

Plan of the Lecture

- ▶ **Review:** control design using frequency response: PI/lead
- ▶ **Today's topic:** control design using frequency response: PD/lag, PID/lead+lag

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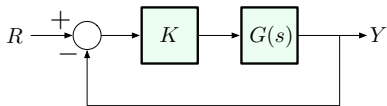
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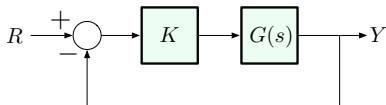
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Reading: FPE, Chapter 6

Review: Bode's Gain-Phase Relationship

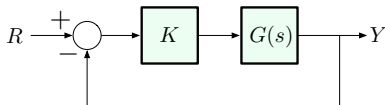


Review: Bode's Gain-Phase Relationship



Assuming that $G(s)$ is *minimum-phase* (i.e., has no RHP zeros), we derived the following for the Bode plot of $KG(s)$:

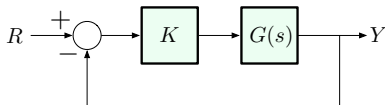
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We can state this succinctly as follows:

Gain-Phase Relationship. Far enough from break-points,

$$\text{Phase} \approx \text{Magnitude Slope} \times 90^\circ$$

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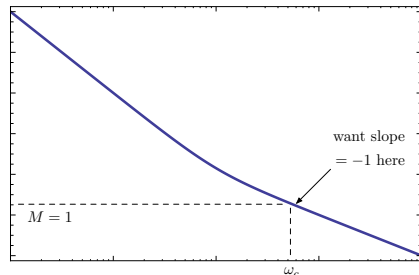
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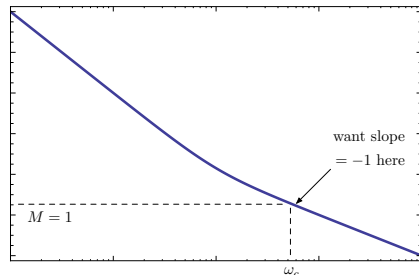


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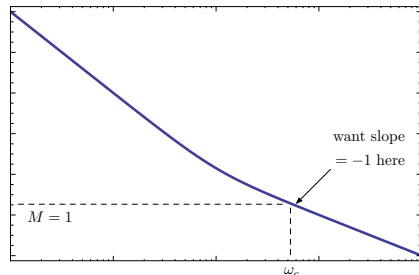
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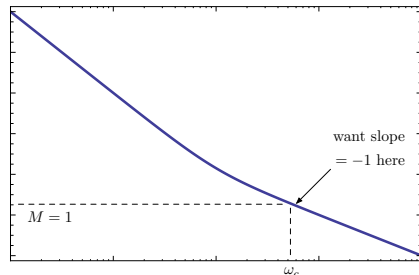
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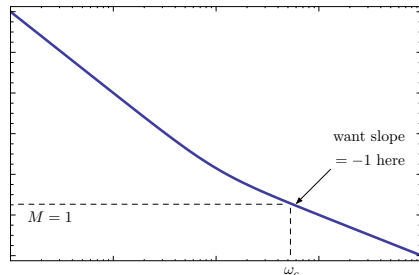
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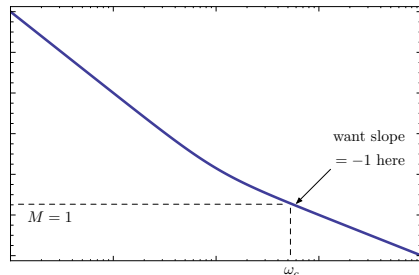
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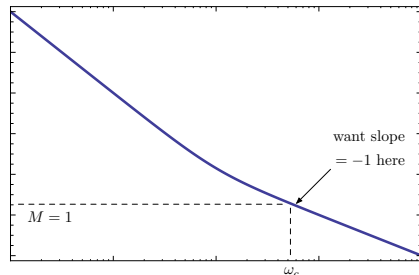
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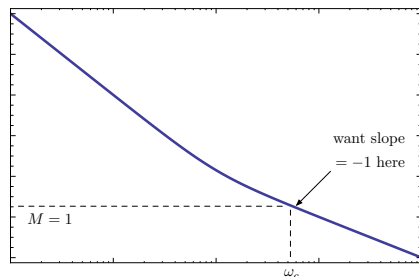
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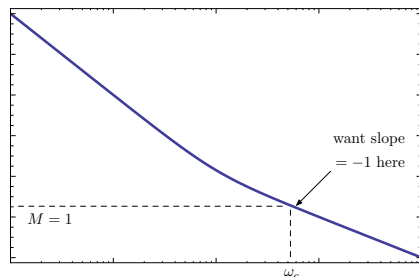
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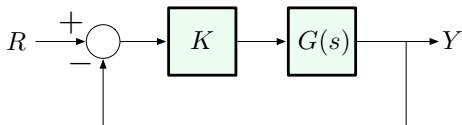


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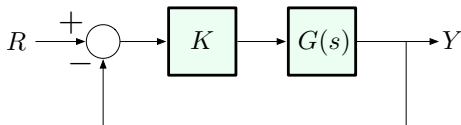
(Similar considerations apply when M -plot has positive slope – depends on the t.f.)

Control Design Using Frequency Response



Bode's Gain-Phase Relationship suggests that we can shape the time response of the *closed-loop* system by choosing K (or, more generally, a dynamic controller $KD(s)$) to tune the Phase Margin.

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In particular, from the quantitative Gain-Phase Relationship,

$$\text{Magnitude slope}(\omega_c) = -1 \quad \implies \quad \text{Phase}(\omega_c) \approx -90^\circ$$

— which gives us PM of 90° and consequently **good damping**.

Lead Controller Design Using Frequency Response

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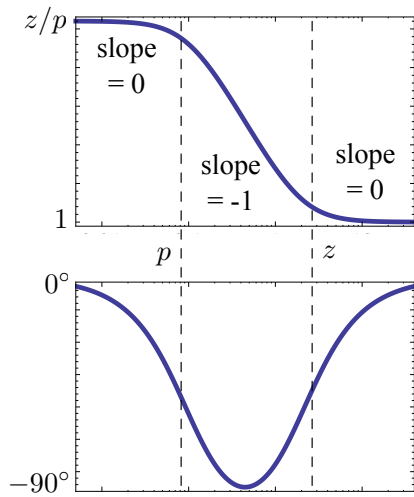
This is an intuitive procedure, but it's not very precise, requires trial & error.

Lag Compensation: Bode Plot

$$D(s) = \frac{s + z}{s + p} = \frac{z \frac{s}{z} + 1}{p \frac{s}{p} + 1}, \quad z \gg p$$

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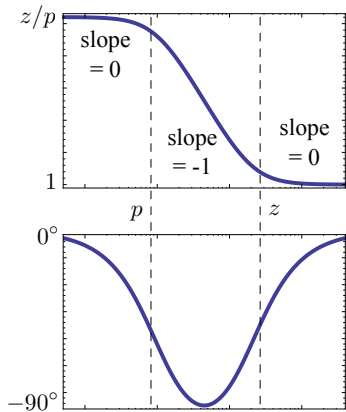
$$D(s) = \frac{s + z}{s + p} = \frac{z \frac{s}{z} + 1}{p \frac{s}{p} + 1}, \quad z \gg p$$



▶ $\frac{j\omega + z}{j\omega + p} \xrightarrow{\omega \rightarrow \infty} 1$
so $M \rightarrow 1$ at high frequencies

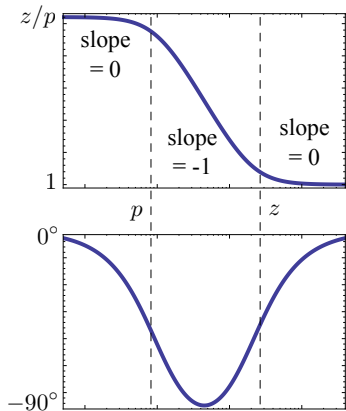
▶ subtracts phase, hence the term “phase lag”

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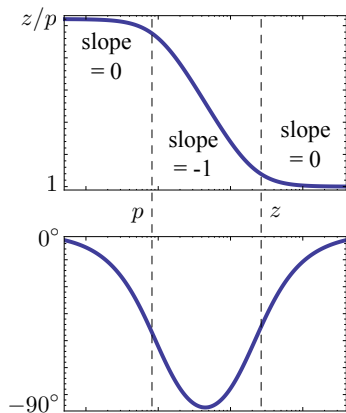


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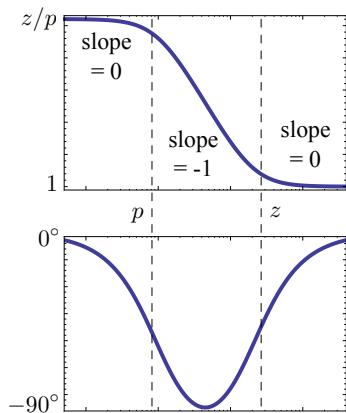


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steady-state tracking error:

$$e(\infty) = \left. \frac{sR(s)}{1 + D(s)G(s)} \right|_{s=0}$$

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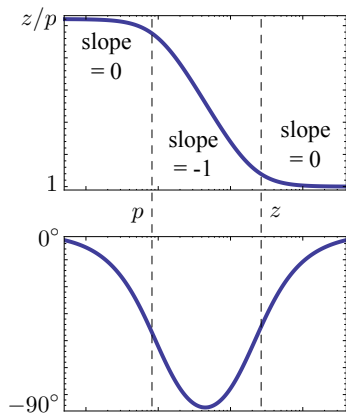
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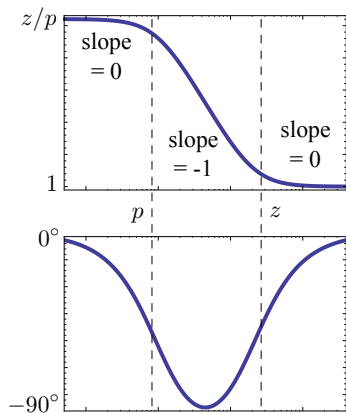
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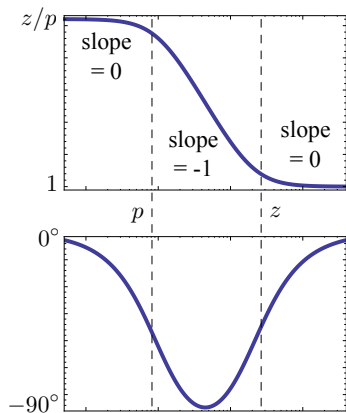
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- \blacktriangleright to mitigate this, choose both z and p very small, while maintaining desired ratio z/p

Example

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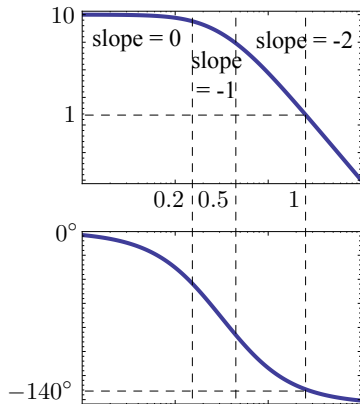
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- ▶ this is different from what we did for lead (used p and z to shape PM, then chose K to get desired bandwidth spec)

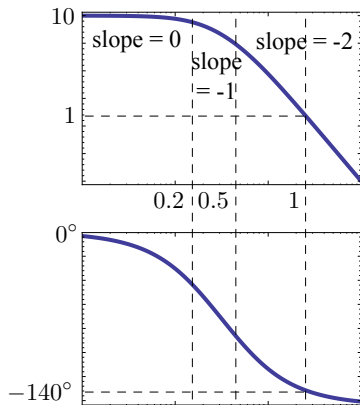
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Check Bode plot of $G(s)$ to see how much PM it already has:



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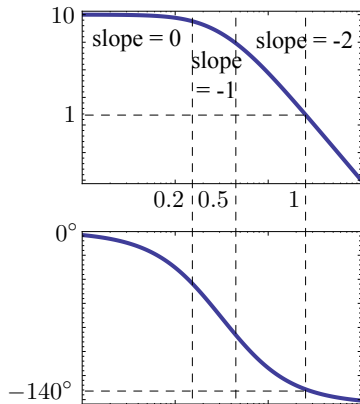
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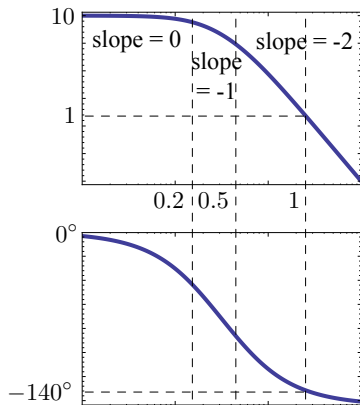


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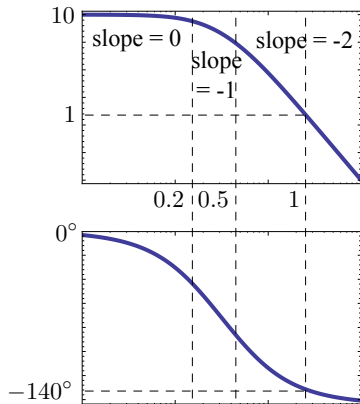
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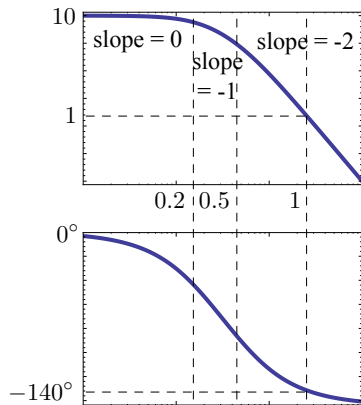
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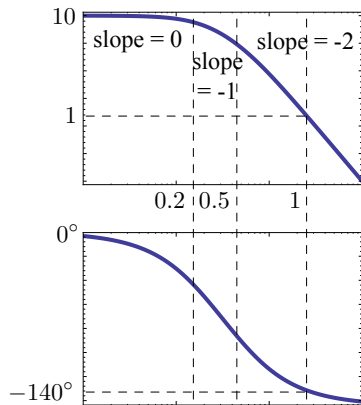
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A conservative choice (to allow some slack) is $K = 1/2.5 = 0.4$, gives $\omega_c \approx 0.52$, PM $\approx 65^\circ$

Step 2: Choose z & p to Shape Tracking Error

So far: $KG(s) = \frac{0.4 \cdot 10}{\left(\frac{s}{0.2} + 1\right) \left(\frac{s}{0.5} + 1\right)}$

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$$e(\infty) = \frac{1}{1 + KG(s)} \Big|_{s=0} = \frac{1}{1 + 4} = \frac{1}{5} = 20\% \quad (\text{too high})$$

To have $e(\infty) \leq 10\%$, need $KD(0)G(0) \geq 9$:

$$e(\infty) = \frac{1}{1 + KD(0)G(0)} \leq \frac{1}{1 + 9} = 10\%.$$

So, we need

$$D(0) = \frac{s + z}{s + p} \Big|_{s=0} = \frac{z}{p} \geq \frac{9}{4} = 2.25 \quad \text{--- say, } z/p = 2.5$$

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Not to distort PM and ω_c , let's pick z and p an order of magnitude smaller than $\omega_c \approx 0.5$: $z = 0.05$, $p = 0.02$

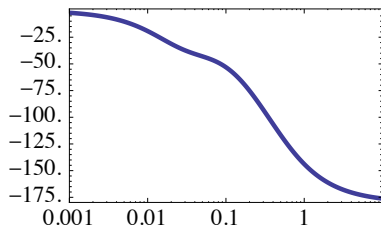
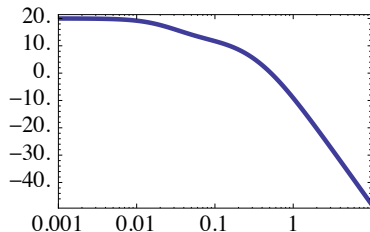
Overall Design

Plant:

$$G(s) = \frac{10}{\left(\frac{s}{0.2} + 1\right) \left(\frac{s}{0.5} + 1\right)}$$

Controller:

$$KD(s) = 0.4 \frac{s + 0.05}{s + 0.02}$$

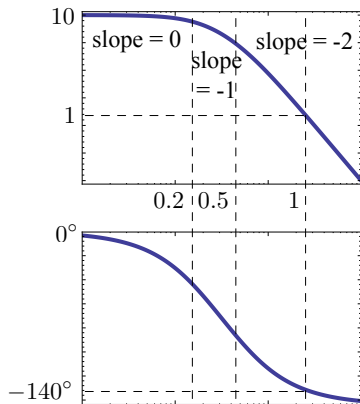


— the design still needs a bit of refinement ...

Lead & Lag Compensation

Let's combine the advantages of PD/lead and PI/lag.

Back to our example:
$$G(s) = \frac{10}{\left(\frac{s}{0.2} + 1\right) \left(\frac{s}{0.5} + 1\right)}$$

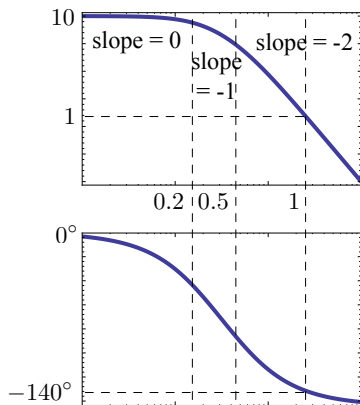


- ▶ from Matlab, $\omega_c \approx 1$
- ▶ PM $\approx 40^\circ$

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New objectives:

- ▶ $\omega_{BW} \geq 2$
- ▶ PM $\geq 60^\circ$
- ▶ $e(\infty) \leq 1\%$ for const. ref.

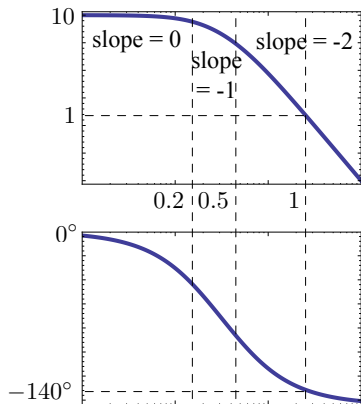
Lead & Lag Compensation

What we got before, with lag only:

- ▶ Improved PM by adjusting K to decrease ω_c .
- ▶ This gave $\omega_c \approx 0.5$, whereas now we want a larger ω_c (recall: $\omega_{BW} \in [\omega_c, 2\omega_c]$, so $\omega_c = 0.5$ is too small)

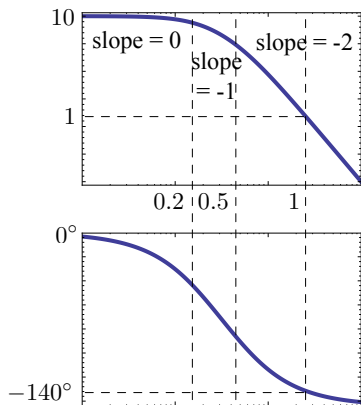
So: we need to reshape the phase curve using lead.

Lead & Lag Compensation



Step 1. Choose K to get $\omega_c \approx 2$
(before lead)

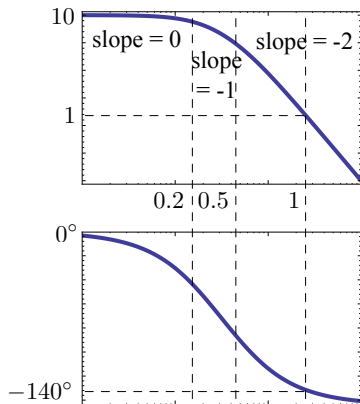
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Lead & Lag Compensation

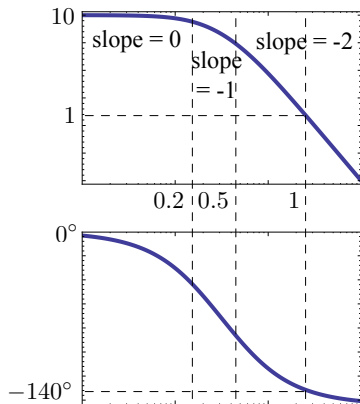


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at $\omega = 2$, $M \approx 0.24$ (with $K = 1$)

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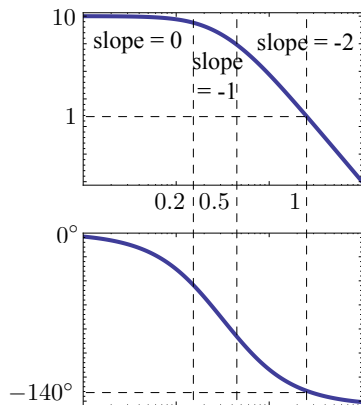
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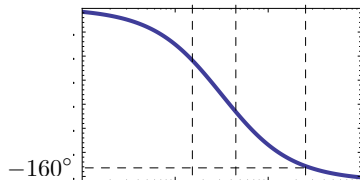
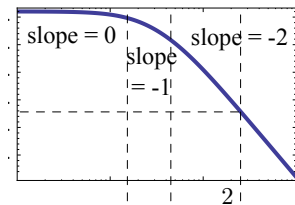
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— choose $K = 4$

(gives ω_c slightly < 2 , but still ok).

Lead & Lag Compensation

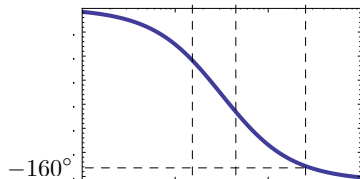
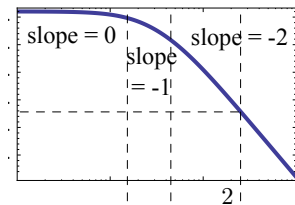
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Step 2. Decide how much phase lead is needed, and choose z_{lead} and p_{lead}

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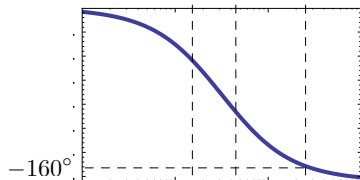
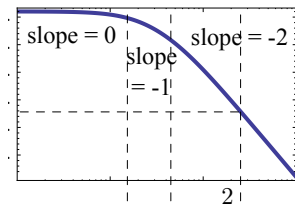


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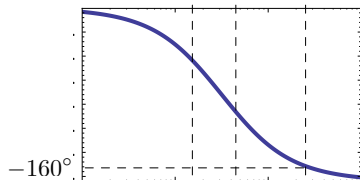
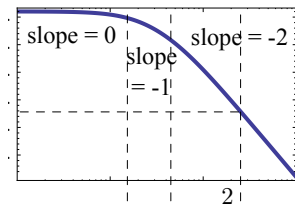
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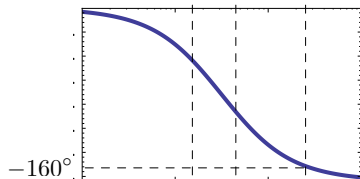
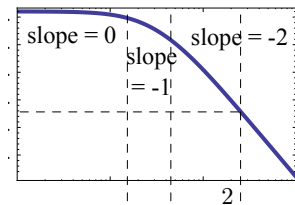
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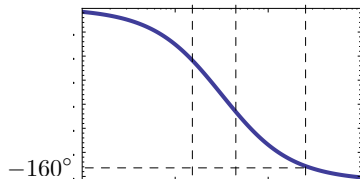
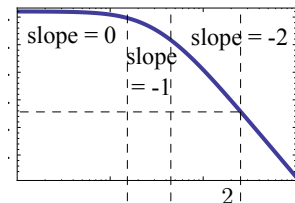
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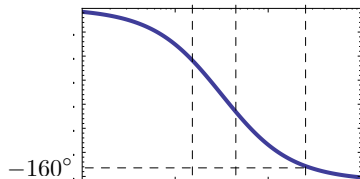
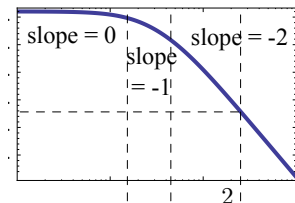
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Lead & Lag Compensation

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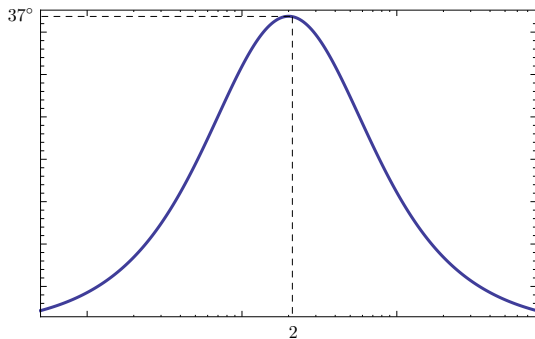
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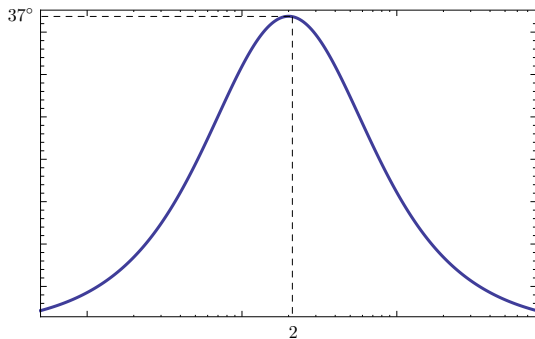


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Phase lead = 37° — not enough!!

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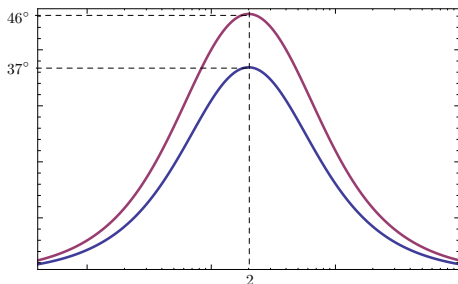
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$$\begin{cases} z_{\text{lead}} = 0.8 \\ p_{\text{lead}} = 5 \end{cases} \implies \text{phase lead} = 46^\circ$$



Lead & Lag Compensation

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We want $D(0) \geq \frac{99}{40}$ with lag $\frac{z_{\text{lag}}}{p_{\text{lag}}} \approx 2.5$ will do

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Overall controller:

$$\underbrace{4 \frac{\frac{s}{0.8} + 1}{\frac{s}{5} + 1}}_{\text{lead (with gain } K = 4 \text{ absorbed)}} \cdot \underbrace{\frac{s + 0.05}{s + 0.02}}_{\text{lag (not in Bode form)}}$$

(Note: we don't rewrite lag in Bode form, because $z_{\text{lag}}/p_{\text{lag}}$ is not incorporated into K .)

Frequency Domain Design Method: Advantages

Design based on Bode plots is good for:

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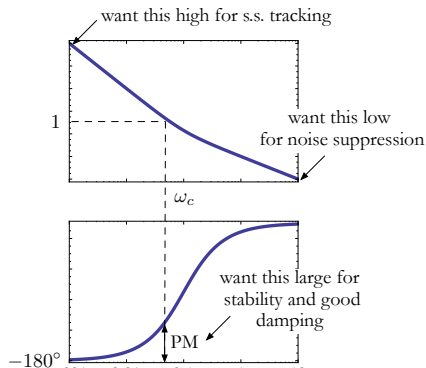
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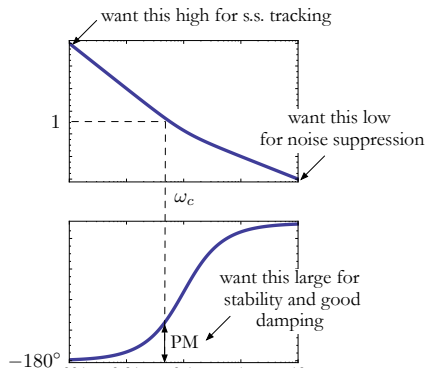
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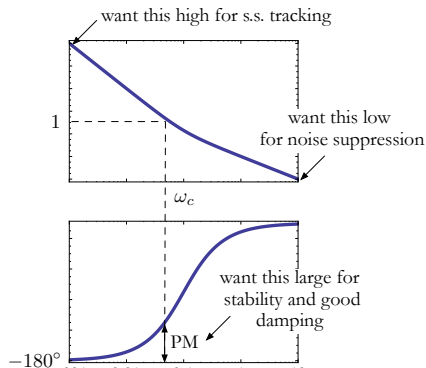


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Frequency Domain Design Method: Advantages

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- ▶ evaluating the design and seeing which way to change it
- ▶ using experimental data (frequency response of the uncontrolled system can be measured experimentally)

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What we want is a frequency-domain substitute for the Routh–Hurwitz criterion — this is the **Nyquist criterion**, which we will discuss in the next lecture.