21 Monochromatic waves and phasor notation

• Recall that we reached the traveling-wave d'Alembert solutions

$$\mathbf{E},\,\mathbf{H}\propto f(t\mp\frac{z}{v})$$

via the superposition of time-shifted and amplitude-scaled versions of

$$f(t) = \cos(\omega t),$$

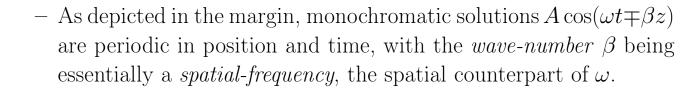
namely the monochromatic waves

$$A\cos[\omega(t\mp\frac{z}{v})] = A\cos(\omega t\mp\beta z),$$

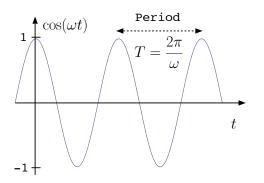
with amplitudes A where

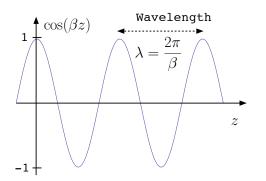
$$\beta \equiv \frac{\omega}{v} = \omega \sqrt{\mu \epsilon}$$

can be called **wave-number** in analogy with **wave-frequency** ω .



This is an important point that you should try to understand well — it has implications for signal processing courses related to images and vision.





- In general, monochromatic solutions of 1D wave-equations obtained in various branches of science and engineering can all be represented in the same format as above in terms of wave-frequency / wave-wavenumber pairs ω and β having a ratio

$$v \equiv \frac{\omega}{\beta}$$

recognized as the **wave-speed** and specific **dispersion relations** such as:

1. **TEM waves** in perfect dielectrics:

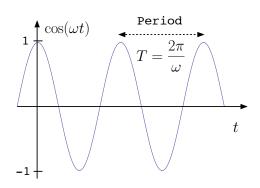
$$\beta = \omega \sqrt{\mu \epsilon},$$

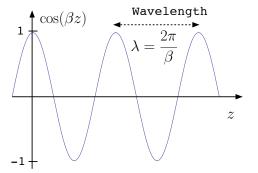
2. Acoustic waves in monoatomic gases with temperature T (K) and atomic mass m (kg):

$$\beta = \omega \sqrt{\frac{m}{\frac{5}{3}KT}},$$

3. TEM waves in collisionless **plasmas** (ionized gases) with plasma frequency $\omega_p = \sqrt{\frac{Ne^2}{m\epsilon_o}}$:

$$\beta = \frac{1}{c} \sqrt{\omega^2 - \omega_p^2}.$$





Dispersion relations between $wave frequency \omega$ and $wave number \beta$ determine the propagation velocity

$$v = \frac{\omega}{\beta} = \lambda f$$

for all types of wave motions.

For any type of wave solution — TEM, acoustic, plasma wave
 — once the dispersion relation is available (meaning that it has been derived from fundamental physical laws governing the specific wave type), wave propagation velocity is always obtained as

$$v = \frac{\omega}{\beta}$$

or, equivalently, as

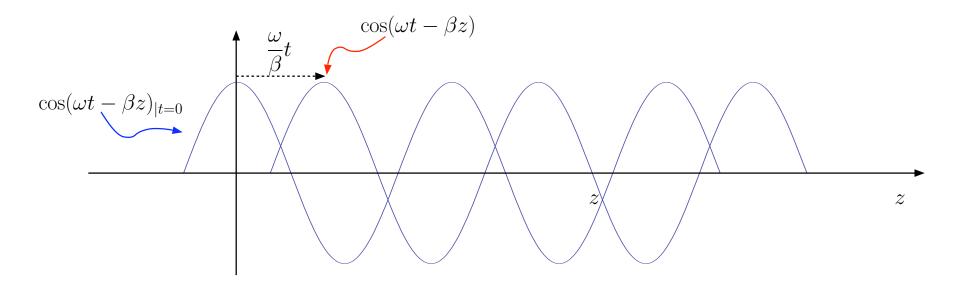
$$v = \frac{\lambda}{T} = \lambda f$$

where

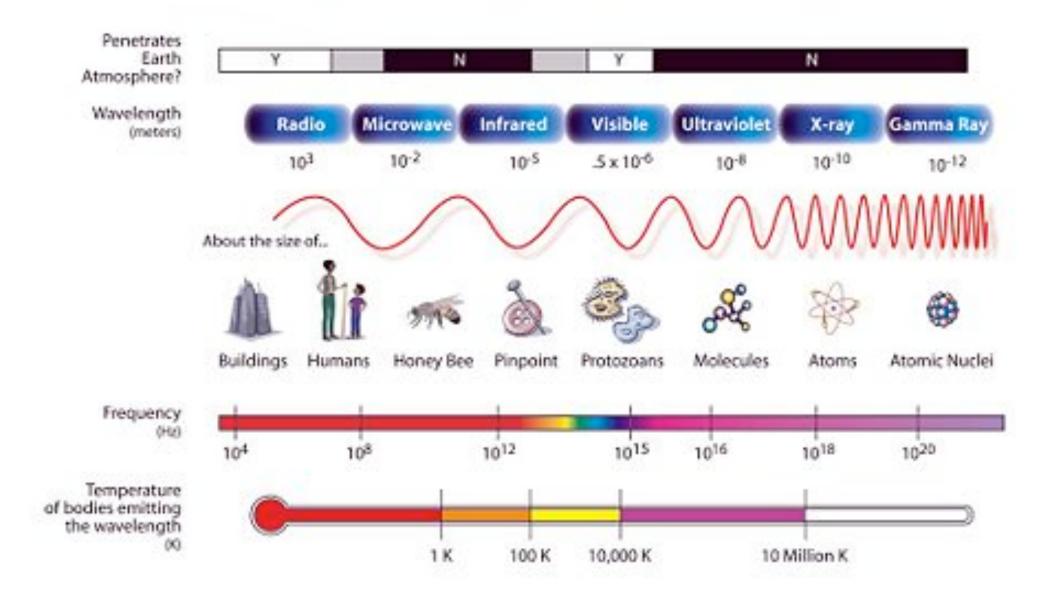
$$\lambda \equiv \frac{2\pi}{\beta}$$
 Wavelength

and

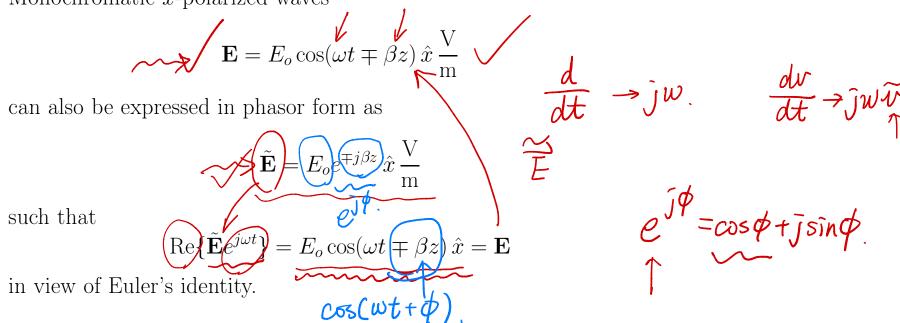
$$T = \frac{2\pi}{\omega} \equiv \frac{1}{f}$$
 Waveperiod.



THE ELECTROMAGNETIC SPECTRUM



 \bullet Monochromatic x-polarized waves



Example 1: Study the following table to understand monochromatic wave

fields and their phasors.

1	Field	Phasor	Comment
~>	$\mathbf{E} = \cos(\omega t + \beta y) \hat{z}$	$\tilde{\mathbf{E}} = e^{i\beta y} \hat{z}$	z-polarized wave propagating in $-y$ direction
		$ ilde{\mathbf{H}} = -rac{e^{jeta y}}{m} \hat{x}$	magnetic phasor that accompanies $\tilde{\mathbf{E}}$ above
~>	$\mathbf{H} = \sin(\omega t - \beta z) \hat{y}$	$\tilde{\mathbf{H}} = (-j)e^{-j\beta z}\hat{y}$	wave propagating in $+z$ direction
		$\tilde{\mathbf{E}} = -j \hat{\mathbf{p}} e^{-j\beta z} \hat{x}$	electric field phasor of H above
	$\mathbf{E} = \eta \sin(\omega t - \beta z) \hat{x}$	3	which is an x-polarized field (see the right column)
,		11	

Example 2: Given that

$$\mathbf{H} = \hat{x}H^{+}\cos(\omega t - \beta z) + \hat{y}H^{-}\sin(\omega t + \beta z)$$

representing the sum of wave fields propagating in opposite directions, the corresponding phasor

$$\tilde{\mathbf{H}} = \hat{x}H^+e^{-j\beta z} - j\hat{y}H^-e^{j\beta z}.$$

The corresponding **E**-field phasor is

$$\tilde{\mathbf{E}} = -\hat{y}\eta H^+ e^{-j\beta z} + j\hat{x}\eta H^- e^{j\beta z},$$

from which

$$\mathbf{E} = -\hat{y}\eta H^{+}\cos(\omega t - \beta z) - \hat{x}\eta H^{-}\sin(\omega t + \beta z).$$

Make sure to check that all the signs make sense, and if you think you have caught an error, let us know.

- In general, we transform between plane TEM wave phasors $\tilde{\mathbf{E}}$ and $\tilde{\mathbf{H}}$ as follows:
- 1. To obtain $\tilde{\mathbf{H}}$ from $\tilde{\mathbf{E}}$: divide $\tilde{\mathbf{E}}$ by η and rotate the vector direction so that vector $\tilde{\mathbf{S}} \equiv \tilde{\mathbf{E}} \times \tilde{\mathbf{H}}^*$ points in the propagation direction of the wave more on complex vector $\tilde{\mathbf{S}}$ later on.
- 2. To obtain $\tilde{\mathbf{E}}$ from $\tilde{\mathbf{H}}$: multiply $\tilde{\mathbf{H}}$ by η and rotate the vector direction so that vector $\tilde{\mathbf{S}} \equiv \tilde{\mathbf{E}} \times \tilde{\mathbf{H}}^*$ points in the propagation direction of the

$$\frac{1}{6} |\vec{E}| = \eta |\vec{H}|$$

wave.

Example 3: On z = 0 plane we have a monochromatic surface current specified as

$$\underbrace{\mathbf{J}_{s}} = \hat{x}f(t) = \underbrace{\hat{x}2\cos(\omega t)\frac{\mathbf{A}}{\mathbf{m}}} = \underbrace{\operatorname{Re}\{\hat{x}2e^{j\omega t}\}}. \quad \underbrace{\overset{\sim}{\mathsf{J}_{s}}} = 2\overset{\wedge}{\mathsf{\chi}} \overset{\mathsf{A}}{\mathsf{M}}$$

Determine wave field phasors $\tilde{\mathbf{E}}^{\pm}$ and $\tilde{\mathbf{H}}^{\pm}$ for plane TEM waves propagating away from the z=0 surface on both sides (assumed vacuum).

Solution: We know that an x-polarized surface current f(t) produces

$$\sqrt{E_x} = -\frac{\eta}{2}f(t\mp\frac{z}{v})$$
 and $H_y = \mp\frac{1}{2}f(t\mp\frac{z}{v})$

in surrounding regions. Given that $f(t) = 2\cos(\omega t)$, this implies

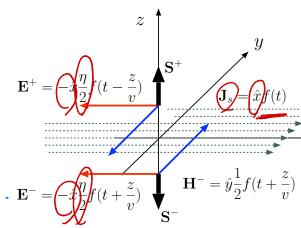
$$\int E_x = -\eta \cos(\omega t \mp \beta z) \text{ and } H_y = \underbrace{\mp \cos(\omega t \mp \beta z)}$$

where

$$\beta = \frac{\omega}{c}$$
 and $\eta = \eta_o \approx 120\pi \,\Omega$

since the current sheet is surrounded by vacuum. Converting these into phasors, we find

$$\tilde{\mathbf{E}}^{\pm} = -\eta e^{\mp j\beta} \hat{\hat{x}} \text{ and } \tilde{\mathbf{H}}^{\pm} = \mp e^{\mp j\beta z} \hat{y}.$$



• In the last lecture we calculated the time-average $\mathbf{E} \times \mathbf{H}$ and $\mathbf{J}_s \cdot \mathbf{E}$ of the fields examined in Example 3 using a time-domain approach. The [005(wt)same calculations can be carried out in terms of phasors $\tilde{\mathbf{E}}$, $\tilde{\mathbf{H}}$, and $\tilde{\mathbf{J}}_s$ as follows:

ift)
$$\langle \mathbf{E} \times \mathbf{H} \rangle = \frac{1}{2} \operatorname{Re} \{ \tilde{\mathbf{E}} \times \tilde{\mathbf{H}}^* \}$$
 and $\langle \mathbf{J}_s \cdot \mathbf{E} \rangle = \frac{1}{2} \operatorname{Re} \{ \tilde{\mathbf{J}}_s \cdot \tilde{\mathbf{E}}^* \}$

where $\tilde{\mathbf{E}} \times \tilde{\mathbf{H}}^* \equiv \tilde{\mathbf{S}}$ is called complex Poynting vector.

The proof of these are analogous to the proof of
$$\sqrt[3]{V} = \sqrt[4]{p(t)} = 5$$
 A. $\sqrt{\langle p(t) \rangle} = \frac{1}{2} \text{Re}\{VI^*\}$

for the average power of a circuit component in terms of voltage and current hasors V and I (see margin).

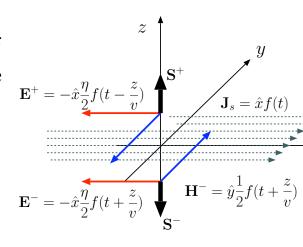
r, instance, given that

$$\tilde{\mathbf{J}}_s = 2\hat{x} \frac{\mathbf{A}}{\mathbf{m}}$$
 and $\tilde{\mathbf{E}}^{\pm}(z) = -\eta e^{\mp j\beta z} \hat{x} \frac{\mathbf{V}}{\mathbf{m}}$

in Example 3, it follows that

$$\langle -\mathbf{J}_s(t) \cdot \mathbf{E}(0,t) \rangle = \frac{1}{2} \operatorname{Re} \{ -\tilde{\mathbf{J}}_s \cdot \tilde{\mathbf{E}}^*(0) \} = \eta \approx 120\pi \frac{W}{m^2},$$

in conformity with the result from last lecture.



Instantaneous power

$$p(t) = v(t)i(t)$$

with time-harmonic signals is

$$v(t)i(t) = (\underbrace{Ve^{j\omega t} + cc}_{2})(\underbrace{\frac{Ie^{j\omega t} + cc}_{2}}_{2})$$
Ref Ve just 3.

where V and I are phasors of v(t) and i(t) and cc indicates the conjugate of the term to the left of + sign. This can be expanded as

$$v(t)i(t) = \frac{VI^* + cc}{4} + \frac{VIe^{j2\omega t} + cc}{4}.$$

The second term has a zero time average. It follows that time-average power

$$\langle v(t)i(t)\rangle = \frac{VI^* + cc}{4} = \frac{1}{2}\text{Re}\{VI^*\}$$

since

$$VI^* + cc = VI^* + V^*I = 2\text{Re}\{VI^*\}.$$

(Also see ECE 210 text.)