# ECE 401 Signal and Image Analysis Homework 3

# UNIVERSITY OF ILLINOIS

Department of Electrical and Computer Engineering

Assigned: 9/25/2023; Due: 10/4/2023Reading:  $DSP\ First\ Chapter\ 5$ 

#### Problem 3.1

In MP3, one of the filters you'll create is a local averaging filter. A local averaging filter produces an output y[n], at time n, which is the average of the previous N samples of x[m]:

$$y[n] = \sum_{m=-\infty}^{\infty} h[m]x[n-m]$$
(3.1-1)

$$h[m] = \begin{cases} \frac{1}{N} & 0 \le m \le N - 1\\ 0 & \text{otherwise} \end{cases}$$
 (3.1-2)

(a) First, consider what happens if x[m] is a pure tone with a period of  $N_0 = \frac{2\pi}{\omega_0}$ , an amplitude of A, and a phase of  $\theta$ :

$$x[n] = A\cos(\omega_0 n - \theta)$$

Suppose that the averaging window, N, is exactly an integer multiple of  $N_0$ . For example, suppose that  $N = 3N_0$ . Draw a picture of x[n] as a function of n, and shade in the regions that would be added together by the summation in Eq. (3.1-1) in order to compute y[0]. Argue based on your figure (with no calculations at all) that y[0] = 0.

**Solution:** The picture should show that we are adding together three complete periods of the cosine up through, and including, the sample x[0]. Every period of the cosine has a positive section and a negative section. When we average these two sections, they cancel each other out.

(b) Adding up the samples of a cosine is easy when N is an integer multiple of  $N_0$ , but hard otherwise. It's actually much easier to add the samples of a complex exponential, because we can use the standard geometric series formula (https://en.wikipedia.org/wiki/Geometric\_series#Formula). Use that formula to find y[N-1] when

$$x[n] = e^{j\omega_0 n}$$

Your result should have the form y[0] = (1-a)/(1-b) for some complex-valued constants, a and b, that depend on  $\pi$ , N, and  $\omega_0$ , but not on m or n.

## Solution:

$$y[0] = \frac{1}{N} \left( \frac{1 - e^{-j\omega_0 N}}{1 - e^{-j\omega_0}} \right)$$

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#### Problem 3.2

Another of the filters you'll create in MP3 is a backward-difference filter. A backward-difference filter is one of several different ways of approximating a first-derivative:

$$y[n] = \sum_{m=-\infty}^{\infty} h[m]x[n-m]$$
 (3.2-1)

$$h[m] = \begin{cases} 1 & m = 0 \\ -1 & m = 1 \\ 0 & \text{otherwise} \end{cases}$$
 (3.2-2)

(a) First, consider what happens if x[m] is a pure tone with a period of  $T_0 = \frac{2\pi}{\omega_0}$ , an amplitude of A, and a phase of  $\theta$ :

$$x[n] = A\cos(\omega_0 n - \theta) \tag{3.2-3}$$

Plug Eq. (3.2-3) into Eq. (3.2-1), then use the following trigonometric identity and the following approximation to discover in exactly what way the first-difference operator approximates a derivative:

$$-2\sin(a)\sin(b) = \cos(a+b) - \cos(a-b)$$
$$\sin(b) \approx b \text{ if } b \text{ is small}$$

In order to apply the approximation, you can assume that  $\omega_0$  is a small number.

Solution: Plugging Eq. (3.2-3) into Eq. (3.2-1) gives

$$y[n] = x[n] - x[n-1]$$
  
=  $A\cos(\omega_0 n - \theta) - A\cos(\omega_0 (n-1) - \theta)$ 

We can apply the trig identity if we set

$$a = \omega_0(n - 0.5) - \theta$$
$$b = 0.5\omega_0$$

which gives

$$y[n] = -2A\sin(0.5\omega_0)\sin(\omega_0(n-0.5) + \theta)$$

Applying the approximation gives

$$y[n] \approx -A\omega_0 \sin(\omega_0(n-0.5) + \theta),$$

which is what you would get if you differentiated  $A\cos(\omega_0(n-0.5)+\theta)$  with respect to n.

(b) Let's try the same thing with a complex exponential. Plug the following value of x[n] into Eq. (3.2-1)

$$x[n] = e^{j\omega_0 n},$$

then assume that  $\omega_0$  is a small number, and simplify using the approximation

$$e^{\phi} \approx 1 + \phi$$
 if  $\phi$  is small

in order to get something that looks like dx[n]/dn.

### Solution:

$$y[n] = x[n] - n[n-1]$$

$$= e^{j\omega_0 n} - e^{j\omega_0(n-1)}$$

$$= e^{j\omega_0 n} (1 - e^{-j\omega_0})$$

$$\approx j\omega_0 e^{-j\omega_0 n}$$

which is dx[n]/dn.