

Lecture 8: Sampling Theorem

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ECE 401: Signal and Image Analysis, Fall 2021

- 1 Review: Sampling
- 2 Spectrum Plots
- 3 Spectrum of Oversampled Signals
- 4 Spectrum of Undersampled Signals
- 5 The Sampling Theorem
- 6 Summary
- 7 Written Example

Outline

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How to sample a continuous-time signal

Suppose you have some continuous-time signal, $x(t)$, and you'd like to sample it, in order to store the sample values in a computer. The samples are collected once every $T_s = \frac{1}{F_s}$ seconds:

$$x[n] = x(t = nT_s)$$

Aliasing

- A sampled sinusoid can be reconstructed perfectly if the Nyquist criterion is met, $f < \frac{F_s}{2}$.
- If the Nyquist criterion is violated, then:
 - If $\frac{F_s}{2} < f < F_s$, then it will be aliased to

$$f_a = F_s - f$$

$$z_a = z^*$$

i.e., the sign of all sines will be reversed.

- If $F_s < f < \frac{3F_s}{2}$, then it will be aliased to

$$f_a = f - F_s$$

$$z_a = z$$

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Spectrum Plots

The **spectrum plot** of a periodic signal is a plot with

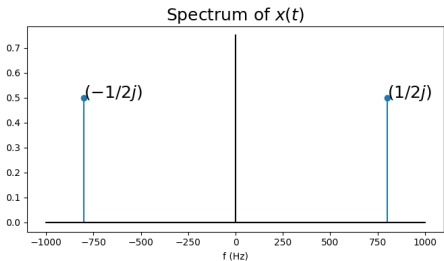
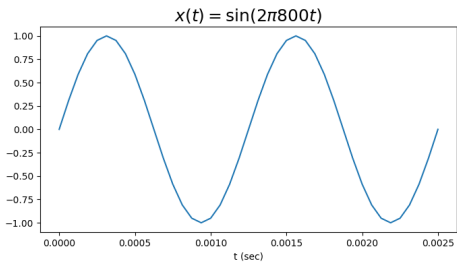
- frequency on the X-axis,
- showing a vertical spike at each frequency component,
- each of which is labeled with the corresponding phasor.

Example: Sine Wave

$$\begin{aligned}x(t) &= \sin(2\pi 800t) \\ &= \frac{1}{2j}e^{j2\pi 800t} - \frac{1}{2j}e^{-j2\pi 800t}\end{aligned}$$

The spectrum of $x(t)$ is $\{(-800, -\frac{1}{2j}), (800, \frac{1}{2j})\}$.

Example: Sine Wave

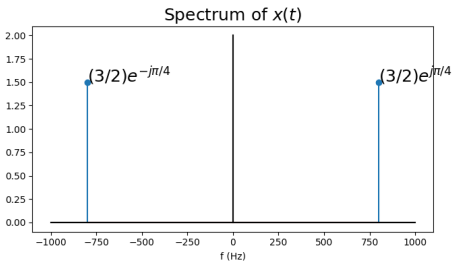
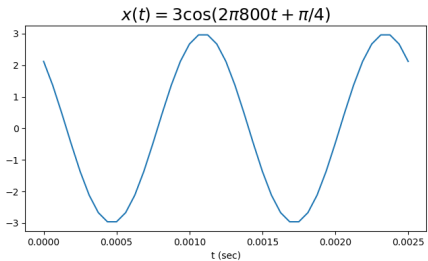


Example: Quadrature Cosine

$$\begin{aligned}x(t) &= 3 \cos\left(2\pi 800t + \frac{\pi}{4}\right) \\ &= \frac{3}{2} e^{j\pi/4} e^{j2\pi 800t} + \frac{3}{2} e^{-j\pi/4} e^{-j2\pi 800t}\end{aligned}$$

The spectrum of $x(t)$ is $\left\{(-800, \frac{3}{2} e^{-j\pi/4}), (800, \frac{3}{2} e^{j\pi/4})\right\}$.

Example: Quadrature Cosine



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Oversampled Signals

A signal is called **oversampled** if $F_s > 2f$ (e.g., so that sinc interpolation can reconstruct it from its samples).

Spectrum Plot of a Discrete-Time Periodic Signal

The spectrum plot of a **discrete-time periodic signal** is a regular spectrum plot, but with the X-axis relabeled. Instead of frequency in Hertz = $\left[\frac{\text{cycles}}{\text{second}} \right]$, we use

$$\omega \left[\frac{\text{radians}}{\text{sample}} \right] = \frac{2\pi \left[\frac{\text{radians}}{\text{cycle}} \right] f \left[\frac{\text{cycles}}{\text{second}} \right]}{F_s \left[\frac{\text{samples}}{\text{second}} \right]}$$

How do we plot the aliasing?

Remember that a discrete-time signal has energy at

- f and $-f$, but also $F_s - f$ and $-F_s + f$, and $F_s + f$ and $-F_s - f$, and...
- ω and $-\omega$, but also $2\pi - \omega$ and $-2\pi + \omega$, and $2\pi + \omega$ and $-2\pi - \omega$, and...

Which ones should we plot? Answer: **plot all of them!** Usually we plot a few nearest the center, then add “...” at either end, to show that the plot continues forever.

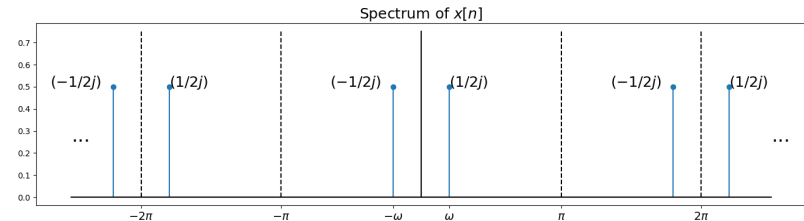
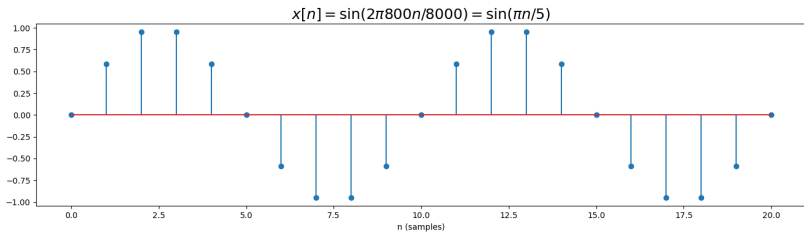
Example: Sine Wave

Let's sample at $F_s = 8000$ samples/second.

$$\begin{aligned}x[n] &= \sin(2\pi 800n/8000) \\ &= \sin(\pi n/5) \\ &= \frac{1}{2j}e^{j\pi n/5} - \frac{1}{2j}e^{-j\pi n/5}\end{aligned}$$

The spectrum of $x[n]$ is $\{\dots, (-\pi/5, -\frac{1}{2j}), (\pi/5, \frac{1}{2j}), \dots\}$.

Example: Sine Wave

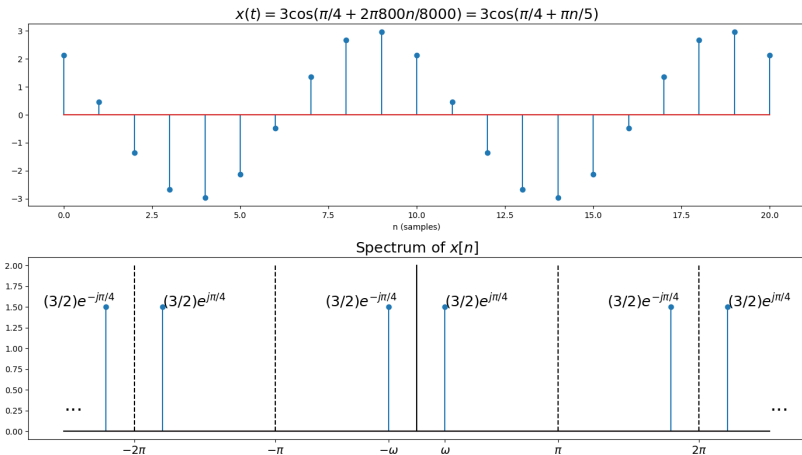


Example: Quadrature Cosine

$$\begin{aligned}x[n] &= 3 \cos \left(2\pi 800n/8000 + \frac{\pi}{4} \right) \\ &= 3 \cos \left(\pi n/5 + \frac{\pi}{4} \right) \\ &= \frac{3}{2} e^{j\pi/4} e^{j\pi n/5} + \frac{3}{2} e^{-j\pi/4} e^{-j\pi n/5}\end{aligned}$$

The spectrum of $x[n]$ is $\{\dots, (-\pi/5, \frac{3}{2}e^{-j\pi/4}), (\pi/5, \frac{3}{2}e^{j\pi/4}), \dots\}$.

Example: Quadrature Cosine



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Undersampled Signals

A signal is called **undersampled** if $F_s < 2f$ (e.g., so that sinc interpolation can't reconstruct it from its samples).

... but Aliasing?

Remember that a discrete-time signal has energy at

- f and $-f$, but also $F_s - f$ and $-F_s + f$, and $F_s + f$ and $-F_s - f$, and...
- ω and $-\omega$, but also $2\pi - \omega$ and $-2\pi + \omega$, and $2\pi + \omega$ and $-2\pi - \omega$, and...

We still want to plot all of these, but now ω and $-\omega$ won't be the spikes closest to the center. Instead, some other spike will be closest to the center.

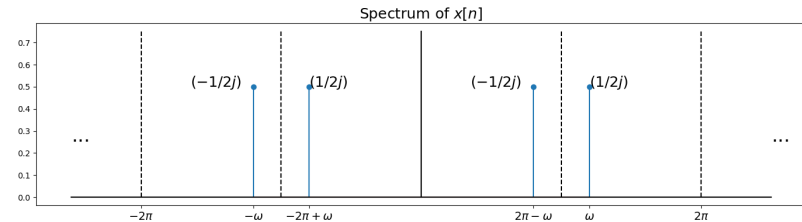
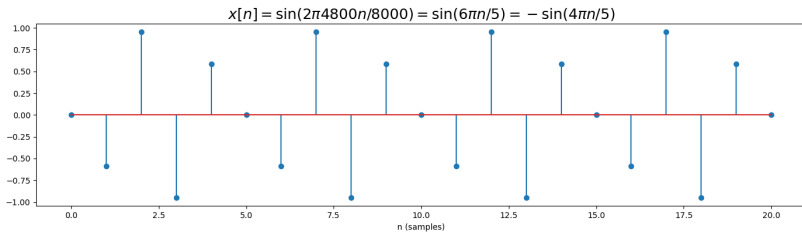
Example: Sine Wave

Let's still sample at $F_s = 8000$, but we'll use a sine wave at $f = 4800\text{Hz}$, so it gets undersampled.

$$\begin{aligned}x[n] &= \sin(2\pi 4800n/8000) \\ &= \sin(6\pi n/5) \\ &= -\sin(4\pi n/5) \\ &= -\frac{1}{2j}e^{j4\pi n/5} + \frac{1}{2j}e^{j4\pi n/5}\end{aligned}$$

The spectrum of $x[n]$ is $\{\dots, (-4\pi/5, \frac{1}{2j}), (4\pi/5, -\frac{1}{2j}), \dots\}$.

Example: Sine Wave



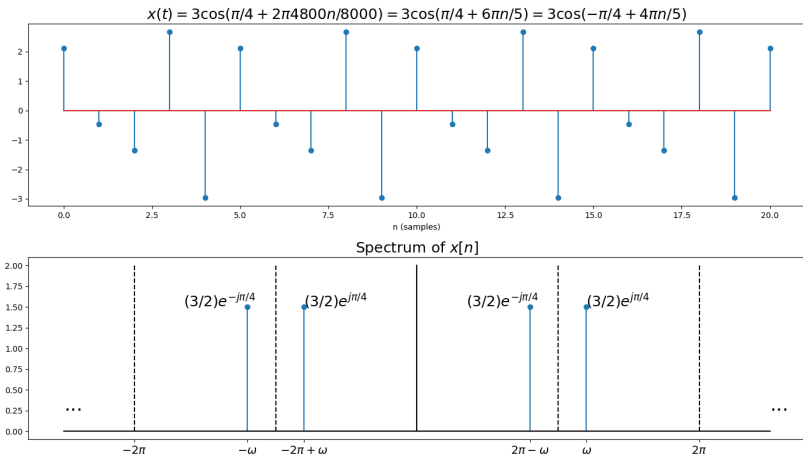
Example: Quadrature Cosine

$$\begin{aligned}x[n] &= 3 \cos \left(2\pi 4800n/8000 + \frac{\pi}{4} \right) \\&= 3 \cos \left(6\pi n/5 + \frac{\pi}{4} \right) \\&= 3 \cos \left(4\pi n/5 - \frac{\pi}{4} \right) \\&= \frac{3}{2} e^{-j\pi/4} e^{j4\pi n/5} + \frac{3}{2} e^{j\pi/4} e^{-j4\pi n/5}\end{aligned}$$

The spectrum of $x[n]$ is

$$\left\{ \dots, (-4\pi/5, \frac{3}{2} e^{j\pi/4}), (4\pi/5, \frac{3}{2} e^{-j\pi/4}), \dots \right\}.$$

Example: Quadrature Cosine



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General periodic continuous-time signals

Let's assume that $x(t)$ is periodic with some period T_0 , therefore it has a Fourier series:

$$x(t) = \sum_{k=-\infty}^{\infty} X_k e^{j2\pi kt/T_0} = \sum_{k=0}^{\infty} |X_k| \cos\left(\frac{2\pi kt}{T_0} + \angle X_k\right)$$

Eliminate the aliased tones

We already know that $e^{j2\pi kt/T_0}$ will be aliased if $|k|/T_0 > F_N$. So let's assume that the signal is **band-limited**: it contains no frequency components with frequencies larger than $F_S/2$.

That means that the only X_k with nonzero energy are the ones in the range $-N/2 \leq k \leq N/2$, where $N \leq F_S T_0$.

$$x(t) = \sum_{k=-N/2}^{N/2} X_k e^{j2\pi kt/T_0} = \sum_{k=0}^{N/2} |X_k| \cos\left(\frac{2\pi kt}{T_0} + \angle X_k\right)$$

Sample that signal!

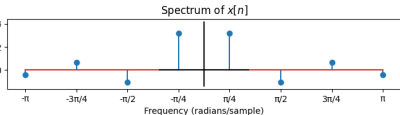
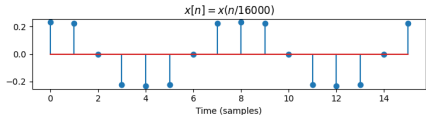
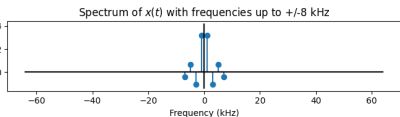
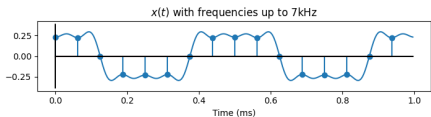
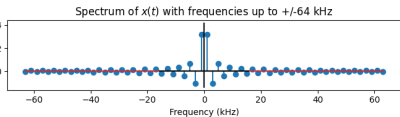
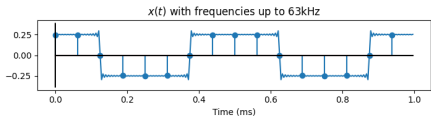
Now let's sample that signal, at sampling frequency F_S :

$$x[n] = \sum_{k=-N/2}^{N/2} X_k e^{j2\pi kn/F_S T_0} = \sum_{k=0}^{N/2} |X_k| \cos\left(\frac{2\pi kn}{N} + \angle X_k\right)$$

So the highest digital frequency, when $k = F_S T_0/2$, is $\omega_k = \pi$.
The lowest is $\omega_0 = 0$.

$$x[n] = \sum_{\omega_k=-\pi}^{\pi} X_k e^{j\omega_k n} = \sum_{\omega_k=0}^{\pi} |X_k| \cos(\omega_k n + \angle X_k)$$

Spectrum of a sampled periodic signal



The sampling theorem

As long as $-\pi \leq \omega_k \leq \pi$, we can recreate the continuous-time signal by either (1) using sinc interpolation, or (2) regenerating a continuous-time signal with the corresponding frequency:

$$f_k \left[\frac{\text{cycles}}{\text{second}} \right] = \frac{\omega_k \left[\frac{\text{radians}}{\text{sample}} \right] \times F_S \left[\frac{\text{samples}}{\text{second}} \right]}{2\pi \left[\frac{\text{radians}}{\text{cycle}} \right]}$$

$$x[n] = \cos(\omega_k n + \theta_k) \rightarrow x(t) = \cos(2\pi f_k t + \theta_k)$$

The sampling theorem

A continuous-time signal $x(t)$ with frequencies no higher than f_{max} can be reconstructed exactly from its samples $x[n] = x(nT_S)$ if the samples are taken at a rate $F_S = 1/T_S$ that is $F_S \geq 2f_{max}$.

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The sampling theorem

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Written Example

Let $x(t)$ be a sinusoid with some amplitude, some phase, and some frequency.

- Plot the spectrum of $x(t)$.
- Choose an F_s that undersamples it. Plot the spectrum of $x[n]$.