

## Lecture 26: DTFT of a Sinusoid

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ECE 401: Signal and Image Analysis

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Magnitude-summable signals have a DTFT:

$$
X(\omega) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}, \quad \Leftrightarrow \quad x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega)e^{j\omega n} d\omega
$$

Periodic signals have a Fourier series:

$$
X_k = \frac{1}{N} \sum_{n=1}^{N-1} x[n] e^{-j\frac{2\pi kn}{N}}, \quad \Leftrightarrow \quad x[n] = \sum_{k=0}^{N-1} X_k e^{j\frac{2\pi kn}{N}}
$$

Finite-length or periodic signals have a DFT:

$$
X[k] = \sum_{n=1}^{N-1} x[n]e^{-j\frac{2\pi kn}{N}}, \quad \Leftrightarrow \quad x[n] = \frac{1}{N}\sum_{k=0}^{N-1} X[k]e^{j\frac{2\pi kn}{N}}
$$

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To find the DFT of a sinusoid, we use the frequency-shift property of the DFT:

$$
x[n] = \cos(\omega_0 n)w[n] = \left(\frac{1}{2}w[n]e^{j\omega_0 n} + \frac{1}{2}w[n]e^{-j\omega_0 n}\right)
$$
  

$$
\leftrightarrow
$$
  

$$
X[k] = \frac{1}{2}W\left(\frac{2\pi k}{N} - \omega_0\right) + \frac{1}{2}W\left(\frac{2\pi k}{N} + \omega_0\right)
$$

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where  $W(\omega)$  is the DTFT of the window.



Today's questions are:

**1** Can we use the frequency-shift property to find the DTFT of a windowed sinusoid?

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2 Can we use something like that to find the DTFT of a non-windowed, infinite length sinusoid?

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First, let's find the DTFT of a windowed sinusoid. This is easy; it's the same as the DFT. Since

$$
x[n] = \cos(\omega_0 n) w[n] = \left(\frac{1}{2}w[n]e^{j\omega_0 n} + \frac{1}{2}w[n]e^{-j\omega_0 n}\right)
$$

We can just use the frequency-shift property of the DTFT to get

$$
X(\omega)=\frac{1}{2}W(\omega-\omega_0)+\frac{1}{2}W(\omega+\omega_0)
$$

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Here are the DTFT and DFT of

$$
x[n] = \cos\left(\frac{2\pi 20.3}{N}n\right) w[n]
$$





Here are the DTFT and DFT of a cosine at a frequency that's a multiple of  $2\pi k/N$ .



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- How about  $x[n] = \cos(\omega_0 n)$ , with no windows? Does it have a DTFT?
- It's not magnitude-summable!

$$
\sum_{n=-\infty}^{\infty} |x[n]| = \infty
$$

Therefore, there's no guarantee that it has a valid DTFT.

• In fact, we will need to make up some new math in order to find the DTFT of this signal.

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The Dirac delta function,  $\delta(\omega)$ , is defined as:

- $\delta(\omega) = 0$  for all  $\omega$  other than  $\omega = 0$ .
- $\delta(0) = \infty$
- The integral of  $\delta(\omega)$ , from any negative  $\omega$  to any positive  $\omega$ , is exactly 1:

 $\int^{\epsilon}$  $-\epsilon$  $\delta(\omega) d\omega = 1$ 

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We usually draw it like this. The arrow has zero width, infinite height, and an area of exactly 1.0.





[https://commons.wikimedia.org/wiki/File:](https://commons.wikimedia.org/wiki/File:Dirac_distribution_PDF.svg)



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The key use of a Dirac delta is that, when we multiply it by any function and integrate,

- All the values of that function at  $\omega \neq 0$  are multiplied by  $\delta(\omega) = 0$
- The value at  $\omega = 0$  is multiplied by  $+\infty$ , in such a way that the integral is exactly:

$$
\int_{-\pi}^{\pi} f(\omega) \delta(\omega) d\omega = f(0)
$$

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The delta function can also be shifted, to some frequency  $\omega_0$ . This is written as  $\delta(\omega - \omega_0)$ .

- All the values of that function at  $\omega \neq \omega_0$  are multiplied by  $\delta(\omega-\omega_0)=0$
- The value at  $\omega = \omega_0$  is multiplied by  $+\infty$ , in such a way that the integral is exactly:

$$
\int_{-\pi}^{\pi} f(\omega) \delta(\omega - \omega_0) d\omega = f(\omega_0)
$$

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Thus, for example,

$$
\frac{1}{2\pi}\int_{-\pi}^{\pi}\delta(\omega-\omega_0)e^{j\omega n}d\omega=\frac{1}{2\pi}e^{j\omega_0 n}
$$

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In other words, the inverse DTFT of  $Y(\omega) = \delta(\omega - \omega_0)$  is  $y[n] = \frac{1}{2\pi} e^{j\omega_0 n}$ , a complex exponential.



By the linearity of the DTFT, we therefore have the following useful DTFT pairs:

$$
e^{j\omega_0 n} \quad \leftrightarrow \quad 2\pi \delta(\omega-\omega_0),
$$

and

$$
\cos(\omega_0 n) \quad \leftrightarrow \quad \pi \delta(\omega - \omega_0) + \pi \delta(\omega + \omega_0)
$$

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Suppose we were to try to find the DTFT of  $x[n] = e^{j\omega_0 n}$  directly:

$$
X(\omega) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} = \sum_{n=-\infty}^{\infty} e^{j(\omega - \omega_0)n}
$$

- $\bullet$  At frequencies  $\omega \neq \omega_0$ , we would be adding the samples of a sinusoid, which would give us  $X(\omega) = 0$ .
- At  $\omega = \omega_0$ , the summation becomes

$$
X(\omega_0)=\sum_{n=-\infty}^\infty 1=\infty
$$

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• So  $X(\omega_0) = \infty$ , and  $X(\omega) = 0$  everywhere else. So it's a Dirac delta! The only thing the forward transform **doesn't** tell us is: what kind of infinity?



- So  $X(\omega_0) = \infty$ , and  $X(\omega) = 0$  everywhere else. So it's a Dirac delta! The only thing the forward transform **doesn't** tell us is: what kind of infinity?
- The inverse DTFT gives us the answer. It needs to be the kind of infinity such that

$$
\frac{1}{2\pi}\int_{-\pi}^{\pi}X(\omega)e^{j\omega n}d\omega=e^{j\omega_0 n},
$$

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and the solution is  $X(\omega) = 2\pi \delta(\omega - \omega_0)$ 

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Remember that windowing in time  $=$  convolution in frequency:

$$
y[n] = x[n]w[n] \quad \leftrightarrow \quad Y(\omega) = \frac{1}{2\pi}X(\omega) * W(\omega).
$$

But if  $x[n] = \cos(\omega_0 n)$ , we already know that

$$
y[n] = \cos(\omega_0 n) w[n] \quad \leftrightarrow \quad Y(\omega) = \frac{1}{2} W(\omega - \omega_0) + \frac{1}{2} W(\omega + \omega_0)
$$

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Can we reconcile these two facts?

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The delta function is defined by this sampling property:

$$
\int_{-\pi}^{\pi} \delta(\omega - \omega_0) f(\omega) d\omega = f(\omega_0)
$$

What does that mean about convolution? Let's try it:

$$
\delta(\omega - \omega_0) * W(\omega) = \int_{-\pi}^{\pi} \delta(\theta - \omega_0) W(\omega - \theta) d\theta
$$

$$
= W(\omega - \omega_0)
$$

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So we see that:

$$
\delta(\omega-\omega_0) * W(\omega) = W(\omega-\omega_0)
$$

This is just like the behavior of impulses in the time domain:





[https://commons.wikimedia.org/wiki/File:Convolution\\_of\\_two\\_pulses\\_with\\_impulse\\_response.svg](https://commons.wikimedia.org/wiki/File:Convolution_of_two_pulses_with_impulse_response.svg)



So if:

$$
\cos(\omega_0 n) \quad \leftrightarrow \quad \pi \delta(\omega - \omega_0) + \pi \delta(\omega + \omega_0),
$$

and

$$
y[n] = x[n]w[n] \quad \leftrightarrow \quad Y(\omega) = \frac{1}{2\pi}X(\omega) * W(\omega),
$$

then

$$
\cos(\omega_0 n) w[n] \quad \leftrightarrow \quad \left(\frac{1}{2}\delta(\omega - \omega_0) * W(\omega) + \frac{1}{2}\delta(\omega + \omega_0) * W(\omega)\right)
$$

$$
= \left(\frac{1}{2}W(\omega - \omega_0) + \frac{1}{2}W(\omega + \omega_0)\right)
$$

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So again, we discover that:

$$
x[n] = \cos\left(\frac{2\pi 20.3}{N}n\right) w[n]
$$





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#### Go to the course web page, and try the quiz!

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DTFT of a complex exponential is a delta function:

$$
e^{j\omega_0 n} \leftrightarrow 2\pi \delta(\omega-\omega_0)
$$

DTFT of a cosine is two delta functions:

$$
\cos(\omega_0 n) \quad \leftrightarrow \quad \pi \delta(\omega - \omega_0) + \pi \delta(\omega + \omega_0)
$$

DTFT of a windowed cosine is frequency-shifted window functions:

$$
\cos(\omega_0 n) w[n] \quad \leftrightarrow \quad \frac{1}{2} W(\omega - \omega_0) + \frac{1}{2} W(\omega + \omega_0)
$$

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Consider the function

$$
x[n] = A\cos(\omega_0 n + \theta)
$$

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What is  $X(\omega)$ ? How about  $y[n] = w[n]x[n]$ . What is  $Y(\omega)$ ?