  $\int_0^\infty \|x\|^2 + ru^2 dt$ , where  $r > 0$ .

- (a) Determine what Theorem 10.6 implies about the LQR for this problem.

**Solution:** The system is in KCCF form and the uncontrollable matrix  $A_c = [-1]$  is Hurwitz so the system is stabilizable (but not controllable). Also,  $Q = I = C^T C$  for  $C = I$  and  $A, C$  is observable. both parts of Theorem 10.6 apply and imply that the closed loop system matrix  $A_{cl}$  for the optimal feedback control is Hurwitz stable and the matrix  $\bar{P}$  is positive definite.

- (b) Compute by hand the LQR matrix  $\bar{P}$ , the optimal feedback control law, the closed loop state matrix  $A_{cl}$ , and the poles of the closed loop system.

**Solution:** Setting  $\bar{P} = \begin{bmatrix} a & b \\ b & c \end{bmatrix}$  the algebraic Riccati equation (ARE) becomes

$$\begin{bmatrix} 0 & 0 \\ a-b & b-c \end{bmatrix} + \begin{bmatrix} 0 & a-b \\ 0 & b-c \end{bmatrix} + \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} + \frac{1}{r} \begin{bmatrix} a^2 & ab \\ ab & b^2 \end{bmatrix} = 0$$

from which we find:

$$\bar{P} = \begin{bmatrix} \sqrt{r} & \\ \frac{r}{1+\sqrt{r}} & \frac{1}{2} + \frac{r}{1+\sqrt{r}} - \frac{r}{2(1+\sqrt{r})^2} \end{bmatrix}$$

$$\text{or } c = \frac{1 + 2\sqrt{r} + 2r + 2r\sqrt{r}}{2(1 + \sqrt{r})^2}$$

$$u = -\frac{1}{r} B^T \bar{P} x = - \begin{bmatrix} \frac{1}{\sqrt{r}} & \frac{1}{1+\sqrt{r}} \end{bmatrix} x$$

$$A_{cl} = \begin{bmatrix} -\frac{1}{\sqrt{r}} & \frac{\sqrt{r}}{1+\sqrt{r}} \\ 0 & -1 \end{bmatrix}$$

The poles of the closed loop system are  $-\frac{1}{\sqrt{r}}$  and  $-1$ . (Pole -1 is associated with the uncontrollable stable mode and is not moved by feedback.)

- (c) Comment on how the feedback law and poles vary as  $r$  gets very large (expensive control) or very small (cheap control).

**Solution:** For  $r$  large the feedback multipliers get small and the controllable pole gets close to 0 (which is the open loop pole). For  $r$  small the first feedback multiplier gets large and the controllable pole gets large magnitude negative corresponding to a fast response time. However, the pole -1 is not controllable so the overall response time will be limited by that pole.

- (d) Find the eigenvalues of the Hamiltonian matrix  $\mathcal{H}$  by hand. Is your answer consistent with part (b)?

**Solution:**

$$\begin{aligned} \det(Is - \mathcal{H}) &= \det \begin{bmatrix} s & -1 & \frac{1}{r} & 0 \\ 0 & s+1 & 0 & 0 \\ 1 & 0 & s & 0 \\ 0 & 1 & 1 & s-1 \end{bmatrix} \\ &= s \det \begin{bmatrix} s+1 & 0 & 0 \\ 0 & s & 0 \\ 1 & 1 & s-1 \end{bmatrix} + \det \begin{bmatrix} -1 & \frac{1}{r} & 0 \\ s+1 & 0 & 0 \\ 1 & 1 & s-1 \end{bmatrix} \\ &= s^2(s^2 - 1) - (s^2 - 1)\frac{1}{r} = \left(s^2 - \frac{1}{r}\right)(s^2 - 1). \end{aligned}$$

The four eigenvalues of  $\mathcal{H}$  are  $\pm 1, \pm \frac{1}{\sqrt{r}}$ , which is consistent with part (b) – they are the two stable eigenvalues found in part (b) and their negatives.

## 2. [Variation of an LQR example]

Consider the same linear system model as in Problem 1.

- (a) Suppose the same cost function is used as in Problem 1 but with  $\|x\|^2$  replaced by  $x_1^2$ . Explain how the LQR regulator and optimal cost are different from those found in Problem 1.

**Solution:** There is very little difference. The  $Q$  matrix is changed from the identity matrix to  $\begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix}$  so the corresponding  $C$  matrix becomes  $C = [1 \ 0]$ . The system remains detectable and observable and Theorem 10.6 applies as before. We find:

$$\bar{P} = \begin{bmatrix} \sqrt{r} & \frac{r}{1+\sqrt{r}} \\ \frac{r}{1+\sqrt{r}} & \frac{r}{2(1+\sqrt{r})^2} \end{bmatrix}$$

The only difference is that the term  $\frac{1}{2}$  is dropped from the bottom right entry of  $\bar{P}$ . In other words, the optimal cost  $V^o(x) = x^T \bar{P} x$  is smaller by  $\frac{1}{2}x_2(0)^2$ . That is due to the fact that  $\int_0^\infty (x^2(t))^2 dt = \int_0^\infty (x_2(0)e^{-t})^2 dt = x_2^2(0) \int_0^\infty e^{-2t} dt = \frac{1}{2}x_2(0)^2$  no matter what control is used. The same feedback control is optimal and  $A_{cl}$  and the closed loop poles are the same as in Problem 1.

- (b) The change in part (a) gives rise to a SISO system. Find the open loop transfer function  $P(s) = C(Is - A)^{-1}B$  for part (a) and find the set of all solutions to the symmetric root locus equation (10.31) in the course notes, namely,  $1 + \frac{1}{r}P(s)P(-s) = 0$ . Do you recover the same two negative closed loop roots as before? Explain.

**Solution:** Computation as usual yields  $P(s) = C(Is - A)^{-1}B = \frac{1}{s}$ . At first this might seem surprising but since transfer functions don't take into account the uncontrollable subsystem we get the same transfer function as a pure integrator:  $\dot{x}_1 = u$ . The symmetric root locus equation becomes  $1 - \frac{1}{rs^2} = 0$  which has solutions  $s = \pm \frac{1}{\sqrt{r}}$ . Therefore this reflects only the controllable stable pole  $-\frac{1}{\sqrt{r}}$  and its negative. It makes sense that since the system is not minimal (because it is not observable), the symmetric root locus equations would pertain to a minimal system model realization only.

- (c) Suppose the same cost function is used as in Problem 1 but with  $\|x\|^2$  replaced by  $x_2^2$ . Explain how the LQR regulator and optimal cost are different from those found in Problem 1. Find  $\bar{P}$  and the optimal control for this variation.

**Solution:** There is a huge difference. The  $Q$  matrix is changed from the identity matrix to  $\begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix}$  so the corresponding  $C$  matrix becomes  $C = [0 \ 1]$ . The system is no longer observable and it is not even detectable, so Theorem 10.6 does not apply. Since the control cannot effect  $x_2$  the optimal choice is to let  $u(t) = 0$  for all  $t$ . We then see that  $\bar{P} = \begin{bmatrix} 0 & 0 \\ 0 & \frac{1}{2} \end{bmatrix}$ . Also  $A_{cl} = A$  so the closed loop poles are 0 and 1.  $A_{cl}$  is not Hurwitz. We see that  $x_2(t) \rightarrow 0$  and  $x_1(t) \rightarrow x_1(0) + x_2(0)/2$  as  $t \rightarrow \infty$ .

### 3. [LQR with no control]

Consider an LQR problem for an LTI system of the form  $\dot{x} = Ax$  with cost  $\int_0^\infty x^T Q x dt$  where  $Q = C^T C$  for some matrix  $C$ . In other words, it is a general infinite horizon LQR problem except  $B = 0$ .

- (a) Under what conditions on  $A$  and  $C$  does Theorem 10.6 part (i) apply and what does it imply under those conditions? Has the ARE appeared earlier in the course?

**Solution:** The system is stabilizable if and only if it is stable – meaning  $A$  is Hurwitz. If  $A$  is Hurwitz the system is also detectable. So part (i) of Theorem 10.6 applies if  $A$  is Hurwitz. If  $A$  is Hurwitz we can conclude there exists a positive semi-definite solution to the ARE which is unique in the class of positive semi-definite matrices. The ARE in this case is  $A^T \bar{P} + \bar{P} A + Q = 0$ , which is the Lyapunov stability equation encountered in Chapter 4. See a connection to the proof of Theorem 4.6.

- (b) Under what conditions on  $A$  and  $C$  does Theorem 10.6 part (ii) apply and what does it imply under those conditions?

**Solution:** In order for Theorem 10.6 to apply we need to assume that  $A$  is Hurwitz (so  $(A, B)$  is detectable) and also that  $(A, C)$  is observable. In this case, in addition to the conclusions in part (i), we know that  $\bar{P}$  is positive definite.

### 4. [Stable subspace of the Hamiltonian matrix for LQR]

Consider the LQR problem for the LTI system  $\dot{x} = Ax + Bu$  and infinite horizon cost  $\int_0^\infty x^T Q x + u^T R u dt$  such that  $Q = C^T C$ . Assume  $R$  is positive definite,  $(A, B)$  is stabilizable, and  $(A, C)$  is detectable. Let  $\bar{P}, A_{cl}$  and  $\mathcal{H}$  be as in Chapter 10 of the notes.

- (a) Show that  $\mathcal{H} \begin{bmatrix} I \\ \bar{P} \end{bmatrix} = \begin{bmatrix} I \\ \bar{P} \end{bmatrix} A_{cl}$ .

**Solution:** We need to show

$$\begin{bmatrix} A & -BR^{-1}B^T \\ -Q & -A^T \end{bmatrix} \begin{bmatrix} I \\ \bar{P} \end{bmatrix} = \begin{bmatrix} I \\ \bar{P} \end{bmatrix} A_{cl}$$

or equivalently

$$A - BR^{-1}B^T\bar{P} = A_{cl} \quad (1)$$

$$A^T\bar{P} + Q = -\bar{P}A_{cl} \quad (2)$$

Equation (1) gives the correct expression for  $A_{cl}$ ; it is the closed loop matrix  $A - BK$  for the optimal feedback control  $u = -Kx$  for  $K = R^{-1}B^T\bar{P}$ . Substituting this expression for  $A_{cl}$  into (2) reduces the equation to the ARE.

(b) Show that  $e^{\mathcal{H}t} \begin{bmatrix} I \\ \bar{P} \end{bmatrix} = \begin{bmatrix} I \\ \bar{P} \end{bmatrix} e^{A_{cl}t}$ .

**Solution:** By part (a) and argument by induction on  $k$ ,  $\mathcal{H}^k \begin{bmatrix} I \\ \bar{P} \end{bmatrix} = \begin{bmatrix} I \\ \bar{P} \end{bmatrix} A_{cl}^k$  for  $k \geq 0$ . Multiplying both sides by  $t^k/k!$  and summing over  $k$  from 0 to  $\infty$  yields the equation to be shown.

(c) Part (b) shows that the columns of  $\begin{bmatrix} I \\ \bar{P} \end{bmatrix}$  span the stable subspace of  $\mathcal{H}$ . This implies the following method for finding  $\bar{P}$  using  $\mathcal{H}$ . First identify a  $2n \times n$  matrix with columns that span the stable subspace of  $\mathcal{H}$  and then do elementary column operations to make the upper half of the matrix the identity matrix. Then the bottom half will be  $\bar{P}$ . Illustrate this method by using it to find the  $\bar{P}$  matrix  $\bar{P} = [p]$  for the scalar LQR problem  $\dot{x} = ax + bu$  with cost  $\int_0^\infty qx^2 + ru^2 dt$ . Verify your answer by comparing to the solution of the ARE.

**Solution:** The characteristic polynomial of  $\mathcal{H} = \begin{bmatrix} a & -b^2/r \\ -c^2 & -a \end{bmatrix}$  is  $s^2 - a^2 - b^2q/r$ , which has roots  $\lambda_1 = -\sqrt{a^2 + b^2q/r}$  and  $\lambda_2 = \sqrt{a^2 + b^2q/r}$ . The eigenvector  $v^1$  for the stable root  $\lambda_1$  is determined by  $(\lambda_1 I - \mathcal{H})v^1 = \vartheta$ . Assuming  $v^1$  is scaled so it has the form  $v^1 = \begin{bmatrix} 1 \\ p \end{bmatrix}$  yields the equation  $\begin{bmatrix} -\sqrt{a^2 + b^2q/r} - a & b^2/r \end{bmatrix} \begin{bmatrix} 1 \\ p \end{bmatrix} = 0$  or  $p = \frac{r}{b^2} \left[ \sqrt{a^2 + b^2q/r} + a \right]$ . This is also the positive solution of the ARE which in this case reduces to  $\frac{b^2 p^2}{r} - 2ap - c^2 = 0$ .

#### 5. [An observability equivalence]

Let  $Q = C^T C$ . Show that  $(A, C)$  is an observable pair if and only if  $(A, Q)$  is an observable pair. (Hint: Apply the eigenvector criterion for observability – closely related to the Hautus Rosenbrock criterion.)

**Solution:** By the eigenvector criterion,  $(A, C)$  is observable if and only if  $Cv \neq \vartheta$  for every eigenvector  $v$  of  $A$  and  $(A, C^T C)$  is observable if and only if  $C^T C v \neq \vartheta$  for every eigenvector  $v$  of  $A$ . So it is enough to show that  $Cv \neq \vartheta$  if and only if  $C^T C v \neq \vartheta$  for any vector  $v$  or equivalently,  $Cv = \vartheta$  if and only if  $C^T C v = \vartheta$  for any vector  $v$ .

If  $Cv = \vartheta$  then left multiplying both sides by  $C^T$  implies  $C^T C v = \vartheta$ . If  $C^T C v = \vartheta$  then left multiplying both sides by  $v^T$  yields  $v^T C^T C v = 0$  or equivalently  $\|Cv\|^2 = 0$  which is equivalent to  $Cv = \vartheta$ .