Vocoder	Random Signals	Power Spectrum	Autocorrelation	Bandpass	Synthesis	Conclusions

Lecture 5, The Vocoder, Part 2: Unvoiced Sounds

Mark Hasegawa-Johnson

ECE 537: Speech Processing Fundamentals

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- 6 Speech Synthesis: Adjusting the Level of Each Band

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Conclusions

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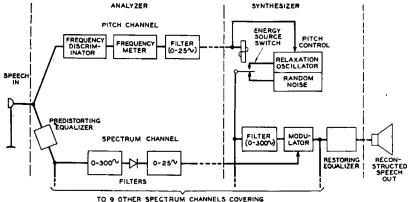
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Conclusions



The Vocoder Block Diagram



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FIG. 2. Schematic arrangement of the Vocoder.

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		Summarv				
Vocoder ○O●	Random Signals 00000	Power Spectrum	Autocorrelation	Bandpass 000000	Synthesis 000000	Conclusions

- What is the spectrum of a "relaxation oscillator"?
 - To answer this question, we need to learn about the Discrete-Time Fourier Series (DTFS).
 - What happens when you bandpass filter it?
 - What happens when you adjust its level?
- What is the spectrum of a "random noise"?
 - To answer this question, we need to learn about autocorrelation and the power spectrum.

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- What happens when you bandpass filter it?
- What happens when you adjust its level?

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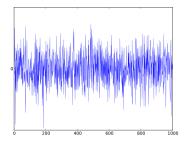
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7 Conclusions

Vocoder 000	Random Signals ○●000	Power Spectrum	Autocorrelation	Bandpass 000000	Synthesis 000000	Conclusions 00
Rando	om Signals					

Let's start out with a zero-mean random signal, x[n].

- Random signal: each x[n] is a random number.
- Zero mean: *E*[*x*[*n*]] = 0 (for all *n*).



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Vocader Random Signals Power Spectrum Autocorrelation Bandpass Synthesis Conclusions 000 0000000 0000000 0000000 0000000 0000000 00 Properties a Random Signal Might Have 0000000 0000000 0000000 00 00

- A random signal is **zero-mean** if E[x[n]] = 0 for all *n*.
- A random signal is **unit-power** if $E\left[|x[n]|^2\right] = 1$, regardless of *n*.

• A random signal is white noise, a.k.a. uncorrelated if $E[x[n]x[m]^*] = 0$ for all $n \neq m$.

Vocoder Random Signals Power Spectrum Autocorrelation Bandpass Synthesis Conclusions 000 00 0000000 0000000 0000000 0000000 00 Wide-Sense Stationary Signals Signals 0000000 0000000 0000000 0000000 00

A random signal is called "wide-sense stationary (WSS)" if its mean, variance, and covariance are independent of n:

- $E[x[n]] = \mu_x$, regardless of n.
- $E\left[|x[n] \mu_x|^2\right] = \sigma_x^2$, regardless of n.
- The **autocorrelation** and **autocovariance** of a WSS signal are defined to be

$$R_{xx}[m] = E[x[n]x^*[n-m]]$$

$$K_{xx}[m] = E[(x[n] - \mu_x)(x[n-m] - \mu_x)^*],$$

regardless of n.



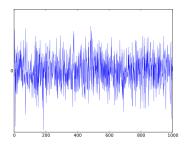
We'll often use zero-mean, unit-variance white noise as a building block:

•
$$E[x[n]] = 0$$

•
$$R_{xx}[m] = \delta[m] =$$

$$\begin{cases} 1 \quad m = 0 \\ 0 \quad m \neq 0 \end{cases}$$

Note: if we add one more assumption (x[n] is Gaussian), then it's also true that x[n] are i.i.d.



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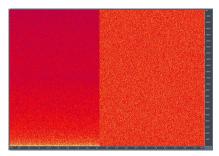
7 Conclusions



The Fourier Transform of a random signal is a random vector.

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

- e^{-jωn} is a constant
- x[n] is random
- X(ω) is the weighted sum of the random variables x[n]



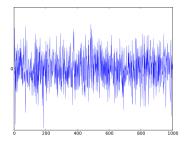
Spectrogram of pink noise (left) and white noise (right), shown with linear frequency axis (vertical). CC-SA 3.0, https: //commons.wikimedia.org/wiki/File:Noise.jpg

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The Fourier Transform of a zero-mean random signal is a zero-mean random vector.

$$E[X(\omega)] = E\left[\sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}\right]$$
$$= \sum_{n=-\infty}^{\infty} E[x[n]]e^{-j\omega n}$$
$$= 0$$



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The magnitude-squared Fourier Transform is also a random variable, but its expected value is not zero.

$$E\left[|X(\omega)|^{2}\right] = E\left[\left(\sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}\right)\left(\sum_{m=-\infty}^{\infty} x[m]e^{-j\omega m}\right)^{*}\right]$$
$$= \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} E\left[x[n]x^{*}[m]\right]e^{-j\omega(n-m)}$$
$$= \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} R_{xx}[n-m]e^{-j\omega(n-m)}$$

Spectrogram of pink noise (left) and white noise (right), shown with linear frequency axis (vertical).

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For most signals, the formula on the previous slide gives $E\left[|X(\omega)|^2\right] \rightarrow \infty$. To make it easier to work with, Norbert Wiener defined the power spectrum to be the time-normalized expected value of the magnitude squared Fourier transform:

$$R_{xx}(\omega) = \lim_{N \to \infty} \frac{1}{N} E \left[\left| \sum_{n=-\left(\frac{N-1}{2}\right)}^{\left(\frac{N-1}{2}\right)} x[n] e^{-j\omega n} \right|^2 \right]$$



Most practical signals are not infinite length. Instead, we usually want to just compute the Fourier transform over N samples, say, $0 \le n \le N-1$. In this case we can define the short-time power spectrum to be

$$R_{xx}(\omega) = rac{1}{N} E\left[\left| \sum_{n=0}^{N-1} x[n] e^{-j\omega n} \right|^2
ight]$$

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Vocoder Random Signals Power Spectrum Autocorrelation Bandpass Synthesis Conclusions 000 00000 0 0000000 0000000 0000000 00

For example, consider white noise: E[x[n]x[m]] = 0 unless n = m. In this case,

$$R_{xx}(\omega) = \frac{1}{N} E\left[|X(\omega)|^2\right]$$

= $\frac{1}{N} \sum_{n=0}^{N-1} \sum_{m=0}^{N-1} E[x[n]x^*[m]] e^{-j\omega(n-m)}$
= $\frac{1}{N} \sum_{n=0}^{N-1} E[|x[n]|^2]$
= $E[|x[n]|^2]$

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Vocoder Random Signals Power Spectrum Autocorrelation Bandpass Synthesis Conclusions 0000 000000 0000000 0000000 0000000 0000000 00 Example: Power Spectrum of White Noise

For example, consider white noise: E[x[n]x[m]] = 0 unless n = m. In this case,

$$R_{xx}(\omega) = E\left[|x[n]|^2\right]$$

This is why we call it white noise: its power spectrum is a constant, $R_{xx}(\omega) = E[|x[n]|^2]$, at every frequency. For example, for zero-mean unit-variance white noise, $R_{xx}(\omega) = E[|x[n]|^2] = \sigma_x^2 = 1$.



Spectrogram of pink noise (left) and white noise (right), shown with linear frequency axis (vertical).

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7 Conclusions

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Power Spectrum of a WSS Signal

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Remember that WSS signals have an autocorrelation function that doesn't depend on n:

$$R_{xx}[m] = E[x[n]x^*[n-m]]$$

For a WSS signal, it's possible to use a dramatic shortcut to compute the power spectrum:

$$R_{xx}(\omega) = \frac{1}{N} E\left[\left| \sum_{n} x[n] e^{-j\omega n} \right|^{2} \right]$$

$$= \frac{1}{N} E\left[\left(\sum_{n} x[n] e^{-j\omega n} \right) \left(\sum_{n-m} x[n-m] e^{-j\omega(n-m)} \right)^{*} \right]$$

$$= \frac{1}{N} \sum_{n} \sum_{m} E[x[n] x^{*}[n-m]] e^{-j\omega m}$$

$$= \sum_{m} R_{xx}[m] e^{-j\omega m}$$



Let me just repeat that, since it's the most important formula today.

$$R_{xx}(\omega) = \sum_{m} R_{xx}[m] e^{-j\omega m}$$

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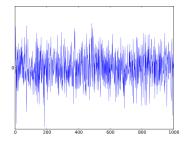


For example, consider white noise:

$$R_{xx}[m] = \delta[m] = \begin{cases} 1 & m = 0\\ 0 & m \neq 0 \end{cases}$$

So its power spectrum is

$$R_{xx}(\omega) = \mathcal{F}\left\{\delta[m]\right\} = 1$$



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Brownian motion, as shown in the video, is motion with independent random increments, i.e., if x[n] is the position and v[n] is an independent increment, then

x[n] = ax[n-1] + bv[n]

Natural Brownian motion uses a = b = 1, but if we want a WSS signal, we need to use $b^2 = 1 - a^2$.

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Suppose we assume v[n] is zero-mean unit-variance white noise, $b^2 = 1 - a^2$, and x[n] = ax[n-1] + bv[n], so that

$$E[x[n]x[n-1]] = E[(ax[n-1] + bv[n])x[n-1]] = a$$

$$E[x[n]x[n-2]] = E[(a^{2}x[n-2] + abv[n-1] + bv[n])x[n-2]] = a^{2}$$

$$R_{xx}[m] = a^{\mid m}$$

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The power spectrum is

$$R_{\mathrm{xx}}(\omega) = \mathcal{F}\left\{a^{|m|}
ight\} = rac{b^2}{\left|1 - ae^{-j\omega}
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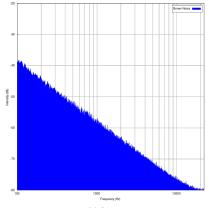
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• Impulse trains and white noise both have flat spectra.

 $|X_k|^2 = R_{xx}(\omega) = 1$

 Square waves and Brownian motion both have Brownian spectra.

$$|X_k|^2 = R_{xx}(\omega) = \frac{1}{\mathcal{O}\left\{\omega^2\right\}}$$



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Autocorrelation isn't a function of n, so it doesn't hurt if we average it over many samples of n:

$$R_{xx}[m] = \frac{1}{N} \sum_{n=0}^{N-1} R_{xx}[m] = \frac{1}{N} E\left[\sum_{n=0}^{N-1} x[n] x^*[n-m]\right]$$
$$= \frac{1}{N} E[x[m] * x^*[-m]]$$

• Convolution: Flip, shift, multiply, and add:

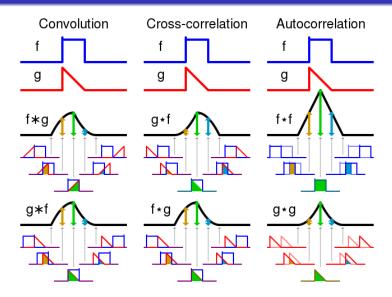
$$x[m] * h[m] = \sum_{n} x[n]h[m-n]$$

• Correlation: DON'T flip. Just shift, multiply and add:

$$x[m] * h^*[-m] = \sum_n x[n]h^*[n-m]$$

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Vocoder Random Signals Power Spectrum Autocorrelation Bandpass Synthesis Conclusions Correlation: Shift, Multiply and Add Add Shift, Shift,



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Facts	about con	volution				
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Convolution is commutative:

$$h[n] * x[n] = x[n] * h[n]$$

It is also associative:

$$g[n] * (x[n] * h[n]) = (g[n] * x[n]) * h[n]$$

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The Vocoder Block Diagram

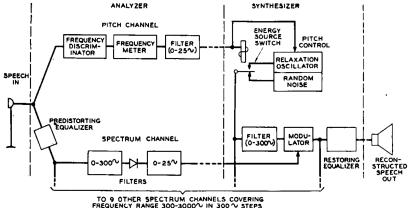


FIG. 2. Schematic arrangement of the Vocoder.

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Spectrum of a Bandpass-Filtered Noise

Suppose

$$y[n] = h[n] * x[n]$$

The autocorrelation of y[n] is defined to be $R_{yy}[m] = E[y[n]y^*[n-m]]$. But remember we can estimate it using the **short-time autocorrelation**:

$$R_{yy}[n] = \frac{1}{N} E[y[n] * y^*[-n]]$$

= $\frac{1}{N} E[h[n] * x[n] * h^*[-n] * x^*[-n]]$
= $\frac{1}{N} (h[n] * h^*[-n] * E[x[n] * x^*[-n]])$
= $h[n] * h^*[-n] * R_{xx}[n]$

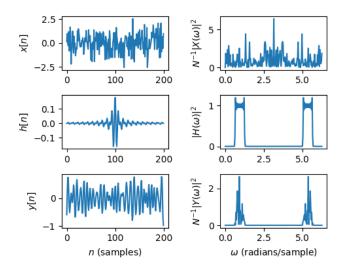
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Spectrum of a Bandpass-Filtered Noise

$$R_{yy}[n] = h[n] * h^*[-n] * R_{xx}[n]$$
$$R_{yy}(\omega) = |H(\omega)|^2 R_{xx}(\omega)$$

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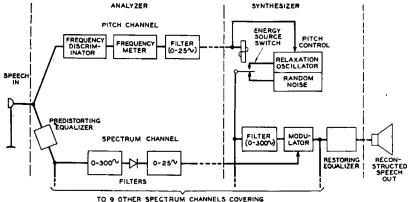
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Conclusions



The Vocoder Block Diagram



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FIG. 2. Schematic arrangement of the Vocoder.

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Suppose that x[n] and y[n] are two uncorrelated random signals, and we add them together:

$$z[n] = ax[n] + by[n]$$

What are the autocorrelation and power spectrum of z[n]?

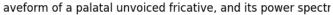
$$\begin{aligned} R_{zz}[m] &= E\left[z[n]z^*[n-m]\right] \\ &= E\left[(ax[n]+by[n])\left(a^*x^*[n-m]+b^*y^*[n-m]\right)\right] \\ &= |a|^2 R_{xx}[m]+|b|^2 R_{yy}[m], \end{aligned}$$

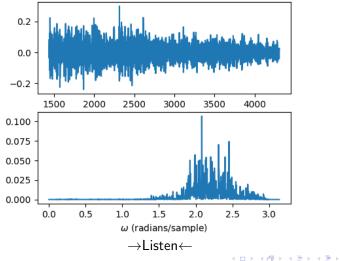
and

$$R_{zz}(\omega) = |a|^2 R_{xx}(\omega) + |b|^2 R_{yy}(\omega)$$

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Start with white noise,

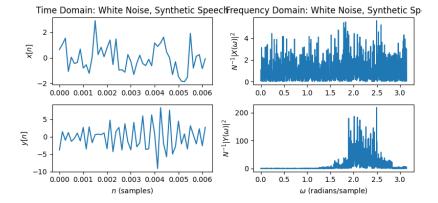
$$R_{xx}(\omega) = 1$$

Filter by a set of 10 bandpass filters $H_I(\omega)$, each about 300Hz wide, then adjust the amplitude of each one (A_I) to match the amplitude of the speech signal in the same band:

$$R_{yy}(\omega) = \sum_{l=1}^{10} A_l \sum_{k=0}^{N-1} |H_l(\omega)|^2 R_{xx}(\omega)$$

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 Conclusions: How to scale the bands of a power spectrum to make fricatives

- White noise has an autocorrelation of R_{xx}[m] = δ[m], and a power spectrum of R_{xx}(ω) = 1.
- Onvolution:

 $y[n] = h[n] * x[n] \quad \leftrightarrow \quad R_{yy}[m] = h[m] * h^*[-m] * R_{xx}[m]$

Linearity:

 $z[n] = ax[n] + by[n] \quad \leftrightarrow \quad R_{zz}[n] = |a|^2 R_{xx}[n] + |b|^2 R_{yy}[n]$