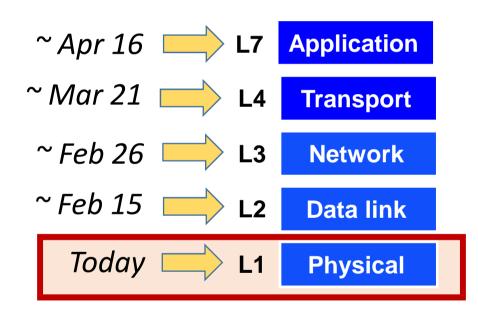
Physical Layer

CS 438: Spring 2014 Instructor: Matthew Caesar http://courses.engr.illinois.edu/cs438/

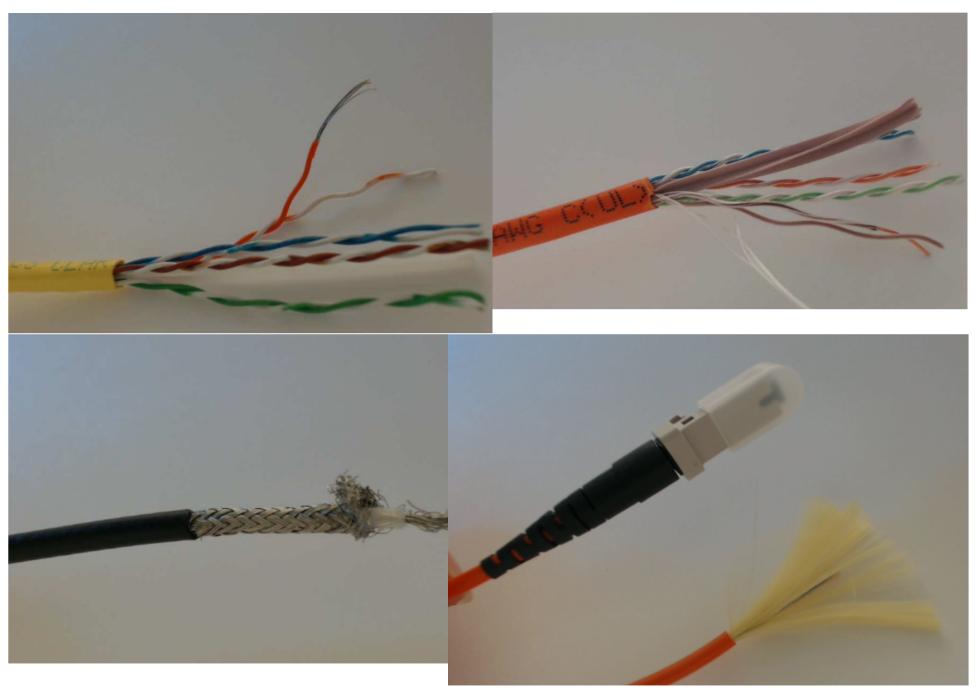
Course Outline



Outline for Today

- Today: The Physical Layer
- How to encode data over a link
- How to detect and correct errors

A Brief Overview of Physical Media



Links - Copper

- Copper-based Media
 - Category 3 Twisted Pair
 - Category 5 Twisted Pair
 - ThinNet Coaxial Cable
 - ThickNet Coaxial Cable

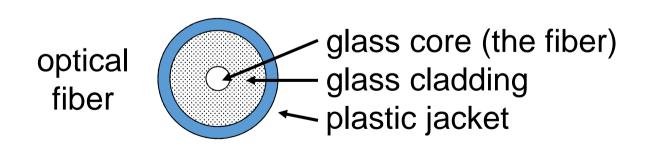
more twists, less crosstalk, better signal over longer distances 10-100Mbps 100m 10-100Mbps 200m 10-100Mbps 500m



coaxial cable (coax) copper core insulation braided outer conductor outer insulation

Links - Optical

- Optical Media
 - Multimode Fiber 100Mbps
 - Single Mode Fiber 100-2400Mbps 40km



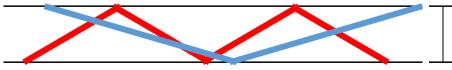
2km

Links - Optical

- Single mode fiber
 - Expensive to drive (Lasers)
 - Lower attenuation (longer distances) ≤ 0.5 dB/km
 - Lower dispersion (higher data rates)
- Multimode fiber
 - Cheap to drive (LED's)
 - Higher attenuation
 - Easier to terminate



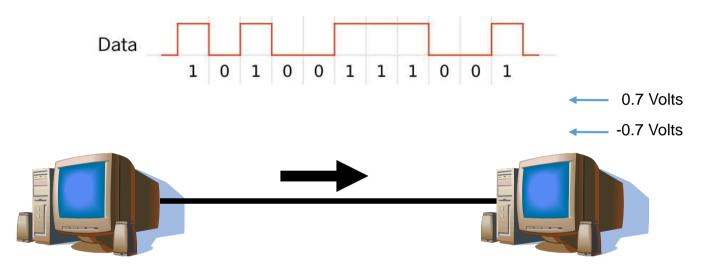
core of multimode fiber (same frequency; colors for clarity)



O(100 microns) thick

Encoding

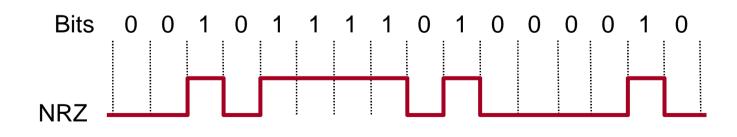
How can two hosts communicate?



- Encode data as variations in electrical/light/EM
 - Phase, frequency, and signal strength modulation, and combinations thereof
 - Simple scheme: voltage encoding
 - Encode 1's and 0's as variations in voltage
 - How to do that?

Non-Return to Zero (NRZ)

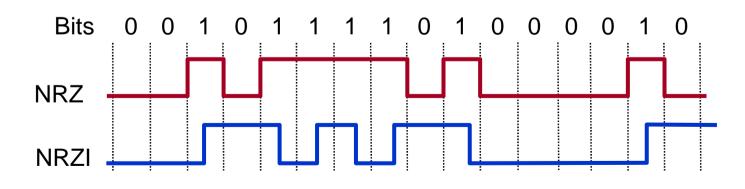
- Signal to Data
 - High ⇒ 1
 - Low ⇒ 0
- Comments
 - Transitions maintain clock synchronization
 - Long strings of 0s confused with no signal
 - Long strings of 1s causes baseline wander
 - Both inhibit clock recovery



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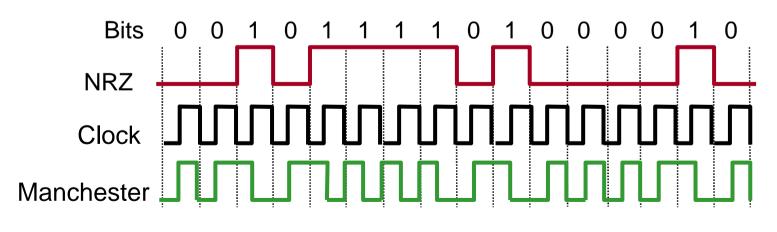
Non-Return to Zero Inverted (NRZI)

- Signal to Data
 - Transition ⇒ 1
 - Maintain ⇒ 0
- Comments
 - Solves series of 1s, but not 0s



Manchester Encoding

- Signal to Data
 - XOR NRZ data with clock
 - High to low transition \Rightarrow 1
 - Low to high transition \Rightarrow 0
- Comments
 - Used by old 10Mbps Ethernet
 - Solves clock recovery problem
 - Only 50% efficient (½ bit per transition)



4B/5B

- Signal to Data
 - Encode every 4 consecutive bits as a 5 bit symbol
- Symbols
 - At most 1 leading 0
 - At most 2 trailing 0s
 - Never more than 3 consecutive 0s
 - Transmit with NRZI
- Comments
 - 16 of 32 possible codes used for data
 - At least two transitions for each code
 - 80% efficient
 - Used by old 100Mbps Ethernet
 - Variation (64B/66B) used by modern 10Gbps Ethernet

4B/5B – Data Symbols

- $0000 \Rightarrow 11110$
- $0001 \Rightarrow 01001$
- $0010 \Rightarrow 10100$
- $0011 \Rightarrow 10101$
- $0100 \Rightarrow 01010$
- $0101 \Rightarrow 01011$
- $0110 \Rightarrow 0110$
- $0111 \Rightarrow 01111$

At most 2 trailing 0s

- $1000 \Rightarrow 10010$
- $1001 \Rightarrow 10011$
- $1010 \Rightarrow 10110$
- $1011 \Rightarrow 10111$
- $1100 \Rightarrow 11010$
- $1101 \Rightarrow 11011$
- $1110 \Rightarrow 11100$
- $1111 \Rightarrow 11101$

4B/5B – Control Symbols

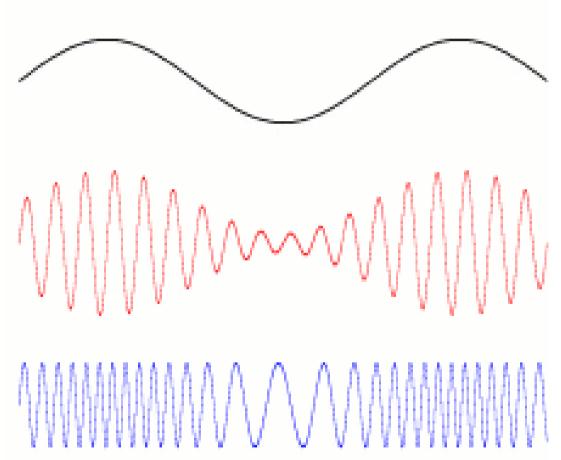
- 11111 \Rightarrow idle
- $11000 \Rightarrow$ start of stream 1
- $10001 \Rightarrow$ start of stream 2
- $01101 \Rightarrow$ end of stream 1
- $00111 \Rightarrow$ end of stream 2
- $00100 \Rightarrow$ transmit error
- Other \Rightarrow invalid

Binary Voltage Encodings

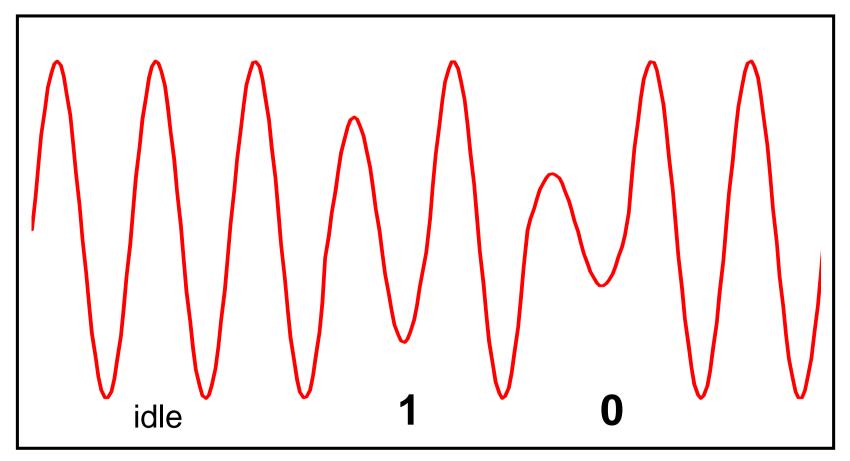
- Problem with binary voltage (square wave) encodings
 - Wide frequency range required, implying
 - Significant dispersion
 - Uneven attenuation
 - Prefer to use narrow frequency band (carrier frequency)
- Types of modulation
 - Amplitude (AM)
 - Frequency (FM)
 - Phase/phase shift
 - Combinations of these
 - Used in wireless Ethernet, optical communications

Example: AM/FM for continuous signal

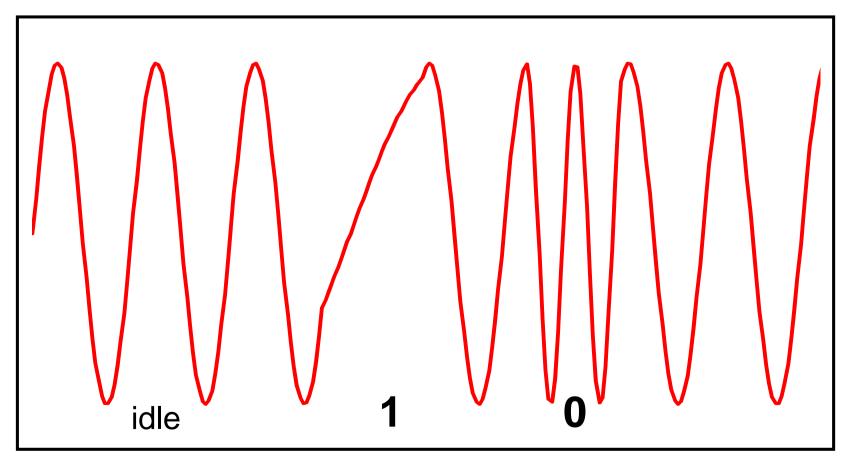
- Original signal
- Amplitude modulation
- Frequency modulation



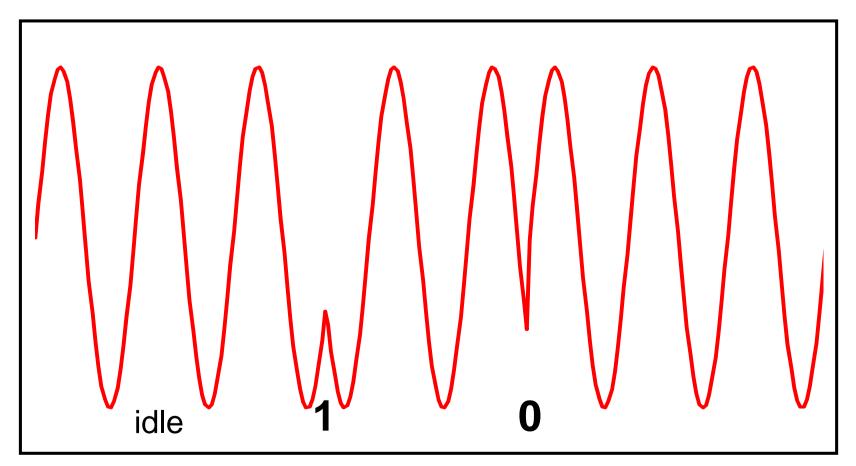
Amplitude Modulation



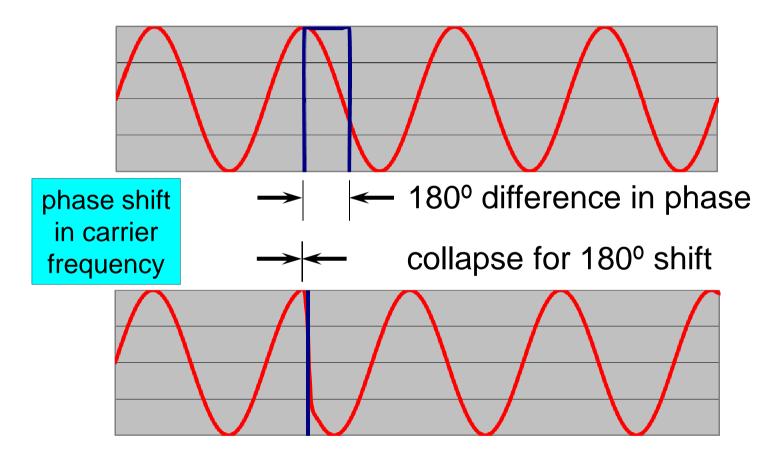
Frequency Modulation



Phase Modulation

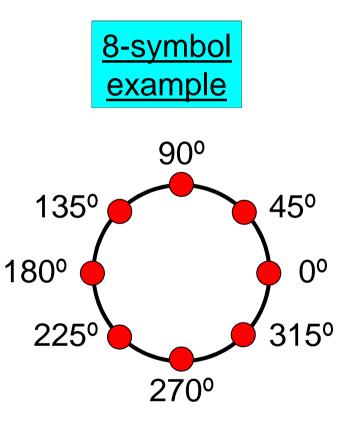


Phase Modulation

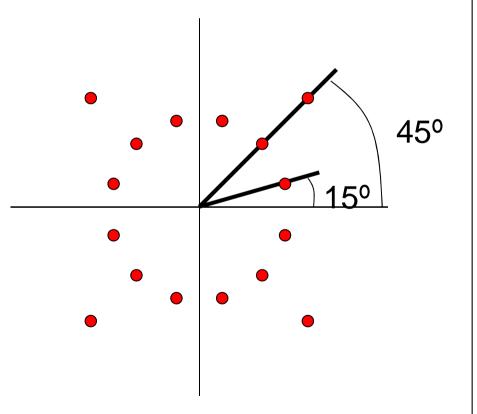


Phase Modulation Algorithm

- Send carrier frequency for one period
 - Perform phase shift
 - Shift value encodes symbol
 - Value in range [0, 360^o)
 - Multiple values for multiple symbols
 - Represent as circle



You can combine modulation schemes



Example: QAM (Quadrature Amplitude Modulation)

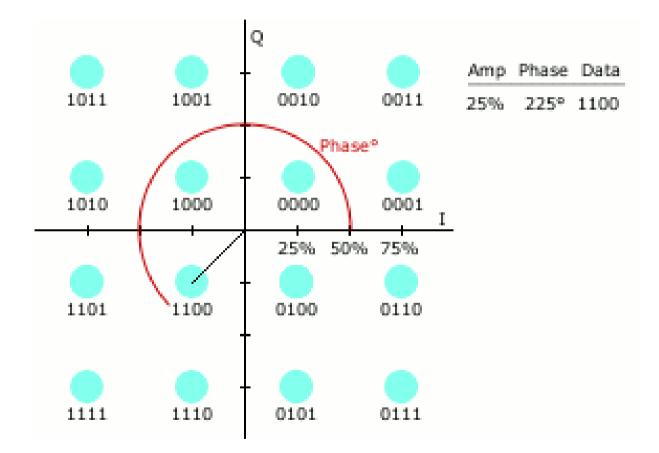
For a given symbol:

• Perform phase shift and change to new amplitude

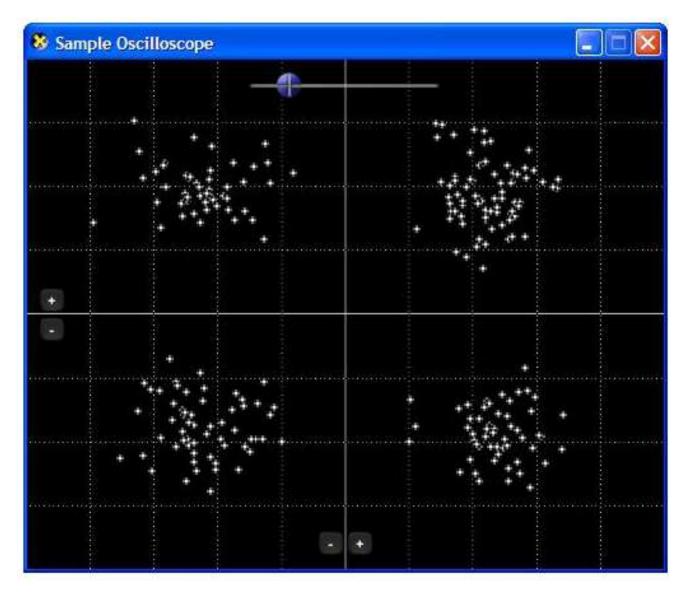
2-dimensional representation:

- Angle is phase shift
- Radial distance is new amplitude

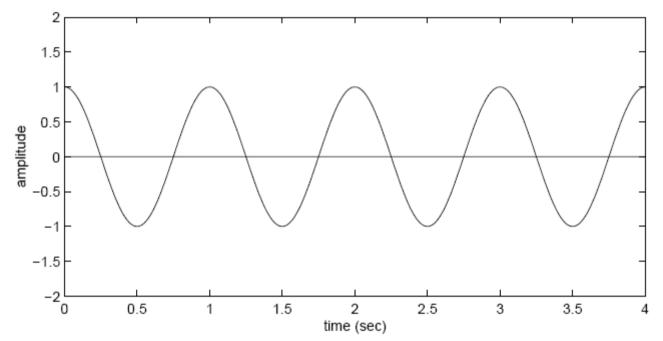
QAM: Example transmission



Real constellation with noise

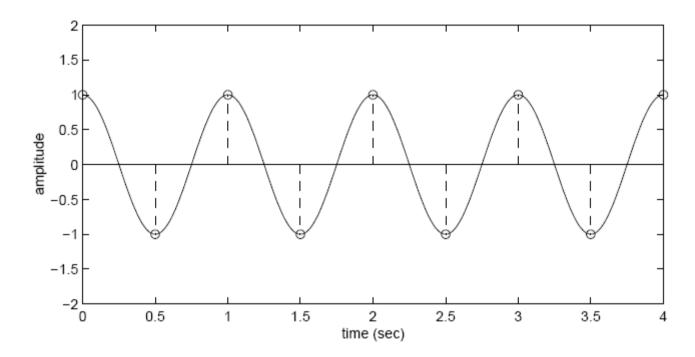






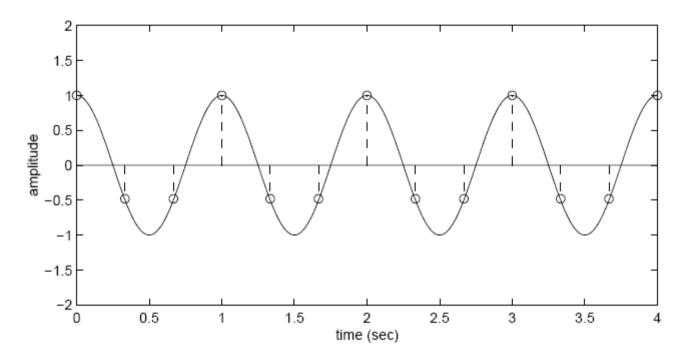
- Suppose you have the following 1Hz signal being received
- How fast to sample, to capture the signal?





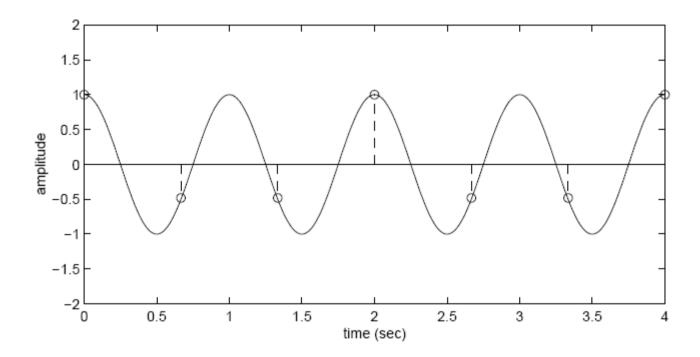
- Sampling a 1 Hz signal at 2 Hz is enough
 - Captures every peak and trough



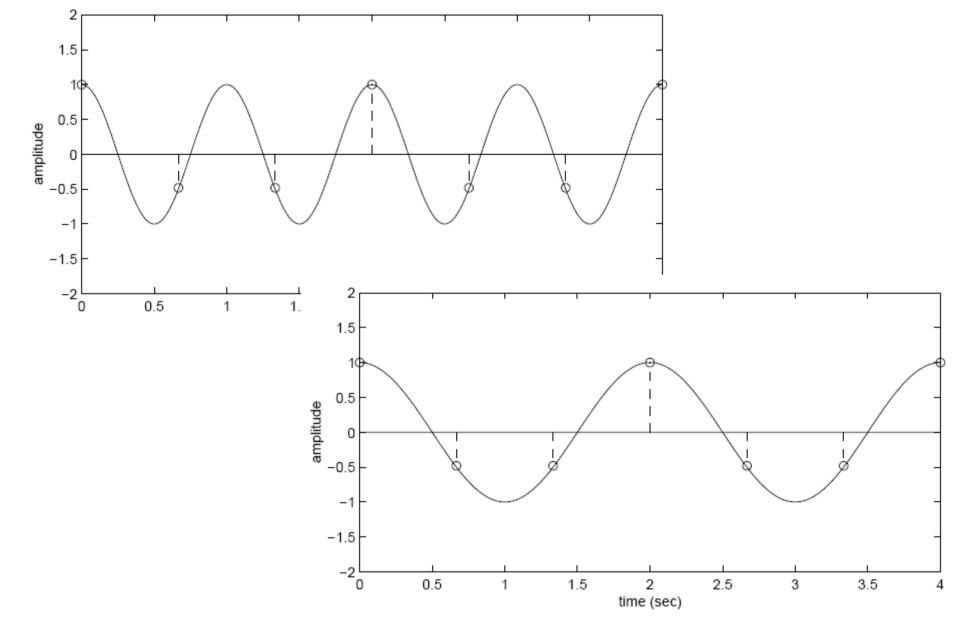


- Sampling a 1 Hz signal at 3 Hz is also enough
 - In fact, more than enough samples to capture variation in signal



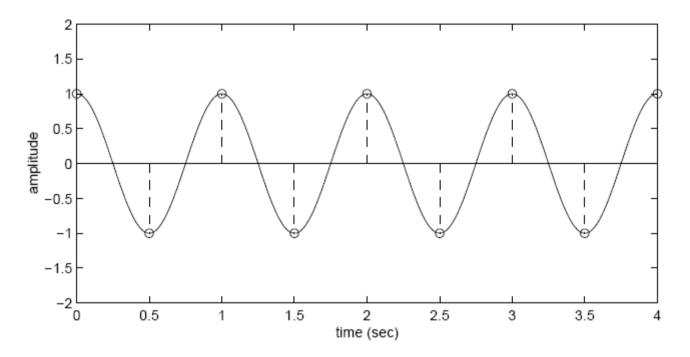


- Sampling a 1 Hz signal at 1.5 Hz is not enough
 - Why?



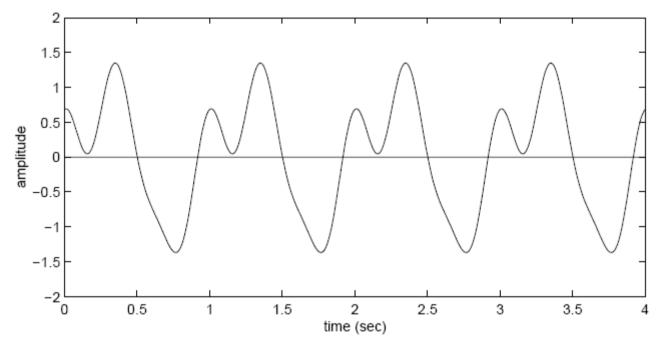
- Sampling a 1 Hz signal at 1.5 Hz is not enough
 - Not enough samples, can't distinguish between multiple possible signals





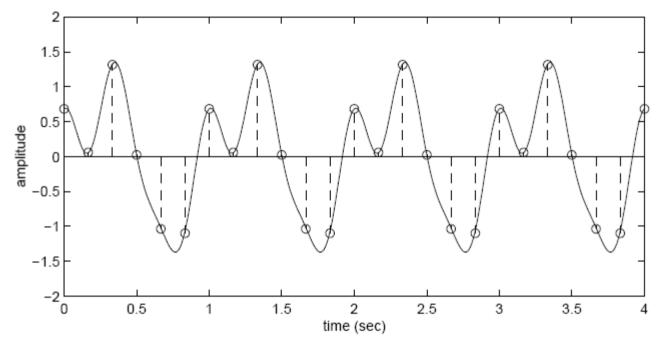
- Sampling a 1 Hz signal at 2 Hz is both necessary and sufficient
- In general: sampling twice rate of signal is enough

What about more complex signals?



- Fourier's theorem: any continuous signal can be decomposed into a sum of sines and cosines at different frequencies
- Example: Sum of 1 Hz, 2 Hz, and 3 Hz sines
 - How fast to sample?

What about more complex signals?



- Fourier's theorem: any continuous signal can be decomposed into a sum of sines and cosines at different frequencies
- Example: Sum of 1 Hz, 2 Hz, and 3 Hz sines
 - How fast to sample?
 - Answer: Twice rate of fastest signal (bandwidth): 6 Hz

Nyquist–Shannon sampling theorem

- If a function x(t) contains no frequencies higher than B hertz, it is completely determined by giving its ordinates at a series of points spaced 1/(2B) seconds apart
- In other words:
 - If the bandwidth of your channel is B
 - Your sampling rate should be 2B
 - Higher sampling rates are pointless
 - Lower sampling rates lead to aliasing/distortion/error

Related Question: How much data can you pack into a channel?

- If I sample at a rate of 2B, I can precisely determine the signal of bandwidth B
- If I have data coming in at rate 2B, I can encode it in a channel of rate B
 - Similar argument to above, but in reverse
 - Instead of "reading" a sample, we "write" a sample
- More generally:
 - Transmitting N distinct signals over a noiseless channel with bandwidth B, we can achieve at most a data rate of
 - 2B log2 N

Noiseless Capacity

- Nyquist's theorem: 2B log₂ N
- Example 1: sampling rate of a phone line
 - B = 4000 Hz
 - 2B = 8000 samples/sec.
 - sample every 125 microseconds
- Example 2: noiseless capacity
 - B = 1200 Hz
 - N = each pulse encodes 16 levels
 - $C = 2B \log_2 (N) = D \times \log_2 (N)$
 - = 2400 x 4 = 9600 bps.

What can Limit Maximum Data Rate?

- Noise
 - E.g., thermal noise (in-band noise) can blur symbols
- Transitions between symbols
 - Introduce high-frequency components into the transmitted signal
 - Such components cannot be recovered (by Nyquist's Theorem), and some information is lost
- Examples
 - Phase modulation
 - Single frequency (with different phases) for each symbol
 - Transitions can require very high frequencies

How does Noise affect these Bounds?

- In-band (thermal, not high-frequency) noise
 - Blurs the symbols, reducing the number of symbols that can be reliably distinguished.
- Claude Shannon (1948)
 - Extended Nyquist's work to channels with additive white Gaussian noise (a good model for thermal noise)

channel capacity $C = B \log_2 (1 + S/N)$

B is the channel bandwidth

S/N is the ratio between

the average signal power and

the average in-band noise power

Noisy Capacity

- Telephone channel
 - 3400 Hz at 40 dB SNR
 - $C = B \log_2 (1+S/N) bits/s$
 - SNR = 40 dB

40 =10 log₁₀ (S/N) S/N =10,000

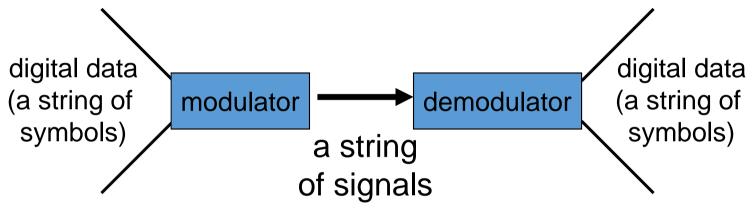
$$\mathrm{SNR}(\mathrm{dB}) = 10 \log_{10} \left(\frac{P_{\mathrm{signal}}}{P_{\mathrm{noise}}} \right)$$

Summary of Encoding

- Problems
 - Attenuation, dispersion, noise
- Digital transmission allows periodic regeneration
- Variety of binary voltage encodings
 - High frequency components limit to short range
 - More voltage levels provide higher data rate
- Carrier frequency and modulation
 - Amplitude, frequency, phase, and combinations
 - Quadrature amplitude modulation: amplitude and phase, many signals
- Nyquist (noiseless) and Shannon (noisy) limits on data rates

Error Detection/Correction

Error Detection



- Encoding translates symbols to signals
- Framing demarcates units of transfer
- Error detection validates correctness of each frame

Error Detection

- Key idea: Add redundant information that can be used to determine if errors have been introduced, and potentially fix them
- Errors checked at many levels
 - Demodulation of signals into symbols (analog)
 - Bit error detection/correction (digital)—our main focus
 - Within network adapter (CRC check)
 - Within IP layer (IP checksum)
 - Possibly within application as well

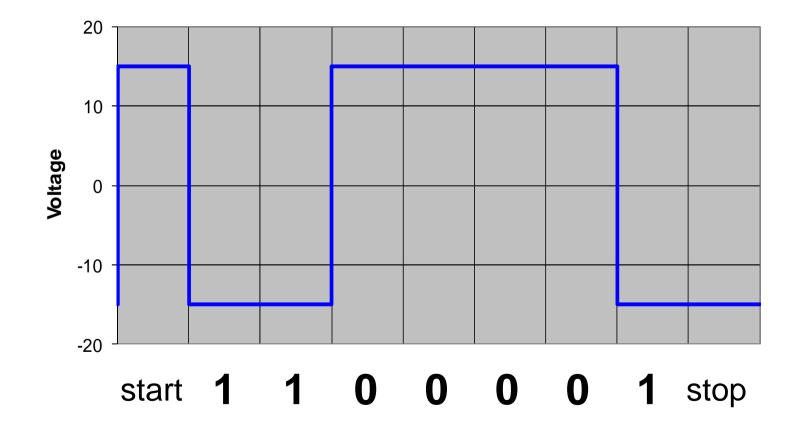
Error Detection

- Analog Errors
 - Example of signal distortion
- Hamming distance
 - Parity and voting
 - Hamming codes
- Error bits or error bursts?
- Digital error detection
 - Two-dimensional parity
 - Checksums
 - Cyclic Redundancy Check (CRC)

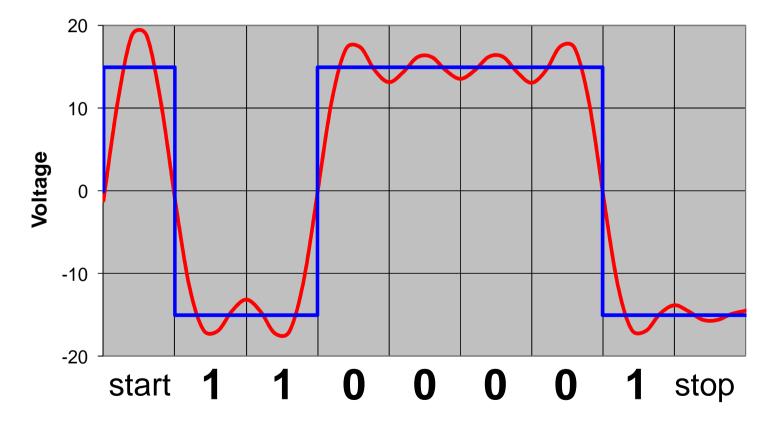
Analog Errors

- Consider RS-232 encoding of character 'Q'
 - ASCII Q = 1100001
- Assume idle wire (-15V) before and after signal

RS-232 Encoding of 'Q'

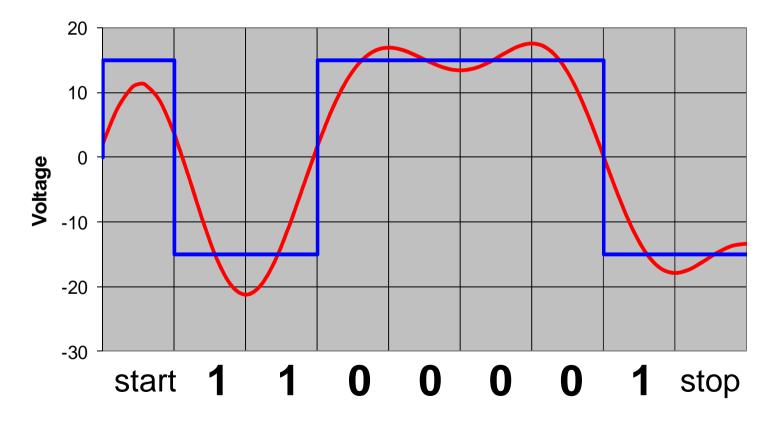


Limited-Frequency Signal Response (bandwidth = baud rate)

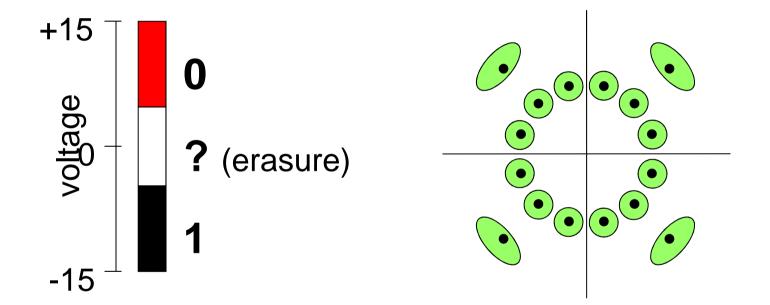


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Limited-Frequency Signal Response (bandwidth = baud rate/2)



Symbols



possible binary voltage encoding possible QAM symbol symbol neighborhoods and erasure neighborhoods in green; all region other space results in erasure

Symbols

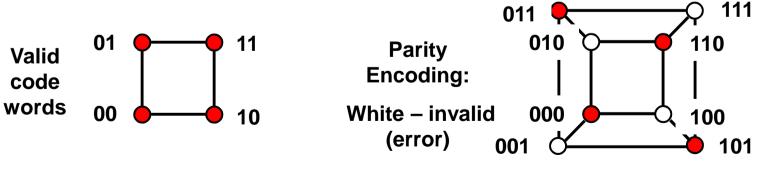
- Inputs to digital level
 - valid symbols
 - erasures
- Hamming distance
 - Definition
 - 1-bit error-detection with parity
 - 1-bit error-correction with voting
 - 2-bit erasure-correction with voting
 - Hamming codes (1-bit error correction)

Hamming Distance

- The Hamming distance between two code words is the minimum number of bit flips to move from one to the other
 - Example:
 - 00101 and 00010
 - Hamming distance of 3

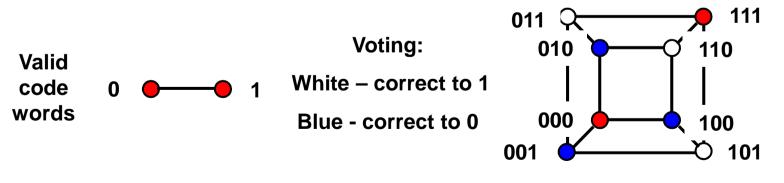
Detecting bit flips with Parity

- 1-bit error detection with parity
 - Add an extra bit to a code to ensure an even (odd) number of 1s
 - Every code word has an even (odd) number of 1s



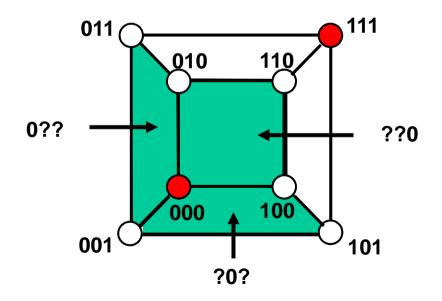
Correcting bit flips with Voting

- 1-bit error correction with voting
 - Every codeword is transmitted n times



2-bit <u>Erasure</u> Correction with Voting

• Every code word is copied 3 times



2-erasure planes in green remaining bit not ambiguous

cannot correct 1-error and 1-erasure

Minimum Hamming Distance

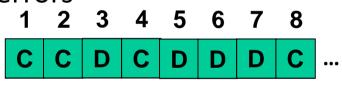
- The minimum Hamming distance of a code is the minimum distance over all pairs of codewords
 - Minimum Hamming Distance for parity
 - 2
 - Minimum Hamming Distance for voting
 - 3

Coverage

- N-bit error detection
 - No code word changed into another code word
 - Requires Hamming distance of N+1
- N-bit error correction
 - N-bit neighborhood: all codewords within N bit flips
 - No overlap between N-bit neighborhoods
 - Requires hamming distance of 2N+1

Hamming Codes

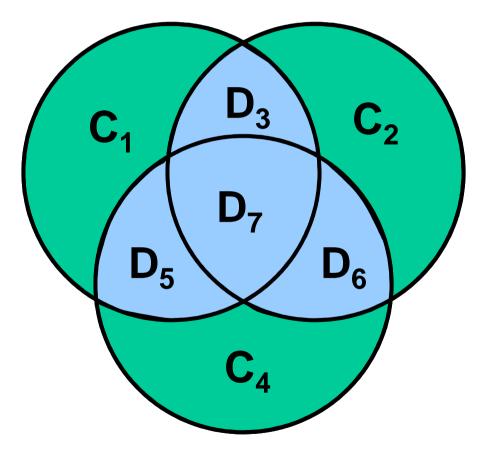
- Linear error-correcting code, Named after Richard Hamming
 - Simple, commonly used in RAM (e.g., ECC-RAM)
- Can detect up to 2 simultaneous bit errors
- Can correct single-bit errors
- Construction
 - number bits from 1 upward
 - powers of 2 are check bits
 - all others are data bits
 - Check bit j is XOR of all bits k such that (j AND k) = j
- Example: 4 bits of data, 3 check bits



Hamming Codes

C1 = D3 XOR D5 XOR D7C2 = D3 XOR D6 XOR D7C4 = D5 XOR D6 XOR D7

Hamming Codes



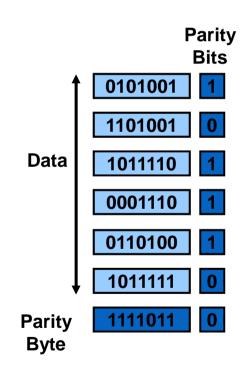
Error Bits or Bursts?

- Common model of errors
 - Probability of error per bit
 - Error in each bit independent of others
 - Value of incorrect bit independent of others
- Burst model
 - Probability of back-to-back bit errors
 - Error probability dependent on adjacent bits
 - Value of errors may have structure
- Why assume bursts?
 - Appropriate for some media (e.g., radio)
 - Faster signaling rate enhances such phenomena

Digital Error Detection Techniques

- Two-dimensional parity
 - Detects up to 3-bit errors
 - Good for burst errors
- IP checksum
 - Simple addition
 - Simple in software
 - Used as backup to CRC
- Cyclic Redundancy Check (CRC)
 - Powerful mathematics
 - Tricky in software, simple in hardware
 - Used in network adapter

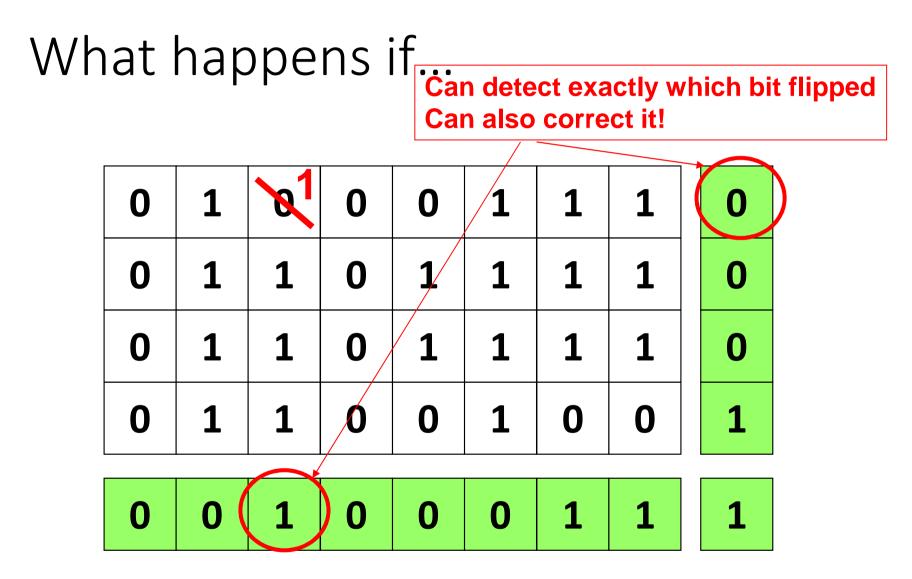
Two-Dimensional Parity

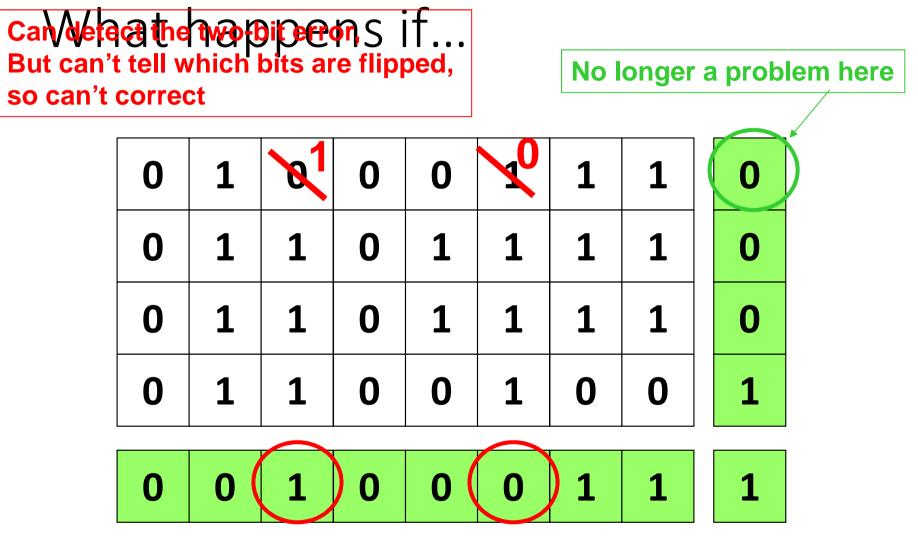


- Use 1-dimensional parity
 - Add one bit to a 7-bit code to ensure an even/odd number of 1s
- Add 2nd dimension
 - Add an extra byte to frame
 - Bits are set to ensure even/odd number of 1s in that position across all bytes in frame
- Comments
 - Can **detect** and **correct** any 1-bit error
 - Can **detect** any 1-, 2- and 3-bit, and most 4-bit errors

Two-Dimensional Parity

0	1	0	0	0	1	1	1	0
0	1	1	0	1	1	1	1	0
0	1	1	0	1	1	1	1	
0	1	1	0	0	1	0	0	1
0	0	1	0	0	0	1	1	1

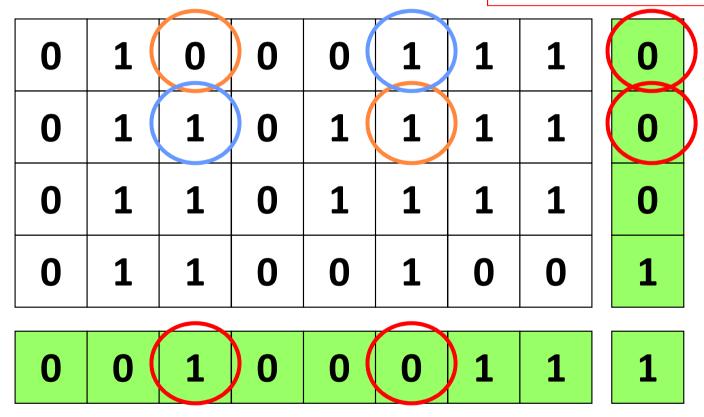




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Suppose these four parity bits don't match Which bits could be in error?

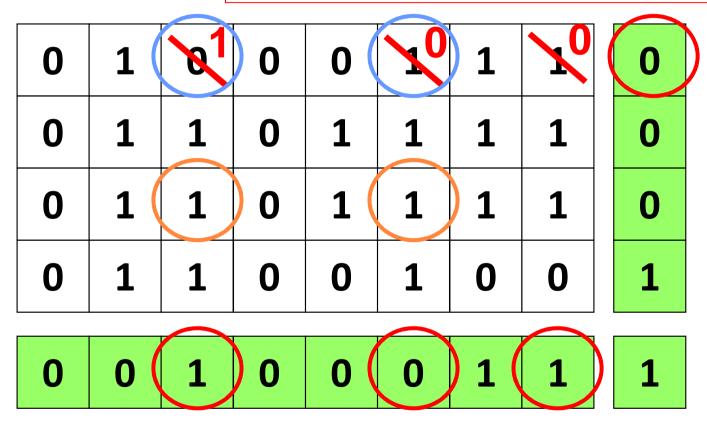
Could be the blue pair, OR, could be the orange pair. So, can't correct.



What about 3-bit errors? Can detect exactly which bit flipped You can correct in this case

0	1	8	0	0	×0	1	10	0
0	1	1	0	1	1	1	1	0
0	1	1	0	1	1	1	1	0
0	1	1	0	0	1	0	0	1
0	0) ()	0 (0) 1 (1
U						_	_	-

What about Gan Odtec exactly which bit flipped But you can't correct (eg if orange bits got flipped instead of the blue ones)



Are there any 4 bit errors this scheme *can* detect?

1	1	81	0	0	1	1	1	0
0	1			1	1	1	1	0
e 1	1	X	0	1	1	1	1	0
0	1	1	0	0	1	0	0	1
0	0	1	0	0	0	1	1	1

Internet Checksum

- Idea: Add up all the words, transmit the sum
- Internet Checksum
 - Use 1's complement addition on 16bit codewords
 - Example

•	Codewords:	-5	-3
---	------------	----	----

- 1's complement binary: 1010 1100
- 1's complement sum 1000

Comments

- Small number of redundant bits
- Easy to implement
- Not very robust

IP Checksum

```
u_short cksum(u_short *buf, int count) {
   register u_long sum = 0;
   while (count--) {
       sum += *buf++;
       if (sum & 0xFFFF0000) {
       /* carry occurred, so wrap around */
              sum \&= 0 \times FFFF;
              sum++;
       }
   return ~(sum & 0xFFFF);
}
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```

Cyclic Redundancy Check (CRC)

- Non-secure hash function based on cyclic codes
- Idea
 - Add **k** bits of redundant data to an **n**-bit message
 - **N**-bit message is represented as a **n**-degree polynomial with each bit in the message being the corresponding coefficient in the polynomial
 - Example
 - Message = 10011010
 - Polynomial

```
= \mathbf{1} * x^{7} + \mathbf{0} * x^{6} + \mathbf{0} * x^{5} + \mathbf{1} * x^{4} + \mathbf{1} * x^{3} + \mathbf{0} * x^{2} + \mathbf{1} * x + \mathbf{0}
= x^{7} + x^{4} + x^{3} + x
```

Overly simplified CRC-like protocol, using regular numbers

- Both endpoints agree in advance on a divisor value C=3
- Sender wants to send a message M=10
- Sender computes a value P=M+X=10+2=12 that is evenly divisible by C
- Sender sends P and M to receiver
- Receiver checks to make sure P=12 is evenly divisible by C=3
 - If it is not, then there's error(s)
 - If it is, then there are probably no errors
- CRC is vaguely like this, but uses polynomials instead of numbers
 - CRC can reconstruct M from P and C, so just needs to send P

CRC Approach

- Given
 - Message M(x) 10011010
 - Represented as $x^7 + x^4 + x^3 + x$
- 1. Select a divisor polynomial C(x) with degree k
 - Example with k = 3:
 - $C(x) = x^3 + x^2 + 1$
 - Represented as 1101
- 2. Transmit a polynomial P(x) that is evenly divisible by C(x)
 - P(x) = M(x) + k bits

How can we determine these k bits?

Properties of Polynomial Arithmetic

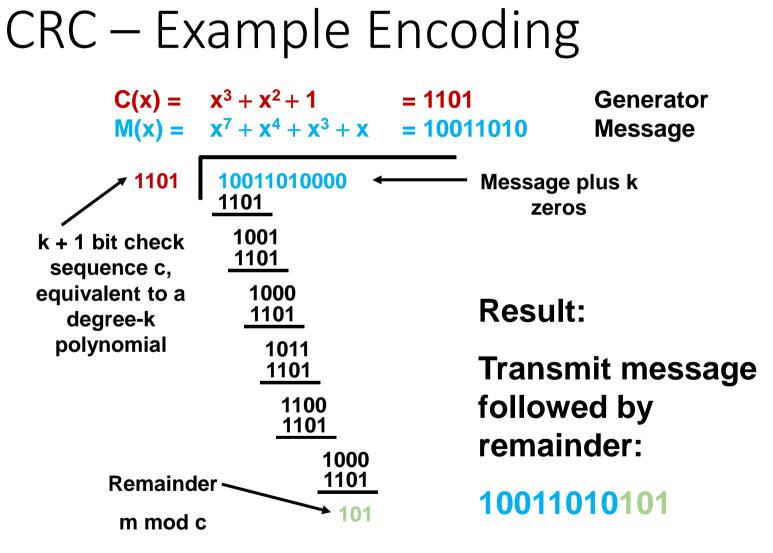
- Divisor
 - Any polynomial B(x) can be divided by a polynomial C(x) if B(x) is of the same or higher degree than C(x)
- Remainder
 - The remainder obtained when B(x) is divided by C(x) is obtained by subtracting C(x) from B(x)
- Subtraction
 - To subtract C(x) from B(x), simply perform an XOR on each pair of matching coefficients
- For example: $(x^3+1)/(x^3+x^2+1) =$

CRC - Sender

- Given
 - $M(x) = 10011010 = x^7 + x^4 + x^3 + x$
 - $C(x) = 1101 = x^3 + x^2 + 1$
- Steps
 - T(x) = M(x) * x^k (add zeros to increase degree of M(x) by k)
 - Find remainder, R(x), from T(x)/C(x)
 - $P(x) = T(x) R(x) \Rightarrow M(x)$ followed by R(x)
- Example
 - T(x) = 10011010000
 - R(x) = 101
 - P(x) = 10011010101

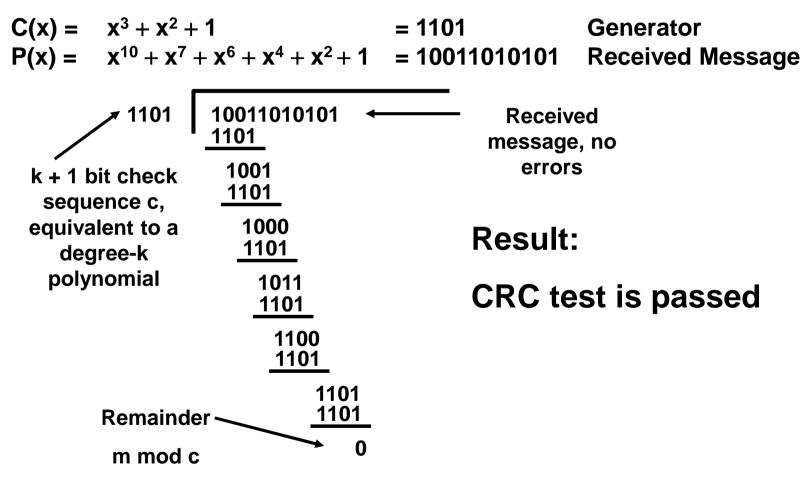
CRC - Receiver

- Receive Polynomial P(x) + E(x)
 - E(x) represents errors
 - (if no errors then E(x) = 0)
- Divide (P(x) + E(x)) by C(x)
 - If result = 0, either
 - No errors (E(x) = 0, and P(x) is evenly divisible by C(x))
 - (P(x) + E(x)) is exactly divisible by C(x), error will not be detected



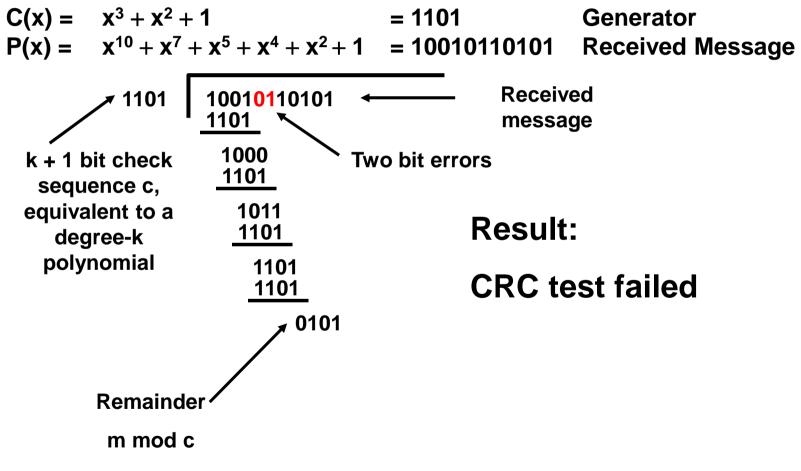
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CRC – Example Decoding – No Errors



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CRC – Example Decoding – with Errors

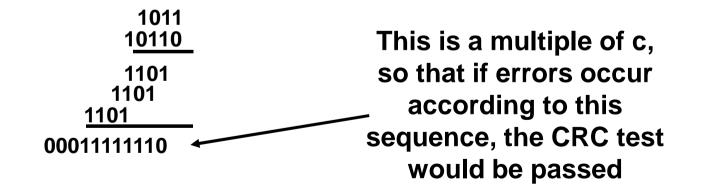


CRC Error Detection

- Properties
 - Characterize error as E(x)
 - Error detected unless C(x) divides E(x)
 - (*i.e.*, E(x) is a multiple of C(x))

Example of Polynomial Multiplication

- Multiply
 - 1101 by 10110
 - $x^3 + x^2 + 1$ by $x^4 + x^2 + x$



On Polynomial Arithmetic

- Polynomial arithmetic
 - A fancy way to think about addition with no carries.
 - Helps in the determination of a good choice of C(x)
 - A non-zero vector is not detected if and only if the error polynomial E(x) is a multiple of C(x)
- Implication
 - Suppose C(x) has the property that C(1) = 0 (i.e. (x + 1) is a factor of C(x))
 - If E(x) corresponds to an undetected error pattern, then it must be that E(1) = 0
 - Therefore, any error pattern with an odd number of error bits is detected

CRC Error Detection

- What errors can we detect?
 - All single-bit errors, if x^k and x⁰ have non-zero coefficients
 - All double-bit errors, if C(x) has at least three terms
 - All odd bit errors, if C(x) contains the factor (x + 1)
 - Any bursts of length < k, if C(x) includes a constant term
 - Most bursts of length $\geq k$

Common	Polynomials for C(x)
CRC	C(x)
CRC-8	$x^8 + x^2 + x^1 + 1$
CRC-10	$x^{10} + x^9 + x^5 + x^4 + x^1 + 1$
CRC-12	$x^{12} + x^{11} + x^3 + x^2 + x^1 + 1$
CRC-16	$x^{16} + x^{15} + x^2 + 1$
CRC-CCITT	$x^{16} + x^{12} + x^5 + 1$
CRC-32	$\begin{array}{c} x^{32} + x^{26} + x^{23} + x^{22} + x^{16} + x^{12} + x^{11} + x^{10} + x^8 + x^7 + x^5 + \\ x^4 + x^2 + x^1 + 1 \end{array}$